AUGUST 1981 / \$2.50

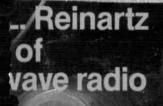




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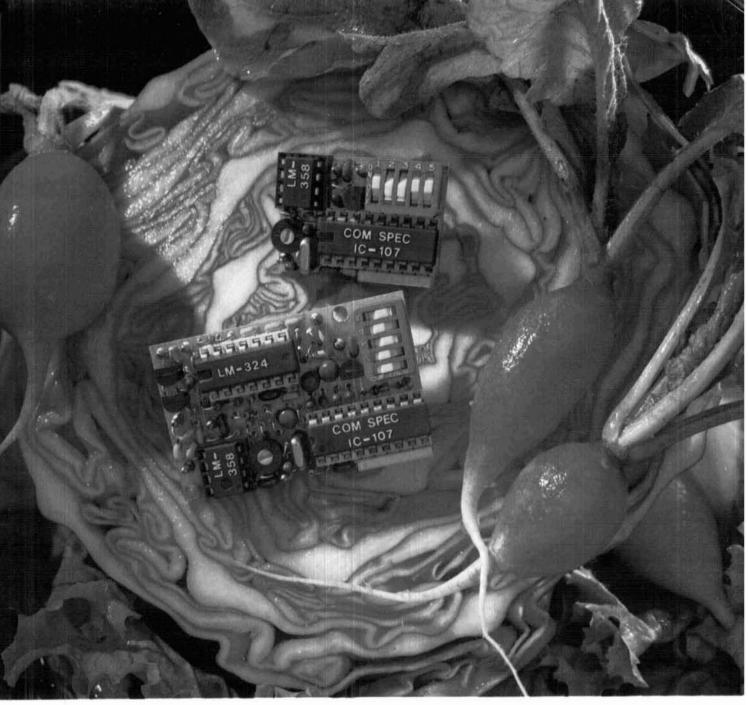
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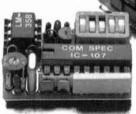


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MFJ Super Keyboards



5 MODES: CW, Baudot, ASCII, memory keyer, Morse code practice. **TWO MODELS:** MFJ-496, **\$339.95.** 256 character buffer, 256 character message memory, automatic messages, serial numbering, repeat/delay. MFJ-494, **\$279.95.** 50 character buffer, 30 character memory, automatic messages.

MFJ brings you a pair of 5 Mode Super Keyboards that gives you more features per dollar than any other keyboard available. You can send CW, Baudot, ASCII. Use it as a memory keyer and for MORSE code practice.

You get text buffer, programmable and automatic message memories, error deletion, buffer preload, buffer hold, plus much more.

MODE 1: CW

The 256 character (50 for 494) text buffer makes sending perfect CW effortless even if you "hunt and peck."

You can preload a message into the buffer and transmit when ready. For break in, you can stop the buffer, send comments on key paddles and then resume sending the buffer content.

Delete errors by backspacing.

A meter gives buffer remaining or speed. Two characters before buffer full the meter lights up red and the sidetone changes pitch.

Four programmable message memories (2 for 494) give a total of 256 characters (30 for 494). Each message starts after one ends for no wasted memory. Delete errors by backspacing.

To use the **automatic messages**, type your call into message A. Then by pressing the CO button you send CO CO DE (message A).

The other automatic messages work the same way: CO TEST DE, DE, ORZ.

Special keys for KN. SK, BT. AS, AA and AR. A lot of thought has gone into human engineering these MFJ Super Keyboards.

For example, you press only a one or two key sequence to execute any command.

All controls and keys are positioned logically and labeled clearly for instant recognition.

Pots are used tor speed, volume, tone, and

weight because they are more human oriented than keystroke sequences and they remember your settings when power is off.

Weight control makes your signal distinctive to penetrate QRM.

MODE 2 & 3 (RTTY): BAUDOT & ASCII

5 level Baudot is transmitted at 60 WPM. Both RTTY and CW ID are provided.

Carriage return, line feed, and "LTRS" are sent automatically on the first space after 63 characters on a line. This gives unbroken words at the receiving end and frees you from sending the carriage return. After 70 characters the function is initiated without a space.

All up and down shift is done automatically. A downshift occurs on every space to quickly clear parbled reception.

The buffer, programmable and automatic messages, backspace delete and PTT control (keys your rig) are included.

The ASCII mode includes all the features of Baudot. Transmission speed is 110 baud. Both upper and lower case are generated.

MODE 4: MEMORY KEYER

Plug in a paddle to use it as a deluxe full feature memory keyer with automatic and programmable memories, iambic operation, dot-dash memories, and all the features of the CW mode.

MODE 5: MORSE CODE PRACTICE

There are two Morse code practice modes. Mode 1: random length groups of random characters. Mode 2: pseudo random 5 character groups in 8 separate repeatable lists (with answers).

Insert space between characters and groups to form high speed characters at slower speed for easy character recognition. Select alphabetic or alphanumeric plus punctuation. You can even pause and then resume.

MORE FEATURES

Automatic incrementing serial number from 0 to 999 can be inserted into buffer or message memory for contests.

Repeat function allows repetition of any message memory with 1 to 99 seconds delay. Lets you call CO and repeat until answered.

Two key lockout operation prevents lost characters during typing speed bursts.

Clock option (496 only) send time in CW, Baudot, ASCII. 24 hour format.

Set CW sending speed before or while sending.

Tune switch with LED keys transmitter for tuning. Tune key provides continuous dots to save finals. Built-in sidetone and speaker.

PTT (push-to-talk) output keys transmitter for Baudot and ASCII modes.

Reliable solid state keying for CW: grid block, cathode, solid state transmitters (-300V, 10 ma Max, + 300V, 100 ma Max). TTL and open collector outputs for RTTY and ASCII.

Fully shielded. RF proof. All aluminum cabinet. Black bottom, eggshell white top. 12"Dx7"Wx11/4"H (front) x31/2"H (back). Red LED indicates on.

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Every single unit is tested for performance and inspected for quality. Solid American construction.

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volume 14, number 8

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ham radio magazine is published monthly by Communications Technology, Inc Greenville, New Hampshire 03048 Telephone: 603-878-1441

subscription rates

United States: one year, \$16.50 two years, \$28.50: three years, \$38.50 Canada and other countries (via Surface Mail) one year, \$21.50; two years, \$40.00 three years, \$57.00

Europe, Japan, Africa (via Air Forwarding Service) one year, \$28.00 All subscription orders payable in United States funds, please

foreign subscription agents

Foreign subscription agents are listed on page 77

Microfilm copies are available from University Microfilms, International Ann Arbor, Michigan 48106 Order publication number 3076

Cassette tapes of selected articles from ham radio are available to the blind and physically handicapped from Recorded Periodicals 919 Walnut Street, 8th Floor Philadelphia, Pennsylvania 19107

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Second-class postage paid at Greenville, N.H. 03048 and at additional mailing offices ISSN 0148-5989

Postmaster send Form 3579 to *hem radio* Greenville, New Hampshire 03048





Automatic Load Control (ALC) was discussed briefly in last month's "Questions and Answers" column. Judging from some of the illegal SSB signals on the air, it seems that ALC is still a mystery to some and should be explored a little more thoroughly. Our guest editorial is by the author of just such an article, which appears in this issue. I'd like to introduce John Weber, K4JW.

Alf Wilson, W6NIF, Editor

To me, it's becoming more and more apparent that some of our phone bands are becoming impossible to enjoy, especially on weekends, because of heavy QRM. Being a "professional loafer," I can enjoy most of my phone operation during the week, usually keeping schedules with old friends around the country.

We have to keep some of these schedules in the General portion of 20 meters because some of our group are restricted to this portion of the band (not to say that the rest of the band is much different!), and many times we are forced to use our linears to plow through the adjacent QRM; thus we too add to the overall QRM level. Although ours is a technical hobby, it gives old friends a chance to meet and talk — no earth-shattering technical or philosophical discussions, just friendly get-togethers. But, when somebody "buckshots" from 4 to 7 kHz away, the enjoyment rapidly disappears.

Just think what a mess we would have if this were still a-m instead of SSB! While NBVM (which hasn't caught on) might help, and future spread-spectrum techniques will allow us to hop, skip, and jump all over the band with minimum QRM, I am afraid we must still deal with the situation as it is today.

One of our problems has been receiver design: old receivers cannot survive alongside present-day linears and high-gain directive antennas, with IMD being the problem usually impeding copy. Fortunately, manufacturers have made considerable improvements to contemporary receivers; with passband shifters, variable passband, and high-performance filters, it is possible to copy signals that would have been impossible to read on older receivers. I hope we all enhance the ability of our receivers to reject or minimize adjacent QRM by using the rf attenuator and backing off on the rf gain when conditions warrant.

Even so, when an Amateur runs his rig "wide-open," fails to monitor his signal output, and shouts into the mike, it gets pretty rough to tweak him out. For some reason, otherwise clean signals get very broad when the operator is running phone patches, suggesting somebody is not paying attention to modulation levels. Of course, these are operator-control problems. One problem it's possible to have and not be aware of is an incompatibility of ALC characteristics between the exciter and linear amplifier. This can occur because the exciter and linear were made by different manufacturers, or because they are of different generations: the 15-year-old linear and the 1981 transceiver. And, of course, maybe the ALC line isn't even connected — that's really an operator-control problem.

I have been using a term not well understood by some Amateurs – ALC. Somebody said it is not even mentioned any more in the *ARRL Handbook*. I think if you look under "Speech Processing" in the chapter on SSB you will find it; it did disappear from the index for some reason. ALC is a form of speech processing.

Now, if you would like to know more about ALC, I suggest you read the article entitled, "An Analysis of ALC Circuits" in this issue. With luck, it will help reduce splatter on the bands, and perhaps it will prod manufacturers into paying more attention to the design problem and into providing us with meaningful numbers so at least we know whether or not a "compatibility" problem exists.

John P. Weber, Jr., K4JW

ICON IC-290A The Latest State of the Art in 2 Meter Mobile



5 Memories/Priority/ Scan/Squelch on SSB.

FM Ease.

Five memories + 2 VFO'S - store your favorite repeaters.

Priority channel - check your most important frequency automatically.

Programmable offsets - for odd repeater splits.
 5 KHz or 1 KHz tuning.

SSB/CW Convenience.

Squelch on SSB - silently scan for signals.
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■ NB - Noise blanker - suppresses pulse type noises on SSB/CW.

Full Capability Scanning.

Scan the whole band/scan between VFO's/scan memories and VFO's.

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 - □ Compact size 6 11/16W×2 1/2H×8 5/8D.



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intruder watch Dear HR:

In regard to W5SAD's letter on Amateur band intruders (October, 1980), W5SAD and others may find, as I have, that the following message sent in CW on the frequency that Russian intruders are using will often get them to shift out of the band:

U/TA OEASTOTA OEASTX MEVDUNARODNOJ L/MBITELXSKOJ POLOSY OEASTOT POVALUJSTA QSY TOTOEASVE

The italicized OE's, UI, and IM are run together to form the additional Cyrillic alphabet characters.

This message when taken down in Cyrillic script reads, "This frequency is part of the International Amateur band — please QSY immediately."

Whether or not you sign the message with your callsign will depend on your local rules. Here in ZL, Amateurs are not permitted to communicate with non-Amateur stations. This legal problem has been overcome by the issue of a special call of ZL6IW. I use this callsign to ask intruders to shift and this works in many cases, although it is sometimes necessary to keep on asking for long periods before they "get the message."

A fact that most Amateurs choose to ignore, whilst still bemoaning the presence of intruders, is that most of the intruders into our exclusive bands are there only because we allow them to be. Given the numbers and geographic spread of Amateur stations around the world, it is within our power to deny an intruder the use of any of our frequencies. Do not move aside for intruders when they come onto your frequency; you are the legitimate user of the Amateur band, not them. Protest the intrusion and continue to protest it in any way you can. Number 115 of the I.T.U. Regulations reads: "Administrations . . .

shall not assign to a station any frequency in derogation of . . . the Table of Frequency Allocations given in this chapter . . . except on the express condition that harmful interference shall not be caused to services carried on by stations operating in accordance with the provisions . . . of these regulations." It follows that, unless an intrusion is protested, the intruders' administration has the right to assume that its stations are not causing harmful interference.

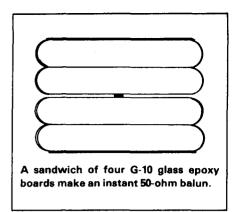
In addition to any protest you may make, your intruder watch should be advised of the date, time, frequency, and nature of the intrusion, along with any relevant information you may have, such as I.D., bearing, and so forth. It is my opinion that Amateurs generally have allowed the present chaotic situation regarding intruders to develop by "leaving it to the other guy" to do something about it while they chased off "up the band" looking for a clear frequency. Unless and until Amateurs as a whole are prepared to act, either through their intruder watch or personally, the situation will not improve.

R.E. Knowles, ZL6IW/ZL1BAD N.Z.A.R.T. Intruder Watch Coordinator Papakura, New Zealand

instant balun Dear HR:

Here is some pleasant news, an addition to the fourth in the series of fine articles by W6GGV on transmission-line circuit design (March, 1981, *ham radio*). Last year I went through Wheeler's 1978 MTT paper (reference 1 in W6GGV's article) of constructing a 4:1 balun for the K2RIW 432 antenna. The results were pleasant indeed.

We have all seen Paul Shuch, N6TX, exploit the fact that a 0.1-inch stripline on the standard G-10 glass epoxy board with groundplane has a characteristic impedance of 50 ohms. What Wheeler's equations show is that a 0.1-inch stripline centered between two parallel plates, constructed of our G-10 boards, is also 50 ohms! (See sketch.) You'll need four onesided boards. Etch a 0.1-inch-stripline on one using the standard tape. Etch a second board free of copper, then



epoxy all four boards together. Instant balun.

> C.R. MacCluer, W8MQW East Lansing, Michigan

ten-second call swaps Dear HR:

As a newcomer to DX, I find myself dismayed when, after exchanging callsigns with a DX station, the other operator rushes off to make another contact. While I realize that he is trying to give other operators in the window a chance to make contact. I feel that these ten-second QSOs should be limited to Field Day and similar contests. It is my opinion that, in order to qualify for a DX award, the operator should acquire some minimum information regarding power, antenna, or even atmospheric conditions. This would add possibly another ten seconds to each QSO, which I don't think would be excessive or prohibitive.

Part 95 states that one of the fundamental purposes for the Amateur Radio Service is to advance radio technology. A call-swap QSO does not help to accurately evaluate a new antenna or an experimental matching network. Armed with the information I suggest to be exchanged, some reasonable judgments can be made and conclusions drawn concerning a stations performance.

I am aware that the system I propose is not perfect, and some modifications would have to be made but I can see nothing constructive or creative coming from the current DX practices, and I think that it is time for a change.

> Bill Marinara, WB1FJE Hamden, Connecticut

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TRAINER FEATURES	MM-1	KT-1	MT-1	CK-1	MK-1	A	B	C	D
Speed Range (WPM)	2-99	1-99	1-99	1-99	2-99	8-50	5-50+	?	8-50
Memory Capacity (Total Characters)	500	or chairing		500		400	100/400	400	
Message Partitioning	Soft			Soft	101000-001	Hard	Hard	Hard	C-10-5-5
Automatic Contest Serial Number	Yes	Mag		Yes	Ves	No	No	No	No
Selectable Dot and Dash Memory	Yes	Yes Yes	Vac	Yes Yes	Yes	No No	No No	No	No No
Independent Dot & Dash (Full) Weighting	Yes Yes	Yes	Yes	Yes	Yes	No	No	Yes	No
Calibrated Speed, 1 WPM Resolution Calibrated Beacon Mode	Yes	ies	les	No	les	No	No	No	NO
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Automatic Stepped Variable Speed	No	No	No	Yes	No	No	No	No	No
2 Presettable Speeds, Instant Recall	No	No	No	Yes	No	No	No	No	No
Automatic Trainer Speed Increase	Yes	Yes	Yes	語言が思いた。	是是是生	211202238	ALL SAMAGED		No
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Random Practice Mode	Yes	Yes	Yes	2543	ACRES CAR				Yes
Standard Letters, Numbers, Punctuation	Yes	Yes Yes	Yes	Mar Nor-Sh				THE REPORT	No
All Morse Characters Advertised Price		\$129.95		\$129.95	\$79.95	\$139.95	\$ 99.50/ \$139.50	\$229.00	The second s
OPTIONS: MT-1P (portable version of MT-1) with batteries, charger, earphone ME-1 2000 character plug-in memore expansion for MM-1 AC-1 600 Ma. 12 Volt wall adaptor for MM-1 with ME-1 AC-2 350 Ma. 12 Volt wall adaptor for all AEA keyer and trainer products except MM-1 w/ ME-1 DC-1 Cigarette lighter cord for all AE keyers and trainers except MT-1P MT-1K Factory conversion of MT-1 to KT-1	\$139.95 y \$59.95 r \$14.95 r \$9.95	single mode keyer intern shown a com from Ask a else I AEA Adva Lynn	e lever of ern amate s are as hal AEA n above. hplete ele the facto a friend he has of keyer an inced E iwood, V	or lambic our transmission easy to compute Each AE evated ten ory. how he l ever tried ad trainer lectronic WA 9803	squeeze nitter with operate rs are al A product nperature likes his I, then J family at c Applica 6. Call 2	e paddle as a for l pre-pro- t is fully e burn-in AEA ke UDGE I your fav ations, I 206/775		key any try requin calcul for the cted and efore it is ared to RSELF. er.	y type of red. AEA ator. The features receives s shipped anything See the
MT-1K Factory conversion	\$ 40.00	Lynn	wood, V	A 9803	6. Call 2	206/775	-7373		La Luga

More Details? CHECK-OFF Page 94

August 1981 🕼 7

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SANTEC•NOLOGY breaks into the 440 band with style! The new ST-7/T synthesizes the entire band in 5 kHz steps, works both up and down repeater splits and does it all right from your hand, with versatile power options of 3 watts, 1 watt or even 50 milliwatts (all nominal), to reach out to where you want. The high power mode of 3 watts radiates on 440 like 5 watts on 2 meters ... and that's a handfull!

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pad is a SANTEC Standard at no extra cost, and the ST-7/T's optional synthesized subtone encoder is controlled by the radio's front panel switch.

All the regular SANTEC accessories used with your HT-1200 fit the ST-7/T as well, meaning that you can enjoy both bands fully with a smaller cash investment. Grab the new SANTEC ST-7/T and join the fun on 440 MHz. See your SANTEC Dealer for delivery details.



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HT-1200

SANTEC'S popular HT-1200 is the incomparable 2 meter leader. This little rig is handing over quality, power and features that you'd expect from something nearer the size of a bread box. SANTEC packs a 2 meter ham shack into the palm of your hand!

SANTEC

. .

4 4 2 7 9 5 M Hz

123A

* 0 # 0

8

5 6 B

9 C

You can carry scan, search, 10 memories and fully synthesized key pad control around with you and still get out with a big 3.5 watts (nominal). Compare them apples to anything you want, and settle for nothing less.

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NAME		CALL		
ADDRESS				
CITY	STATE	ZIP		



ALL 1.8-1.9 MHZ POWER LIMITS WERE LIFTED on June 10, when the Commission also reduced restrictions on the top half of the 160-meter band. The remaining restrictions, listed here by state, are:

States	1900-192	25 & 1975-2000 MHz	1925-1975 MHz
ME, MA, N H, RI	100W	day/25W night	no operation
CT, DE, MD, NJ, NY, PA,	VT 200W	day/50W night	no operation
KY, NC, OH, SC, TN, VA,	WV 500W	day/100W night	no operation
FL, GA, IL, IN, MI, WI	500W	day/100W night	100W day/25W night
AL, AR, IA, MN, MS, MO	1000W	day/200W night	200W day/50W night
All remaining states	1000W	day/200W night, entir	e 1900-2000 kHz
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in its announcement of the relaxation, that new limits could be The FCC Also Warned, in its announcement of the relaxation, that new limits could b imposed at the end of 1982 at which time the WARC agreements go into effect. However, 1.8-1.9 should not be affected.

ARRL Deserves A Big Vote of thanks from all Amateurs for its leadership of the effort to restore 160 meter privileges.

PROHIBITIONS AGAINST 420-430 MHZ USE by U.S. Amateurs near the Canadian border, a change in 220-225 MHz status, and a new 902-928 MHz band are all included in the FCC's just-out General Docket 80-739. In this "Second Notice of Inquiry in the Matter of Implementation of the Final Acts of the World Administrative Radio Conference," the Cor mission reviews the spectrum from 28 through 1215 MHz with no changes proposed for 10, the Com-6, or 2 meters

<u>However</u>, 220-225 Is To Go from its present Radiolocation-Primary, Amateur Radio-Secon-dary status to one with Amateur, Fixed, and Mobile all sharing primary status. In ac-cordance with the WARC Region 2 allocation, this would leave the door open for later FCC allocation of part of the band to one of the other services.

Operation Between 420 And 430 MHz would be forbidden to Amateurs located in a strip of the northern U.S. stretching from the Atlantic to the Pacific, to prevent interference with Canadian commercial users of that segment formally assigned worldwide for the Fixed and Mobile services at WARC. Among the Amateurs affected are those in or near such major cities as Seattle, Detroit, Toledo, Cleveland, Buffalo, and Bangor. A precise descrip-tion of the area affected appears in Part 1.955 of the Commission's rules. <u>A New Band At 902-928 MHz</u> is the good news, with Amateur Radio a secondary allocation to Radiolocation. The only part of the country that won't have use of this new band is

in Colorado, in an area that unfortunately includes Denver. <u>Reply Comments</u> on this Notice of Inquiry are due at the FCC by July 30.

ROBERT W. STANKUS (KESWICK SALES) WAS INDICTED on 22 counts of using the U.S. mails to defraud by a Federal grand jury in Roanoke, Virginia, June 10. He'd advertised new Kenwood TS-520SEs in the Ham Trader for \$369.95 last fall, and though he'd even delivered a couple (purchased from legitimate dealers at near list prices), more than 50 Amateurs had eventually filed complaints with the Postal Service over his failure to deliver rigs as promised.

Eleven Of The Complaining Amateurs were selected for the indictments, with two indictments returned for each complainant. Stankus, NIAAR, was to have been arraigned July 1 and will probably go to trial this fall. If convicted, he faces a maximum fine of \$1000 and five years in prison on each of the 22 counts.

PROPOSED PLAIN-LANGUAGE AMATEUR RULES, PR Docket 80-729, has received a last-minute two-month extension on the due date for Comments. Acting on a request from the ARRL, the Commissioners agreed June 15 on an August 22 due date for Comments, with Reply Comments due on or before October 21.

OSCAR 7 IS PROBABLY DEAD FOR GOOD, following a sudden failure June 12. Thermal stress induced by the solar eclipse it began experiencing on each orbit after May 22 may have triggered the failure. The current theory is that the battery pack (which had been open since a cell failed in 1978) has shorted and pulled the power bus down. KA9Q, whose calculations predicted the solar eclipse, estimates the spacecraft will return to full sun-light about July 12. There is a possibility, considered remote, that it could recover then. Users are asked to monitor OSCAR 7's beacon frequencies, 29.502 and 145.972, and to report any signs of life to AMSAT.

to report any signs of life to AMSAT. OSCAR 7 Has Been An Exemplary Satellite. Launched November 15, 1974, with a design life of three years, it has provided Amateurs throughout the world with literally mil-lions of contacts during more than six and a half years of operation. OSCAR 6, with a one-year design life, lasted over four and a half years. <u>The Ariane Launch June 19</u> from French Guiana was a complete success, reinforcing ex-pectations that the Phase 3B spacecraft will go up next year as scheduled. Two satel-lites, each weighing about 1500 pounds, were launched into orbit by the French rocket. One, the experimental Indian "Apple" satellite, is reported to be having problems ex-tending its solar panels. W6VIO at the Jet Propulsion Laboratory joined WA3NAN and WA2LQQ in reporting the 1203Z launch to listening Amateurs worldwide.

John L. Remartz Father of Shortwave Radio

By Leonard Spencer, WA6CBQ

As a Novice-class Amateur attending one of my first radio-club meetings, I was recruited to help the guest speaker with his demonstration of a "tuna-fishcan rf bridge." It was the Monterey Bay Radio Club in Salinas, California, during September of 1959, and that speaker was John L. Reinartz. Little did I realize at the time the scope of his electronic accomplishments.

We met again when I served as president of the same club and invited him to furnish a program. He was accompanied by Mrs. Reinartz, who was a true helpmate and at that time also an Amateur with the call K6MJH. She was an observer and participant in much of the pioneer experimentation which brought about shortwave communications. It is with her cooperation that facts pertaining to these events have been obtained for this article.

Between the years 1921 and 1925, John L. Reinartz led the Amateur Radio experimenters of that time in development of shortwave communications below 200 meters wavelength. Recognized engineers and scientists of that time considered these wavelengths useless.

Reinartz engineered the receiving and transmitting circuits, developed tune-up procedures, and did the research necessary to explain how the shortwaves performed as they did. For this Amateur Radio accomplishment, he truly earned recognition as the *father of shortwave radio*. He upset many of the accepted theories of fundamental radio communications and probably for this reason did not get the recognition he deserved at that time.

greatest contribution

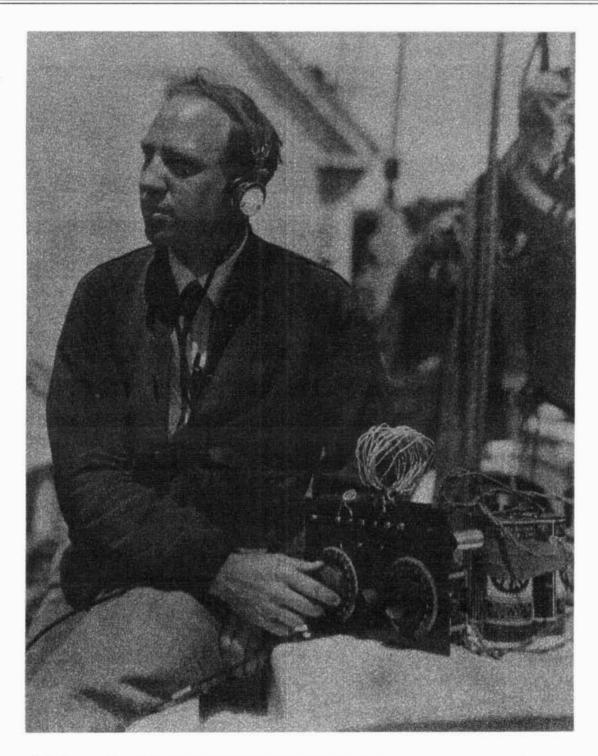
In 1964, Herbert Hoover, Jr., W6ZH, then president of the American Radio Relay League, awarded Mr. Reinartz the first Hiram Percy Maxim Gold Medal for the greatest contribution to Amateur Radio in the past 50 years. The ceremony took place on the 50th anniversary of QST, at the Pacific Division Convention in Sacramento, California. Actual presentation was made later at the Fort Ord Hospital, as Reinartz was unable to attend the convention.

Mr. Hoover summed up his address with this tribute to Reinartz, then retired and living in Aptos, California, and using the call K6BJ:

"In the immediate scramble for shortwaves that followed his basic concepts, Reinartz' pioneering work became strangely overlooked. Perhaps the fact that he was an obscure electrician in a New England textile mill, who had overturned the accepted theories of the scientific authorities of the day, had something to do with it.

"But John Reinartz should not be forgotten. Quiet, modest, and unassuming as he may be, he perhaps more than any other individual — is the father of shortwave radio. When we realize that today, 40 (now 57) years later, the great bulk of the world's long distance radio communications broadcast, point-to-point, marine, aviation and all others — still takes place on these same shortwaves that were first demonstrated by Reinartz, we can justly be proud of his Amateur accomplishment."

When Mr. Hoover's activities at that time are considered, these statements give more perspective to



Reinartz was radio operator on board Bowdoin for MacMillan's 1925 Arctic Expedition using WNP (Wireless North Pole). Many firsts occurred that summer when shortwave radio provided daily communications from the far north in the daytime. Commander MacMillan wrote in a National Geographic story, "Much of the success of the Expedition's radio work depended upon the eager cooperation of Reinartz' associates of the American Radio Relay League in transmitting messages. More than 30,000 words of news dispatches alone were sent out from Bowdoin addressed to the National Geographic Society and released by it, day and night, to the press associations." All without financial remuneration – remember the regulations!

events in early-day radio. With the call 3ZH he had a kilowatt tube-type transmitter in Washington, D.C., and during the 1922 transatlantic tests his station was heard in Switzerland and elsewhere in Europe. He was trying to communicate across the Atlantic at much the same time as Reinartz. It was the shorter wavelength 100-meter equipment used by Reinartz that made the accomplishment possible, using a 100-watt transmitter with 400 watts input. As Reinartz said in a magazine article at the time, "I didn't give a darn whether I burnt up the tubes or not."

honored as a professional engineer

More professional recognition was given Mr. Reinartz in January, 1958, when he was named a Fellow of the Institute of Radio Engineers for his early work on radio-wave propagation. He had 28 patents in the electronics field, but his early circuits were not controlled, as he wished everyone to make use of them, even to the extent of letting commercial enterprises manufacture them.

Reinartz received a bronze-plaque life membership in the Institute of Electrical and Electronics Engineers, Inc. A medal was awarded him by the Manchester, Connecticut, Chamber of Commerce on May 14, 1925. Manchester was his home town. The 1960 session of the California Legislature passed a resolution, dated March 8, commending Mr. Reinartz for his contributions to microwave radar. A beautiful plaque from the Central California Radio Council was presented to Reinartz, K6BJ, on April 6, 1960, in recognition of his many years of service to Amateur Radio. Among his other honors he was a member of the Explorers Club of New York, the American Polar Society, an original member of the American Radio Relay League, and an associate member of the Naval Institute.

spark station in 1908

John's youthful interest in radio started in 1908 when he bought an issue of *The Electrical Experimenter* at a magazine counter near school. He bought the secondary of a one-inch spark coil he saw advertised there by saving the 10 cents a day he earned working for a blacksmith. The transformer was completed by using iron wire for the core and bell wire for the primary winding. He made the electrolytic interrupter for the spark coil, and fashioned a coherer from a quarter-inch glass tube filled with filings from nickel coins. Using his initials, JL, as his call (as was done in the days before controls), he connected his spark transmitter to a 600-foot antenna installed between some trees and went on the air.

By 1921, Reinartz, using the call 1QP and working as superintendent of the local power company, published a magazine, *How to Build Receivers and Transmitters at Low Cost*. It was distributed free, and it was at this time that his famous tuner was developed. These circuits were reproduced and information on construction reprinted in the radio magazines of nearly every country in the world. Because of this wide publicity, thousands of his tuners were built.

QST in October, 1922, carried a symposium of further improvements on the Reinartz tuner. An editor's note states: "It is impossible for us to keep up with this man Reinartz. Since preparing the foregoing for publication he has dropped around with another 'trigger circuit' that knocks its predecessors cold."

first across Atlantic

on 100 meters

Reinartz participated in the planning of, and designed the circuits used for transmitting, the first transatlantic two-way radio communications. Monsieur Leon Deloy, f8AB, of Nice, France, had been in the United States to attend the 1923 ARRL National Convention in Chicago. At that time, Deloy visited the Reinartz home in South Manchester, Connecticut, where he was given the circuits and construction details of the famous transmitters at a session that lasted until 3 AM. Together, they arranged for the transatlantic tests on 100 meters. Reinartz had developed a single tuner for receiving which was variable from 200 meters down to 28 meters, along with the technique of setting the transmitting and receiving equipment to precisely the desired frequencies so each station was compatible with the others.

This great event in shortwave radio took place the night before Thanksgiving Eve at exactly 10:30 EST, the time at which the restrictions on Amateurs, to prevent interference with broadcast listening, were lifted.

F.H. Schnell, a1MO, traffic manager for the ARRL in Hartford, Connecticut, had secured special permission from the Supervisor of Radio in Boston to use the 100-meter wavelength for these experiments.

Deloy, f8AB, on the night of November 27, 1923, called America and sent two messages prior to the 10:30 schedule. Both of these messages were copied by Reinartz at his home station in South Manchester.

At 10:30 long calls to France by a1XAM, the special experimental call assigned to John Reinartz, and a1MO, the call of the station owned jointly by Schnell and Kenneth B. Warner, secretary of the

Remartz

ARRL, were heard "QSA vy one foot from phones," across the Atlantic. Steady and reliable communications were carried on for two hours and five minutes that night, which was until 6:35 AM in France.

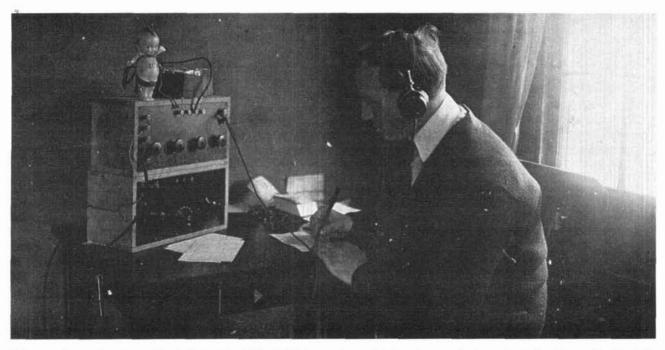
Message number one from America was addressed to General Ferrie, Director of Telegraphs for the French government, which was acknowledged at 11:06, and read as follows:

"America greets you for the first time by Amateur Radio across the ocean on 100 meters." Signed ARRL.

long-wave communications

Another sidelight to the problems of this era was that very little commercial equipment was available for use below 200 meters. Reinartz' circuits accomplished this, making the short waves practical for experimentation.

Early transmitting circuits were not shielded. They were of open breadboard construction, so rf potentials were prevalent pretty much everywhere. Reinartz seems to have been the first to establish a nodal



This operating position shows John L. Reinartz in 1921 with his famous tuner of that era in use, with three stages of audio amplification in the top cabinet. The spark transmitter, located in the cellar, was remotely controlled from this upstairs room. The Kewpie doll on top of the receiver was sent to him by a Canadian Amateur as a mascot — he had the famous 1QP call at that time and signed some of his message traffic "Kewpie."

Two nights later, Reinartz' station, 1XAM, succeeded in connecting with Deloy's f8AB in France twice, and at 10:40 asked Deloy to change his wavelength as he was being interfered with by station KDKA at about 103 meters. Deloy was not heard after that.

A complication arose when international DX became possible, that of identifying call letters for the various countries. Assignment of calls specified a district by number and a suffix of alphabetical characters, but no prefix to the call. The early DX workers added prefix "u" for United States, "a" for America, or "y" for Yankee, to establish country of origin as the United States of America. Great Britain used "g" and France "f" to do the same. point, or point of zero rf potential, at the filament tap on the oscillating circuit. He did this by connecting an rf choke from filament to ground and tuning the circuit until no current flowed through the choke. This later became part of the standard technique.

A really practical shortwave transmitting circuit designed by Reinartz was published in an Amateur Radio magazine, *The Modulator*, in May of 1923. This was the official organ of the Executive Radio Council, Second District, published in New York City.

The story was tied in with another article concerning the failure of two-way transatlantic work between the American and English Amateurs. Because of the great amount of interference, anything in the

Reinartz -



Here is a panel view of 1922 version of the Reinartz shortwave tuner, built by W. W. "Woody" Wilson, WA6KVW, from an early magazine article. He used authentic parts of that period from his extensive antique wireless collection. Switch points picked up progressive taps on the coils — an early version of band switching, you might say.

way of continuous reception had been prevented when stations were operating between 180 and 350 meters. Most of the transmitting and receiving equipment was designed to function somewhere in that part of the spectrum.

shortwave experimentation

Following are quoted paragraphs from *The Modu*lator:

"... real shortwave transmission did not exist until the recent ARRL tests between stations 3ALN, 1HX and 1AW, in cooperation with 1QP were consummated.

"The results of these tests were of great interest to the Amateur world. They proved that *transmission* on 100 meters was a practical proposition. It was carried on for several weeks before the officials would commit themselves.

"And now, with the announcement of the Hoover Conference recommendations, we find that we may have to transmit on those shorter waves! The study, therefore, of a successful transmitter will be of immediate interest.

"One of the best shortwave transmitters used in the preliminary tests was that of station 1QP, John L. Reinartz, of Tuner fame, located in South Manchester, Connecticut. Reinartz developed for these tests a type of transmitting circuit which is not less remarkable than his receiving circuit. With it, he reached the low limit of 100 meters, with mighty good radiation.

"Reinartz tells us that the success of his transmitter depends on the adjusting of the counterpoise and antenna current so that there is equal amount of current flowing through both. To obtain this, both the antenna and counterpoise are tuned separately with a radiation meter, and the adjustment of the helix (coil) turns in both circuits. At this adjustment, the plate current is lowest. To further regulate the circuit, the grid condenser is adjusted so that the plate current falls no further.

"Now that the Amateur has proved his efficiency on what were considered 'useless' waves, he will have the opportunity to show what he can do on the lower ranges and limits. Here is the field — some of the pioneer work has been done — so let's get at it. The field is large in its opportunities, and if the Amateur's skill is equal to it, he will contribute many things to the science and art of communication on the shorter waves."

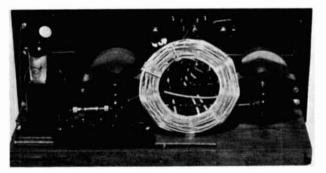
Reinartz' station on 100 meters, 1XAM, determined a new principle in antenna operation. For the first time, he used a counterpoise as the other half of a balanced antenna system instead of a capacity ground connection as had been common practice.

explains shortwave propagation

During 1923 and 1924, Reinartz compiled statistics on the stations he worked and those which reported hearing his experimental shortwave transmissions, some 5,000 in number. He hoped to determine why the shortwaves performed as they did. Experiments were carried on during an eclipse of sun to prove that sunshine made shortwave communications possible.

From his experiments, he came up with a "skip distance" theory, based on reflection or refraction from the ionized Kennelly-Heaviside layer, to explain the shortwave phenomena, and he predicted that communications would be possible across the nation in the daytime hours, which was an entirely new concept at that time.

Reinartz had schedules with stations NKF (call of Naval Research Laboratory at Bellevue, D.C.), 8XC, and 2EB on wavelengths below 40 meters in 1924. Tests were made at night, with three miles or so being the only completed QSOs.



Inside view of the Reinartz tuner with spiderweb inductance, which was an improvement over the original version in June, 1921 QST. The tube is a UV-199. Gridleak resistor was in clip holder for easy changing. Note hole in panel to check on tube filament glow, controlled by varying rheostat on front panel.

Then daylight tests were made on December 21. At 3:25 PM CST, 9EK, Hoffman, called 1XAM, Reinartz. They had been hearing each other for some time and suspected they could work together easily. This they did on 21 meters until 4:30 PM when signals suddenly dropped out. As usual with this band, darkness had changed the transmission pattern. The most important part of this QSO was that 6AJF, Frank C. Jones, at Berkeley, California, copied both stations — 9EK in Madison, Wisconsin, and 1XAM in South Manchester, Connecticut!

daytime transcontinental signals

Two-way communications on 21 meters were conducted between Reinartz, 1XAM, from South Manchester, Connecticut, and Ed N. Willis, 6TS, Santa Monica, California, at noon on January 22, 1925. They maintained continuous communications for 25 minutes and had to stop so Reinartz could get back to work as electrician at the silk mills.

The April, 1925, issue of *QST* contained "The Reflection of Short Waves," by John L. Reinartz, 1XAM, in which the editor said: "We consider this article one of the most important contributions made to radio literature." This evaluation proved to be true.

The article explains with drawings and text how the shortwaves propagated, providing the theories for the Maximum Usable Frequency and Lowest Usable Frequency as we know them today. These theories contradicted the then-accepted action of reflection by the Kennelly-Heaviside layer surrounding the earth. Reinartz' theory of "skip" is now scientific fact.

There was a controversy as to who was first to work across the Atlantic Ocean on 20 meters during daylight, but credit is given to Reinartz, u1XAM, and Secretan, g5LF, in England, for being the first to accomplish the feat.

On May 3, 1925, communication was made on 20 meters between British Amateur station g2OD and Australian a2CM, following planned experiments, so the use of shortwaves for DX rapidly became popular once the Amateurs could make the equipment work on these wavelengths. (Notice that frequency is not mentioned in the early days of radio. It was always wavelength in meters — metric units from the start!)

In 1923, Amateur Don Mix, 1TS, was assigned as radio operator with the MacMillan Expedition to the Arctic. His equipment was built to work in the 180 to 220 meter range, and this proved to be unreliable for consistent communications from the north because of the long period of daylight during the summertime. This was the eighth polar exploration of which MacMillan had been a part — and the first to use Amateur Radio to maintain contact with civilization.

Arctic expedition of 1925

John Reinartz' participation in the Navy-MacMillan Arctic Expedition of 1925 began when he received a letter from Hiram Percy Maxim, president of the ARRL, inviting him to cooperate with Commander Donald B. MacMillan, USNR, and E.F. McDonald, Jr., president of Zenith Radio, which was to supply the radio equipment. Reinartz helped to design and construct the shortwave station installed on *Bowdoin*, a schooner with an auxiliary oil-burning engine, commanded by MacMillan.

Captain McDonald, USNR, was in command of *Peary*, a coal-burning steamship, on the expedition. Lieutenant Commander Richard E. Byrd, Jr., on active duty in U.S. Navy, was in command of the Naval Aviation Detachment with the expedition in his first attempt to fly over the North Pole.

This expedition was organized under auspices of the National Geographic Society and the U.S. Navy. Here is a quotation from the *National Geographic* magazine of that time:

"Airplanes will explore in days icy areas of the Arctic which would take months to traverse by dog sleds. Radio is telling the daily program of work which in years past would be shrouded in silence for months. For the first time in Arctic history, color photographers are recording, for members in early issues of their *Geographic*, the surprising tints of the far north, the native life, birds and many beautiful Arctic flowers.

"These new aids to travel and communication enable the expedition to engage upon a program perhaps of broader exploration and scientific study than any expedition heretofore has attempted.

"The flying of the U.S. Navy airplanes under the direction of Lieutenant Commander Richard E. Byrd, Jr., and his splendid personnel, not only is epochmaking, but marks an important experiment in aviation that will focus the world's attention."

First date of the personal radio log kept by Reinartz on the Arctic expedition is June 16, 1925, while the crew was preparing to depart from Wiscasset, Maine. Amateur calls worked on 40 meters are entered in the record for several days and the *National Geographic* message of the day was transmitted using Amateur Radio for relay. Some of the interesting entries follow:

"June 26 – Left Sydney (Nova Scotia) at 9:15 AM. During radio watches, worked NKF and 1QP (his home station in Connecticut), 1MY (a personal friend), and several other Amateurs in 40-meter band.

"June 28 — At sea bound for Battle Harbor. Gave 1MY a message by voice at 7:20 PM in the 40-meter band.

"July 1 — At Battle Harbor. Met Stanley W. Brazil who operates the northernmost Marconi radio station on Labrador coast. Made him a transmitter and receiver operated entirely from dry batteries and expected that he would be able to keep in touch with me on the trip north, which he did!

"July 2 — Don't forget to learn the code better please — I want to hear you on the air when we are up north." (This was in a letter from John to Mrs. Reinartz mailed from the last post office on the way north, Battle Harbor, Labrador.)

"July 5 – Left Battle Harbor at 12:30 AM aboard *Peary* for Hopedale (Labrador). Sailed on *Peary* in order to put their defunct radio equipment in commission. Was necessary to tear out all the radio equipment and rebuild more compactly.

"July 6 — Left *Peary* after fixing radio equipment and reboarded *Bowdoin* when she stopped on account of fog.

"July 23 — From WNP to 1MY 6 PM via f8QQ. Have Gertie (Mrs. Reinartz — Gertrude) keep 1XAM going on 20-meter band with bug 8 AM to 8 PM Saturday and Sunday. Sig. Kewpie." (This was for experiments in determining time when 20 meters would be open to the United States from the northern end of Greenland. Reinartz had an automatic code sending device at his home station, which he had used earlier in his experiments on shortwaves. The keyer was made by John from a clock mechanism and the spring had to be wound up every half hour. Code signals were made by timing from a revolving disk attached to the clock.)

"August 2 — Etah at last. Had a great time today. Heard many 20-meter Amateur stations and worked 9CXX, Cedar Rapids, Iowa, on 16 meters. Gave him message number 218 to Mrs. QP."

Quote

Am well and happy. Love via 20 meters from Etah the first time in history of radio at 3:40 PM. Kewpie

Unquote

This message was telegraphed to Mrs. John L. Reinartz that same day by Arthur Collins, and several other Amateur stations who received it, and was the first message ever to be sent on short wavelengths from the frozen north in the daytime.

Etah, Greenland, was the ship base from which the

planes explored the uncharted regions of the North Pole. WNP (Wireless North Pole) operated by Reinartz, was aboard *Bowdoin* at Etah. (9CXX was the station of Arthur A. Collins, then a student in high school. He cut classes to get back to his rig to handle these communications. He was to become famous for the manufacture of quality electronic equipment, and Collins Radio is now part of Rockwell International conglomerate.)

"August 11 — The three planes off to Ellesmere Island at 10:47 AM. The NA-3 back at 2:15 PM, the NA-2 back at 2:30 PM, and the NA-1 back at 2:44 PM. They could not land due to fog. Was in radio communication with Washington, D.C., through Cedar Rapids, Iowa, and gave notification the instant the planes returned. Newspapers had news within 15 minutes of their return.

"August 12 - It is evident that planes are not going to be useful up here. The Los Angeles (dirigible) would be much better. A drop means certain death. Looks as if the flyers knew it too well.

"August 15 — Made world's record by using radio telephone to Cedar Rapids at 12:30 PM.

"August 17 — Not much doing. Planes having hard time finding landing place on Ellesmere Island. Again talked with Cedar Rapids at 2 PM. It was noon at Cedar Rapids.

"August 18 — Same old story. Planes not suited for job." (These single engine Loening biplanes were specifically equipped with dry-battery-operated receivers and transmitters capable of operating between 37 and 42 meters wavelength, plus standard Navy aircraft equipment operating on 500 meters.)

"August 20 - Bad weather. Packing up to go home.

"August 21 - Left Etah at 6:29 AM for home.

"September 14-20 — At Godthaab (Greenland). While there visited the Norse ruins, going to the ruins on September 15 at 7 AM and returning on the 16th. The ruins are about 15 miles from Godthaab, at the end of the fjord.

"October 10 — stayed at Monhegan due to storm until Monday morning when *Bowdoin* left for Wiscasset. The end of the journey."

The next year, after the expedition of which Reinartz was a part, Lieutenant Commander Richard E. Byrd and his pilot, Floyd Bennett, flew over the North Pole in a commercial type Fokker three-motor plane, *Josephine Ford*. Bennett had been a Navy Pilot with the 1925 MacMillan Expedition. The flight left from Kings Bay, Spitsbergen, May 9, 1926, making a 15-hour, 1500-mile trip over the pole, returning to Spitsbergen.

Three days later, May 11, 1926, the Amundsen-

REINARTZ -

Ellsworth-Nobile flight, leaving from the same place, using a dirigible 347 feet long, dropped Norwegian, American, and Italian flags over the North Pole and continued on to Teller, Alaska. Each expedition added more knowledge about the Arctic region, making it easier for other explorers to follow them.

After the Arctic expedition, Reinartz did further experimentation for the Navy, and worked on measurement of voltages produced by growing plants at what is now the University of Connecticut. He was commissioned a lieutenant in the Naval Reserve in 1929 and conducted weekly classes via radio for Third Naval District personnel.

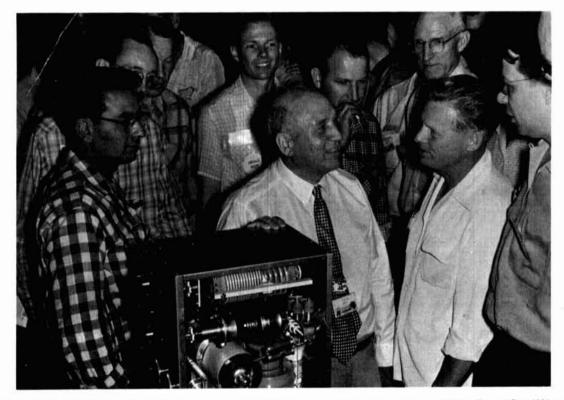
fying fm radio.

Professional engineering status was earned in the State of Connecticut, July, 1936, when he was issued Certificate of Registration number 947.

active Navy duty

In 1938, Reinartz was called to active duty in the Navy as a personnel officer, assigned to assemble eligible, experienced radio personnel for training and research. By the time of Pearl Harbor, he had assembled a list of 720 reserve officers and 3500 enlisted reserves who were quickly assigned to communications duties.

His initial job done, Reinartz was assigned to other



At the 1955 ARRL Pacific Division Convention in Fresno, California, the tecnnical discussion topic was "Your Future One-kW SSB Final." Here John L. Reinartz shows the finished transmitter to very interested Amateurs at the conclusion of the lecture. About his work he said, "I'm not selling anything, but basically these hams are the men who eventually move into places of prominence as operators of the big commercial stations, engineers in the telephone industry, ship operators, executives in the communications field, and military signal transmission experts. They have the know-how for the important jobs."

Reinartz' formal education ended at the eighth grade level. He took correspondence courses and attended evening school classes to keep learning. In 1930, he decided on a college education. He attended Connecticut University for two years, majoring in radio and mathematics, but received a job offer from Radio Corporation of America that was too good to turn down, so went to work for RCA in 1933. Twentythree Reinartz patents were assigned to RCA in this five-year period, some of them on methods of simpliWashington duties, including a tour as head of the Radio and Radar Division of the Naval Research Laboratories, at which time he supervised the work of 1200 physicists.

As a member of the Joint Chiefs of Staff communications setup, he had a lot to do with the vital nerve system which meant so much to Allied victory. He was assigned to the office of the Chief of Naval Operations.

Reinartz was a member of the Defense Communi-

John L. Reinartz Father of Shortwave Radio



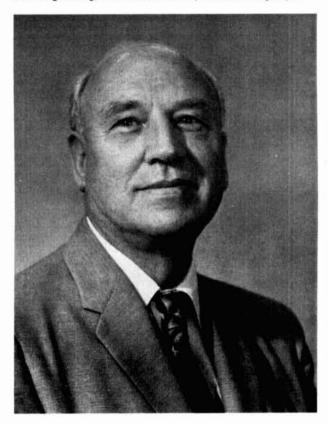
Here Mrs. Reinartz points to an entry in the personal log kept by John L. Reinartz while he was radio operator of station WNP on board Bowdoin during 1925 Arctic Expedition to North Pole regions. Captain Reinartz is in photograph beside replica of 1922 version of his famous tuner. Scrapbook behind photo is of Arctic trip and larger book contains clippings and programs of events in which John participated while with the EIMAC firm.

cations Board at its formative meeting January 6, 1941. For a time, he was on the West Coast in charge of modification of all airborne radar for the Pacific areas.

In 1946, Reinartz rejoined RCA, and in 1949 moved west to become part of the Eitel-McCullough, Inc., organization at San Bruno, California, in a similar capacity, as field engineer and manager of the Amateur Radio Service Department, until retirement January 30, 1960. (The Eimac transmitting-tube manufacturing firm was founded in 1934 by two Amateurs, W.W. "Bill" Eitel, W6UF, and Jack A. McCullough, W6CHE.)

honored in retirement

His retirement prompted a John Reinartz Testimonial Amateur Radio Banquet February 1, 1960, in San Mateo, California. Seventeen speakers praised John for his pioneer accomplishments in the field of electronics. Two hundred invited guests paid him tribute by their attendance at the banquet, almost all of them with Amateur Radio calls. This portrait was taken while Reinartz worked for EIMAC. Shortly after World War II, many photos showed him in his Captain's uniform in advance publicity of his many tours thoughout the United States, giving lectures and demonstrations before professional engineering, Amateur Radio, and public services groups.



Check your QSL cards to see if you have worked John L. Reinartz on the air. Calls he held and used were JL, 1QP, 1XT (a special land station call issued by Bureau of Navigation, U.S. Department of Commerce), 1XAM (special shortwave experimentation call), NDF, WNP, N1QP, W3RB, W3IBS, and K6BJ. The Santa Cruz County Amateur Radio Club now uses the K6BJ call for its John L. Reinartz Memorial Station in the Santa Cruz, California, area, so listen for it Field Day, and work it.

This quotation comes from an interview with John Reinartz published in the Augusta, Georgia, *Chronicle* of Thursday, May 3, 1956:

"'It's a wonderful hobby — radio,' he says. 'I can't recommend it too highly. A normal lad can build himself a receiver for as little as \$10 — and a transmitter for only \$25. If he uses the right frequencies he can cover a good part of the world with such a rig. And experience in radio can stand him in mighty good stead in this Atomic Age.'"

ham radio

an analysis of ALC **circuits**

Automatic load control circuits are investigated as a means for avoiding distortion in SSB transmitters

This article dispels the mystery of the ALC circuit used in Amateur SSB transmitters and addresses problems encountered when using linear amplifiers made by one manufacturer with exciters made by another manufacturer. Such problems include splatter, which may occur even though the ALC line is connected between the two pieces of equipment.

ALC defined

ALC is an acronym for *automatic load control* when used with SSB transmitters. The term goes back to the days when Bell Labs made a high-frequency SSB transmitter whose output was in four channels. The term *loading* in telephone work indicated the number of channels in a multichannel system that were in use at any given time. Their auto-

My thanks to Professor Charles Gould, W4LZO, for reviewing this manuscript – John P. Weber.

matic load control kept the peak envelope power from the SSB transmitter about the same, regardless of whether only one or all four channels were in use. Probably there would be less confusion as to the function of automatic load control if it had meant automatic level control.* Even automatic gain control might be appropriate, except for the fact that it would be confused with the system used in receivers.

So in terms of today's technology, what is ALC? What does it do? Why use it? These questions are explored in this article, which includes a discussion of the principles of ALC and why, if used properly, it helps to eliminate interference on the Amateur bands by reducing spurious products in SSB transmitters.

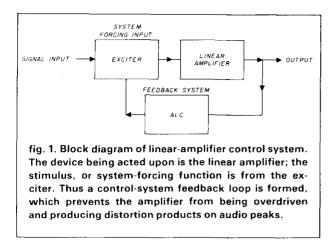
some background information

In a-m communications systems it was common to use speech compressors and clippers, so that the power amplifier could operate near its maximum power-output capability without excessive distortion products. As used in single-sideband, ALC is part of a system that is intended to maintain the distortion products of a *linear* amplifier within acceptable limits.

ALC may be thought of as an rf compressor. You might ask, Why not use audio compression to hold

*To add to the confusion, automatic level control has been used in Broadcast engineering to define a method for taming speech amplifiers. Editor

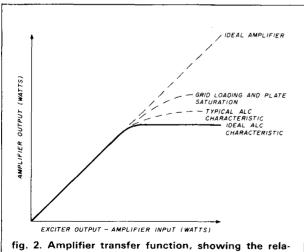
By John P. Weber, Jr., K4JW, 102 Southgate Boulevard, Melbourne, Florida 32901

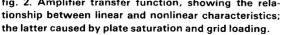


the transmitter output to an acceptable maximum? Well, the peak levels of the SSB signal don't always correspond to those of the audio signal. So the method referred to as ALC in SSB controls the drive power from the exciter (input to the linear amplifier) by sampling the rf output of the linear amplifier, rectifying this output signal to obtain an audio signal, and feeding the filtered dc back to a stage in the exciter that controls the exciter's gain. It is similar in operation to the AGC circuit in a receiver.

The point is that we are trying to maintain a *linear* input-output relationship for the amplifier over the range we're using. Of course, if the linear amplifier could maintain its linearity over the desired operating range by its design, ALC would not be needed. But in today's world, it's possible to overdrive the linear amplifier, which creates splatter.

Now if you don't care about your signal quality or what you're doing to the hams 7 kHz away, stop here! When contacting manufacturers for technical





information, an engineer at one company replied, "It is our considered opinion that most DXers do not use the ALC feature inasmuch as they wish to obtain maximum power output, splatter or no splatter." While I don't think it applies only to DXers, this gentleman's observation is borne out by listening on the Amateur phone bands. I'm interested in the ham who has his ALC line hooked up and is *still* splattering.

analysis

For a thorough discussion of linear distortion mechanisms and measurements, I refer you to the article by Warren B. Bruene;¹ this is must reading. References 2, 3, and 4 are excellent material, not only on ALC circuits, but on all facets of SSB. They may not, however, be easy to obtain.

Fig. 1 is a block diagram of a basic ALC system, from the standpoint of control-system design. The device being acted upon is the *linear amplifier*; the stimulus, or system-forcing input, is the exciter. The linear amplifier has a transfer characteristic; that is, its output related to its input stimulus. If we sample the amplifier output, process it, feed it back to control the stimulus (or exciter), we have a closed-loop control system. Why this roundabout approach? I've found that some hams get confused if I use the term *feedback*, relating it only to an audio amplifier; but they seem to understand the control system concept.

What we are trying to accomplish is shown in the curves of **fig. 2**. We see the ideal linear transfer characteristic that is linear over the entire range, regardless of the input. Then we see a nonlinear condition occur as we go over a certain value of input. In a practical grounded-grid SSB vacuum-tube linear amplifier, this is usually the result of two effects:

1. *Plate voltage saturation* — the output of the tubes tries to exceed twice the value of the supply voltage.

2. Grid loading — as the grids begin to draw current, they present a varying load to the exciter, and because of the exciter's regulation, the exciter will drop its output, generate harmonics, and the power output from the linear will decrease.

Ideally, we want to cut the output power at a point where further exciter drive will not result in additional amplifier output that is distorted. Our compressor would have a curve that suddenly flattens off at the threshold point; but practically, we will get some further increase in output, as shown.

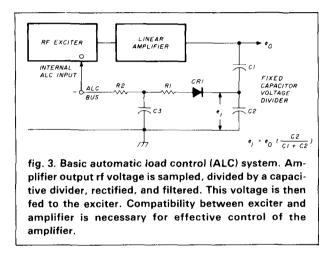
basic circuit

A basic ALC circuit is shown in **fig. 3**. The output rf voltage of the amplifier is sampled, divided by a capacitive divider consisting of C1 and C2, rectified by diode CR1, and fed into a filter consisting of R1, C3,

and R2. This ALC voltage is fed into the exciter.

Most modern exciters have an *internal* ALC system that works from the exciter's output stage back into earlier stages to control its linearity. We can take the ALC from the power linear amplifier and feed it into the internal loop of the exciter to reduce its output to prevent overdrive of the linear amplifier.

In the case of this simple form of ALC system, the threshold point at which ALC action starts is determined by the exciter's ALC circuitry. Many times, this takes the form of a zener or biased diode arrangement. Since the exciter controls the point where the threshold is, how do we know we have the right voltage from the linear amplifier to initiate this action? If both designs were by the same person, he could ensure that these points coincide. But, if the linear amplifier comes from one company and the exciter from



another, how can we assume that the ALC action is optimum, or even exists?

examples of ALC circuits

Now let's look at some circuits. The sampling was from 24 models made by nine manufacturers. Examining fig. 4 we see the first step above the basic circuit. In fig. 4-A, again we have the capacitive voltage divider, this time across the cathode or input circuit of the grounded-grid linear amplifier. Why sample the cathode rf voltage? Well, the power gain of the tube or tubes is very nearly equal to the ratio of rf plate voltage to the rf cathode voltage, since the tubes' plate current is common to both the input and output circuit. Actually, it's a little different because of the power absorbed by the grid(s) and input- and output-circuit losses; but it is a convenient low-voltage point in which to work. This divided voltage is rectified by CR1, filtered by R1, C1, and R2, then fed to the high side of a pot, R3. This pot enables adjustment of the output voltage (ALC) to match the requirements of the particular exciter with which it's

being used. However, the threshold is still controlled by the exciter's internal circuitry.

the most common ALC circuit

Fig. 4B shows the most common ALC circuit in use, according to my study of manufacturers' diagrams and also from knowing which are the most popular units by on-the-air listening.

The components are all established by the designer and probably are intended for use with exciters manufactured by his company — remember, he has extensive knowledge of the exciter.

Again, a capacitive divider is in the cathode circuit; a diode is used to rectify this rf voltage. However the diode is biased so that it will not conduct — this bias is usually derived from a resistor tap in the high-voltage supply bleeder circuit; or, in some cases, from a separate low-voltage power supply that is used to energize relays and pilot lights.

Since the diode is biased, no ALC voltage appears from the linear amplifier until the amplifier reaches a power level where this bias is overcome and the diode conducts.

In effect, the designer has set the threshold point in the linear amplifier. But if that point is below the required threshold voltage needed by your exciter's ALC bus, no ALC action will occur until you overdrive the linear sufficiently to start ALC action. The older exciters needed -2.0 to -4.0 volts for threshold; some of the newer ones need -6.0 to -10.0 volts. For ultra-conservatives, who are safety conscious, examine this circuit in terms of the bleeder resistor tap-to-ground opening up!

biased diode, variable threshold circuit

Fig. 4C is a circuit that allows you to overcome the restrictions mentioned above. Simply stated, the threshold point is made variable by using a pot to set the diode bias. If you note the curves shown beside each circuit, it becomes apparent that the most flexibility to work with different exciters is afforded by this circuit.

This convenience is negated, however, if the ALC control is put inside the linear or on its back panel. The control should be placed on the front panel of the linear — it is an operating control. Why? Well, first (and please don't laugh), so you can set the linear to 1 kW legal limit while in the CW-TUNE position. Second, as we've been discussing, different exciters require different values of ALC voltage. Third, the ALC setting can vary from band to band, the extent depending on its gain variation. Fourth, the input impedance of the linear can vary from band to band, changing the loading on the exciter.

The input impedance of a grounded-grid amplifier is a function of the tube design and the plate load impedance presented to the tube. This plate load impedance is determined by the characteristic of the load (antenna plus line) and how the output network is tuned. (Is loading accomplished by plate current tuned for dip or for maximum output?) Unfortunately, only one currently available unit (two-piece) has this facility; an older unit is no longer in production, and a yetto-be-proved unit has just appeared on the market.

ALC using grid and plate detection

A unique circuit bears discussion and is shown in **fig. 5**. Unfortunately, to the best of my knowledge, it was never exported to this country. This circuit is unique in that it samples both the grid and plate signals of the amplifier. The detector connected to the grid circuit senses the pulses of grid current when the grid is driven into the positive voltage region. This voltage is doubled and fed along with the sensed plate signals can be achieved by individual pot settings. The thing about this circuit that struck me was the similarity to the method of measuring linearity des-

cribed by Bruene in 1954. His linearity tracer sampled the input and output of the amplifier and portrayed them on a scope; why not use them to generate an error signal for the ALC function?

time constants

The discussion above has dealt with static operating conditions: measurements made on a point-bypoint basis. The amplifier, however, is operated in a dynamic condition with the stimulus being voice excitation. Now we must deal with additional considerations - the transient behavior of the system. First, there's attack time; the time for the ALC system to respond to the stimulus, or voice peaks. If this attack time is too long, the system will overshoot and "buckshot" will result. Second, the release time must be considered; this is how long it takes the ALC system to recover or return to normal gain. If release time is too long, the amplifier gain is unnecessarily reduced, especially after a loud blast into the microphone — it can result in an unpleasant pumping effect if too short.

what can be done

I've painted a gloomy picture for those who don't

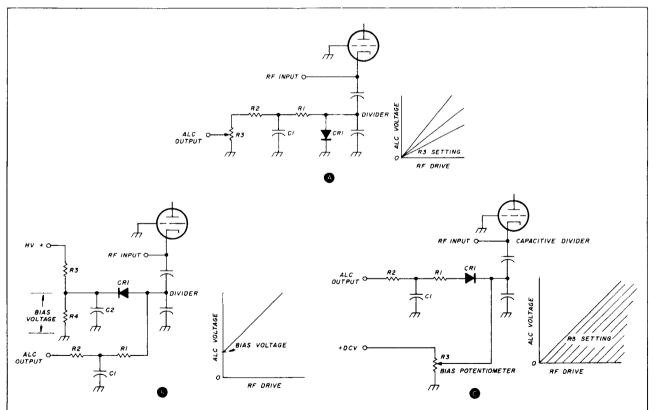


fig. 4. Examples of practical ALC circuits. Circuit at (A) is a simple system with no threshold and is adjustable by potentiometer R3. The most common ALC circuit is shown in (B), a system using a biased diode to determine the threshold point at which ALC action begins. In (C), the threshold point is made variable by using the potentiometer (R3) to set the bias to the diode.

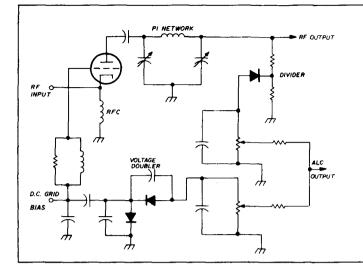


fig. 5. A unique ALC circuit using both grid and plate detection. When the grid is driven into the positive-voltage region, this voltage is doubled and fed together with the sensed plate signal to the ALC bus. The two voltages are balanced by individual potentiometer settings.

know the characteristics of their exciter or linear ALC systems — don't feel bad, neither do I! For those who say, "I watch my grid-current meter and keep the gain down until the meter just kicks," I ask, Do you know the ballistic characteristics of your meter? For those who say, "I watch my monitor scope pattern for signs of flat-topping," I ask, Do you know how much distortion is represented by how much flattening on the small pattern as it moves across the screen?

I think the same applies to those who use a trapezoidal scope pattern, although I feel it gives a fair indication of linearity. The correct method is to use a spectrum analyzer for a monitor; but I can neither afford one nor do I want to stare at scope patterns while I am talking!

If you're beginning to have reservations about using your 10-year old linear with your new exciter, or vice-versa, with ALC incompatibility problems, why not compare the manufacturer's specs on the ALC system constants? Can't find them? Neither can !!

What's needed is standardization in the industry — ALC voltage polarity, magnitude, intended load impedance, attack and release times. Having sat in on many industry standardization committee meetings, I can't give you much hope that this will happen. If military procurement or industrial safety requirements prevailed, it might happen. I don't find anything in FCC type-acceptance procedures that will help. As a matter of fact, examining one such submittal I find no mention of ALC connection, and the harmonic analysis was run with single, rather than two-tone, input and at a 50-watt input level!

Perhaps manufacturers will publish the ALC characteristics of their current and older models; it sure would be helpful. In the meantime, you can measure the voltage required to decrease your exciter's output by using a variable low-voltage dc supply, or batteries and a pot. You can a) measure your amplifier's ALC capability by terminating its ALC output with a resistor equal to your estimate of your exciter's ALC input impedance, b) excite the linear with a two-tone generator into the exciter with the linear amplifier terminated by a dummy load, and, c) measure the ALC voltage available (beware of meter loading) at the point where nonlinearity of the trapezoid scope pattern begins. At least you'll know if you need more or less ALC voltage from the linear amplifier. If you're lucky enough to have too much ALC voltage, you can add a pot on the front panel (or in a separate but accessible box) and attenuate the ALC voltage until you obtain the desired results.

final remarks

The problems of connecting ALC circuits in equipment of different manufacture have been discussed. Perhaps we can get more comprehensive data from the equipment manufacturers to assist us in properly interfacing our equipment. Consider measuring your equipment's ALC characteristics, putting a threshold control where you can get at it, and using a monitor scope. And remember, when you have that speech processor working full bore and your ALC line disconnected, you are *not* making many friends a few kHz up or down the band.

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ham radio

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a new look at **dip meters**

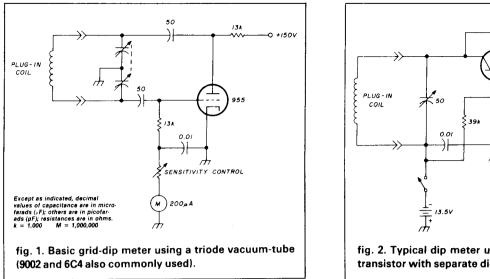
Some updated versions of that most useful tool — the grid-dip meter

One of the most useful accessories in the Amateur station is the grid-dip oscillator, grid-dip meter, or just plain dip meter. It can help solve many problems associated with resonant circuits. With proper calibration the dip meter can be used to determine the frequency of resonant circuits and, with capacitors and inductors of known values, it can be used to measure unknown inductances and capacitances. VHF parasitic oscillations can be detected in rf amplifiers, and antennas and transmission lines can be checked for resonance if proper precautions are used to determine harmonic responses with respect to fundamental resonant frequency.

The original grid-dip meter used a tube-type, class-C oscillator with a meter in its grid circuit. An external resonant circuit coupled to the oscillator's tank caused a decrease, or dip, in grid current; hence the name of the instrument. If this external circuit was a high-Q tuned circuit, the dip in grid current was noticed at a rather sharply defined frequency as the oscillator was tuned. It is precisely because the grid-cathode diode of the tube rectifies rf voltage that the oscillator operates in a stable class-C mode; and this rectified rf provides the dc grid current to operate a conventional dc current meter. A typical griddip meter is shown in **fig. 1**.

Historically, the oldest Amateur reference to griddip oscillators was that made by Bud Bane, W6WB, in the March, 1947, issue of *CQ* magazine. W6WB's meter used a 3A5 for battery operation, and covered 80 through 10 meters with three coils. Still earlier use of the grid-dip oscillator is discussed in the "Zero

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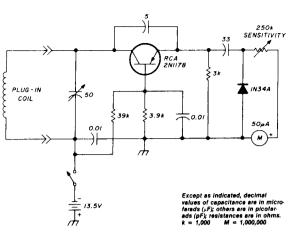
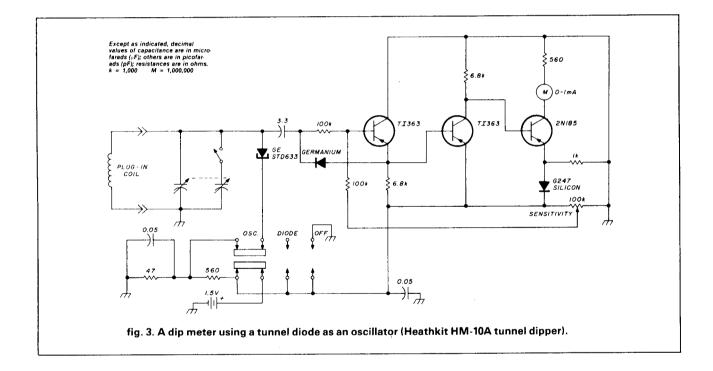


fig. 2. Typical dip meter using germanium PNP bipolar transistor with separate diode detector.



Bias" column of CQ, April, 1979, and in "Our Readers Say," CQ, October, 1979.

When bipolar transistors became available, they were almost immediately applied as the active elements in rf oscillators — and also applied in dip meters. The term *grid-dip* had to be shortened to *dip* since, of course, there are no grids in bipolar transistors. The bipolar transistor (whether NPN or PNP) is unlike the tube in a number of ways. The bipolar transistor is a current-operated device; that is, it requires base current to control collector current. This generally means that it is a lower impedance active device, and that the base has dc current flowing in it

whether oscillation is present or not. Even though the base-emitter junction of a bipolar transistor is a diode, the base current is not a very sensitive indicator of oscillation level because a good portion of the base current is dc bias current.

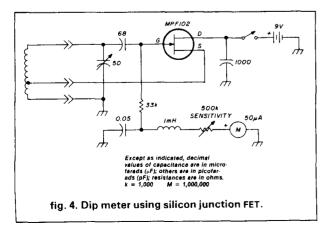
So we find, that, although a number of dip meters have been built using bipolar transistors, all of them make use of a separate diode detector to provide dc current for the meter.^{1,2} Most of these circuits use a 1N34A or some other germanium point-contact diode as the diode detector, and might be updated by use of one of the newer (silicon) Schottky diodes. A typical bipolar dip meter is shown in **fig. 2**. It is even possible (and one was commercially built by Heathkit for a time) to use a two-terminal form of oscillator: a tunnel-diode. In the Heath HM-10A Tunnel Dipper, a tunnel-diode was used as the oscillator and a standard diode detector was used as the rf detector to provide dc meter current. The Tunnel-Dipper is shown in **fig. 3** (reprinting is by permission of the Heath Company).

junction field effect transistors

The junction field effect transistor (JFET) is the first solid-state device to come along that really behaves in a "dip-meter" circuit the way tubes used to behave. Like a tube, the JFET requires voltage on the gate (grid) to control current in the drain (plate), and the gate-source junction operates like a "grid-leak" diode. So, with a lower supply voltage, an N-channel JFET can be put in an old tube-type dip-meter circuit and it will usually work immediately.³ Early silicon JFETs had rather limited performance at VHF, and it was not until the arrival of the TIS34/2N3823 (Texas Instruments) that a good, cheap, silicon JFET was available for Amateur dip-meter construction. Subsequently, the Motorola MPF102, Union-Carbide 2N4416, Siliconix 2N5197, and other excellent Nchannel, silicon JFETs became available inexpensively; these too fit dip-meter service quite well. An example of a dip-meter using an N-channel JFET is shown in fig. 4.4

the MOSFET

The MOSFET, unlike the JFET, does not have an inherent gate-source diode, and there are many MOSFETs that operate very well up through the VHF range. RCA and Motorola have a number of N-channel (depletion mode) MOSFETs that were designed for TV tuner service and are therefore inexpensive enough for Amateur dip-meter service. The 3N128 is a favorite in Amateur designs, and has been incorporated into at least one dip-meter design (see **fig. 5**).⁵ Since there is no gate-source diode in a MOSFET,

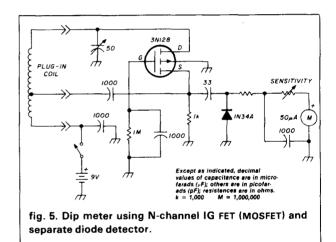


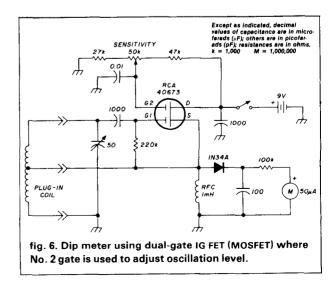
rectification of the rf is accomplished by adding a diode. It is even feasible to use one of the dual-gate MOSFETs, using the No. 2 gate as an oscillation level control (see **fig. 6**).⁶

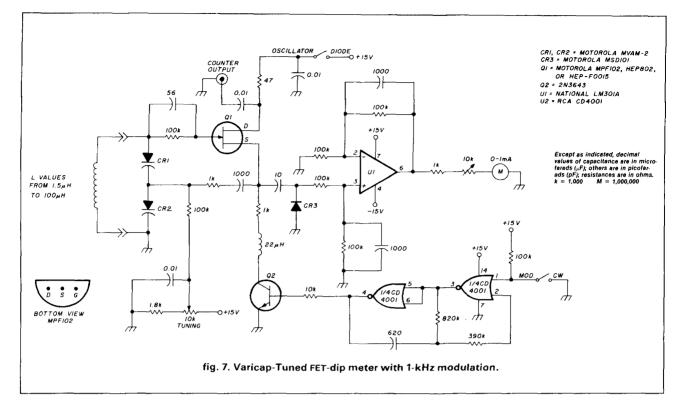
Note that in **figs**. **1** through **6** some of the oscillators use the Colpitts configuration and some use the Hartley circuit. Either form of oscillator will work, or any other form will do, as demonstrated in **fig**. **3**. The Colpitts circuit has the advantage over the Hartley that only two leads must be provided for the plug-in coils.

It is possible to build a dip meter without plug-in coils. This has the advantage that the accessory coils don't get misplaced. One author has written at least two articles on bandswitching dip meters.^{7,8} In reference 8, Fred Brown adds two worthwhile innovations to dip-meter technology: 1) a broadband 50-ohm output for driving a frequency counter, and 2) a square-wave modulation system which amplitude modulates at 1 kHz without fm.

With frequency counters now available for less than \$200.00 that cover the range 0 to 1000 MHz, the







time is perhaps right to dispense with the calibration dial on the modern dip meter. Once this step has been taken, it is only a logical, short jump to replace the tuning capacitor with varicaps and use a remote dc pot to tune the dip oscillator, since no calibration scales are now needed. If the counter is not desired, the dc pot can be calibrated just as well as could the traditional tuning capacitor dial.

One tempting way to combine voltage-tuning the dip meter and using a frequency counter as a readout would be a microcomputer-controlled dipper. The oscillator varicap input would be a ramp (sawtooth) generated by the microprocessor; the detected dc output of the oscillator would be A to D converted and fed to the microprocessor. The program would sense a point of minimum detection voltage (where the dc slope changes from negative-going to positive-going) and remember the frequency counter diaital reading at that point. This system could then simply be placed near the resonant circuit to be measured and the frequency would be read out digitally. The details of oscillator design, programming, integration with a frequency-counter, and provision for multiple dips could be a big development job; but it's certainly a system that seems feasible.

a simple dip meter

As a simple example of how one might build up a voltage-tuned, counter read-out dip meter, the circuit of **fig. 7** is shown. It was built to accept plug-in

coils with banana-pins with 3/4 inch spacing; the coils were borrowed for this breadboard circuit from an old tube-type grid-dip meter.

As shown, the circuit operates over the 2-32 MHz range with each coil covering about an octave of frequency. It would probably be worthwhile to have coil units made up with three terminals, so that varicaps are changed with the plug-in coil, to allow for lower minimum capacitance at the higher frequencies, a 2N5197 or other similar VHF FET might also improve things above 32 MHz.

Look at the circuit of **fig. 7**. The 1-kHz squarewave oscillator has been redesigned around a CMOS IC, and the meter's dc amplifier redesigned around an LM301A op amp IC.

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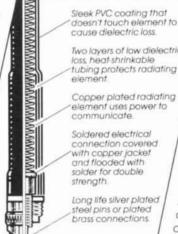
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For thousands of Amateurs, the following story is typical of what might have happened some years ago on a date "that will live in infamy."

the greatest DXer in the world

The streetcar gave a long wheeze, as if it were dying, and slowly ground to a halt at the corner. The accordian door creaked open and I jumped out, to be greeted by a swirl of dust and dead leaves stirred up by the ancient vehicle. With a clang of the bell it slowly resumed its journey down the cobblestone street.

Dodging a new Pierce-Arrow, I reached the curb and headed down a sidestreet. I could see Pendergast's house ahead and noted that he'd put up a brand-new Zepp antenna. It ran from the high chimney over to a tree in the back yard. The new copper wires gleamed in the winter sun.

I hadn't seen my friend for over a year. We had both been away at college, and this was the first time we both found ourselves home together. I knew from the few quick notes I'd received from him that he was working on a new DX-type transmitter. Now, he'd gotten home a few days ahead of me and possibly was already on the air.

I hastened my steps and trotted up the gravel walk. It was only a few steps to the wooden porch, which looked cold and empty. The summer lawn swing and folding chairs had long ago been put away for the winter and the storm windows were already up, even though the weather had been unusually warm for early winter.

Through the glass door I saw Pendergast's sister. She was sitting on the davenport, listening to a symphony on the radio. She was wearing an angora sweater, plaid skirt, bobby sox, and white loafers. I thought to myself, "Little sister is growing up."

"Hi, Miriam, long time no see," I said as she answered the door.

Miriam was going to be a knockout in a few years, I mentally noted. It would be fun to date the sister of a fellow Radio Amateur.

After a few pleasantries, Miriam motioned for me to go upstairs to the radio room, and she returned to the symphony, curling up once again on the davenport.

"Come on up," yelled Pendergast from the rear bedroom. I bounded up the stairs, two at a time, and saw my friend. He looked quite a bit older and had grown a mustache. He had on a white sweater with a big letter *C* on it. He looked very collegiate.

"Good to see you," said Pendergast, shaking my hand warmly. "Come into the shack and bring me up to date . . . and, incidentally, look at my new DX rock crusher." I followed Pendergast into what had once been his bedroom. How things had changed! Gone was the push-pull 210 TNT oscillator on 40 meters. Gone was the National SW-3. Down from the wall were all of Pendergast's cherished QSL cards from the stations he had worked while he was in high school. The wall was blank now.

Replacing all of the old gear was a brand-new National HRO receiver on the operating table and a new, gleaming transmitter sitting to one side on a small table. And there was a chromium-plated bug key on the table next to the receiver.

"You look older," said Pendergast, giving me a penetrating glance.

"I feel older," I replied. "That damned course in Descriptive Geometry last semester nearly finished me off. I pulled a *D* in it."

"I don't know why we have to study that stuff for an Electrical Engineering degree," Pendergast said. "Anyway, I'm running a good B-plus average so I can't complain."

Pendergast sat down in his operating chair and sighed.

"Is the transmitter ready to play?" I asked.

Pendergast smiled. "Yep," he said. "And 20 meters seems to be pretty good. I heard Guam this morning. Boy, could I use a KB6!"

"You have plenty of room on the wall for DX QSLs," I noted. "Now, suppose you tell me about this transmitter. It looks beautiful."

"Doesn't it?" said Pendergast. "I've worked on it night and day for two weeks to get it going in time for this vacation. Here's the circuit." (See fig. 1.) He handed me a large, white sheet with a carefully drawn schematic on it.

"Watch my lips as I explain everything and you'll understand all," he joked.

My friend flipped a switch and the filaments of the transmitting tubes lit up.

"O.K. Now, over here is a 6F6 metal-tube crystal oscillator on 7199.75 kilocycles. I am using a variable crystal so I can QSY from 14,390 to about 14,405 kilocycles."

"Plenty of DX to work outside the band edge," I remarked.

"No, no!" said Pendergast. "I have the band edge marked on the knob of the crystal with ink. No problem, there. Anyway, after the oscillator stage I have an 807 doubler to 20 meters."

"Were you able to tame the 807? It's a tricky tube," I remarked doubtfully.

"No problem. My double-E lab instructor is a ham and he told me to place a parasitic choke in the plate lead. It's a new idea and it really works."

I examined the tiny choke hanging

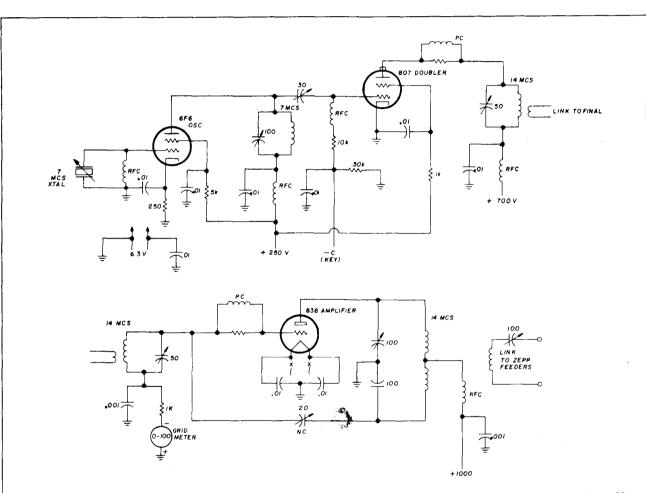


fig. 1. The 1941-model 20-meter DX transmitter. This transmitter delivers over 100 watts CW crystal controlled, on 20 meters. Representative component values are marked on the schematic. Pendergast built his transmitter in two sections. The 6F6 oscillator and 807 doubler were built on a small aluminum chassis. A variable crystal was used for quick frequency change. This was accomplished by a special crystal holder that had a variable air gap between the crystal blank and one holder plate. Adjusting the top plate changed crystal frequency. The 807 was keyed in the bias circuit. Cutoff bias was applied to the 807 and was removed when the key was closed. Pendergast soon found that a more complex keying system that reduced screen voltage under key-up conditions was required — otherwise a small back wave was heard in the keying intervals between characters.

The exciter was link coupled to the final amplifier stage, which was built on a separate aluminum chassis. Coaxial cable was generally unavailable in the 1940s, so Pendergast used twisted lamp cord for his link line. The 838 zero-bias triode was a forerunner of today's modern "grounded grid" tubes, such as the 3-500Z and the 8877. In this circuit, it was grid driven and plate neutralized. The split-stator tank circuit was coupled to a simple series-tuned circuit in the openwire feeders to the end-fed Zepp antenna mounted high atop Pendergast's roof.

Want to build an old-time DX transmitter? This circuit will work today for you!

from the plate cap of the 807.

"Maybe the reason the 807 has such a bad reputation is that it goes into a parasitic oscillation," I guessed.

Pendergast ignored my remark and hurried on. "I have link coupled the 807 to the final amplifier, which is an 838 tube, one of those new zero-bias triodes. A really great tube. No bias required at all. So I key the 807 and the 838 goes along for the ride."

"Look at it!" said Pendergast with pride. "The 838 has 100 watts of plate dissipation. I run it at 1000 volts and nearly 200 mils plate current. Just about 200 watts input. And with the link coupling to the 807 I get plenty of grid drive."

"How is it to neutralize?" I asked.

"No problem. I just apply excitation with plate voltage off and watch the grid meter as I increase neutralizing capacity. When I can turn the plate tank condenser all the way around without a kick in the grid current, it is completely neutralized."

"Very good," I admitted. "I see you are using a split-stator tank circuit."

"Yep," said Pendergast. "And now I think it's about time to fire up and work some DX before the band goes dead for the evening."

Pendergast turned on the receiver and soon I could hear the crystal note of the oscillator's harmonic on 20 meters. He closed the side contact on the bug and quickly tuned the 807 to resonance. The note in the receiver grew intolerably loud. Pendergast flicked the rf gain control and my ears returned to normal.

I noticed my friend tuned his transmitter like a pro. Now, he quickly neutralized the final and adjusted the variable link coupling. Then he reached down and threw another switch. With a growl the power supply came to life, the 866s casting a blue tint across the lower portion of the panel. Pendergast retuned a bit, fussed with the antenna coupling and then said, proudly, "We're on the air. How about a little DX now?"

"Suits me fine," I replied. "What

do you hear?"

Pendergast gestured to a second pair of headphones. "Let's tune the band," he said.

I donned the extra phones and listened. It had been almost a year since I had been on the air, and I was sure my CW was rusty. Pendergast tuned through a bunch of W9s and W6s as he approached the edge of the band.

"Listen!" Pendergast whispered to me. I closed my eyes to concentrate better.

"CQ...CQ...CQ..." Who was it? I opened my eyes and looked at my friend. "Certainly sounds DX-y," I said. "Listen to that fade on him."

The station faded into the noise and then started to come up again. It was a chirpy note, with a bit of an echo on it. Finally after an interminable wait, it signed:

"CQ de KF6ROV . . . KF6ROV . . . KF6ROV"

Pendergast grabbed a late copy of QST and rapidly flipped the pages until he came to the DX column.

"KF6 . . . KF6," he muttered to himself. "Who is *that*? Almost instantly he answered his own question.

"KF6 . . . that's *Baker Island*! Baker Island in the *Pacific*!"

"What a long CQ," I muttered. "By the time he signs, every ham in the world will be here."

"CQ de KF6ROV . . . KF6ROV . . . KF6ROV . . . K"

"Go get him, buddy-boy," I yelled.

Pendergast fired up the rig and reached for his key.

Out of the corner of my ear I heard a commotion downstairs. Miriam was calling something to us. There were excited voices talking back and forth. Hurried voices on the radio.

I looked at Pendergast. He was calling the KF6 now with his left hand and writing in the log with his right. Now, he was standing by. The earphones came to life.

We listened to the background noise for a second. Pendergast frowned as he heard his sister calling loudly to him as she ran up the stairs. Suddenly, there was the KF6. Calling Pendergast.

My friend let out a whoop and started to write in the log as Miriam burst into the room. Her face was white and she shook Pendergast by the shoulder as he wrote. I glanced in the log. Pendergast had just written: *December 7, 1941*.

The Japanese attack came just about sunrise, Hawaiian time, or about 1 o'clock EST. One of the first radio announcements was made by breaking into the New York Philharmonic Orchestra broadcast to announce the news flash.

Amateur Radio operation continued for a day or so. Many ham contacts were made with Hawaii, and confused reports of the raid filtered back via Amateur Radio and were picked up in the newspapers. On Monday, the FCC met in a special session and issued the following:

Order No. 87

At a session of the Federal Communications Commission held at its offices in Washington, D.C., on the 8th day of December, 1941;

Whereas a state of war exists between the United States and the Imperial Japanese Government, and the withdrawal from private use of all Amateur frequencies is required for the purpose of the National Defense;

IT IS ORDERED, that except as may hereafter be specifically authorized by the Commission, no person shall engage in any Amateur Radio operation in the Continental United States, its territories and possessions, and that all frequencies heretofore allocated to Amateur Radio stations under Part 12 of the Rules and regulations, BE, AND THEY ARE HEREBY, WITHDRAWN from use by any person except as may hereafter be authorized by the Commission.

By order of the Commission: T.J. Slowie, Secretary

ham radio

how important is **low** SWR?

This controversy is put to rest once and for all

What, exactly, is the meaning of SWR? What significance does SWR have in Amateur Radio applications? Is a low SWR important?

This subject has always been a popular topic for discussion, and one ham may tell you something completely different from another. This article is written to tell you the facts about SWR. Then, you should be able to determine whether or not you should try to improve the SWR of your antenna system, or perhaps change to a better feed line. (Chances are quite good that you won't have to do anything at all to your present system!)

definition of SWR

When asked exactly what SWR is, many people will answer with something like, "Well, SWR is what you get a reading of when you hook up an SWR meter to your feed line." Then, although they have just admitted they don't know what SWR is, they'll probably add, "If it's more than 1.5:1, it's bad."

SWR is the relationship, or ratio, between the maximum and minimum currents and voltages along a transmission line; hence its name: standing-wave ratio. Ideally, the current and voltage are constant all along a transmission line. This ideal condition occurs when, and only when, the load (antenna) and the line have *equal impedances*. However, this ideal situation is seldom the case in this imperfect world of ours. More often than not, the load and line have impedances that are different — sometimes by quite a lot.

When the load and the line have different impedances, the current and voltage distribution along the line is not uniform. In some places the voltage will be high, and in other places it will be low; the same is true of the current. The VSWR (voltage standing-wave ratio) is the ratio of the highest voltage on the line to the lowest voltage on the line. When talking about SWR, it is generally understood that we are talking about VSWR, although the current standing-wave ratio is theoretically the same.

The SWR is also the relationship between the antenna, or load, impedance, Z_L , and the characteristic impedance of the line, Z_O . Mathematically it gets rather complicated, but if the load is resistive, SWR is the ratio Z_L/Z_O or Z_O/Z_L (whichever is greater than 1).

Often, you'll see SWR given as a ratio, such as 3:1. However, since 3:1 simply means 3/1, mathematically we might just as well do away with the "1" and say that the SWR is 3. (Any number divided by 1 is just the number, right?)

reflected power

Nowhere is the SWR misconception more widespread than with respect to "reflected power."¹ To understand what this really is, we must first know what power is. Power is defined as the rate at which energy is expended. Direction has nothing to do with power. Power is just dissipated someplace or places. In an antenna system, most of the power is dissipat-

By Stan Gibilisco, W1GV/4, 6630 SW 72 Court, Miami, Florida 33143 ed, one hopes, as radiation. Some power is dissipated as heat in the transmitter final-amplifier tank coil, some is dissipated in the feed line, and some is dissipated in objects surrounding the antenna.

Reflected power is, literally, what a directional wattmeter reads when put in the reflected power mode! Reflected power is *not* a force (expended energy) moving toward your rig from your antenna.

Even though reflected power is somewhat fictitious, being caused as a wattmeter reading only because of certain phase relationships and ratios between feed-line currents and voltages,¹ it is very useful in certain antenna calculations. That is, *if* it is properly interpreted.

forward power

If reflected power is a fictitious invention because power has no direction, the same is true of "forward power." The reading on a directional wattmeter in the forward-power mode is the sum of the actual transmitter output power and the "reflected power."

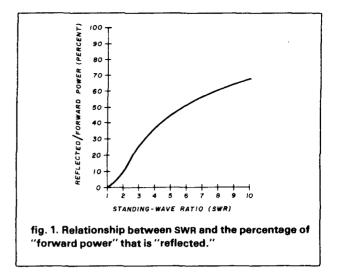
If you have ever used a directional wattmeter and observed that your transmitter was putting out 170 watts for a dc plate input of 200 watts, did you conclude that your rig was 85 percent efficient? If you did, you were probably wrong. (You know, of course, that 85-percent efficiency is virtually unheard of, even with class-C amplifiers.) Most likely, your SWR was high, causing an inflated "forward-power" reading. The actual output of the transmitter is the difference between the "forward power" and the "reflected power." (Think about it!)*

The relationship between the forward and reflected readings on a directional wattmeter is a function of the SWR. **Fig. 1** illustrates this function.

problems a high SWR may cause

A high SWR can, and sometimes does, cause certain problems. These problems include damage to the feed line, damage to transmitter components, reduced transmitter power output, and increased harmonic output. Let's discuss these one at a time.

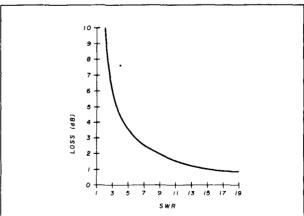
The transmission line. Coaxial feed line in particular, and also certain other types such as TV ribbon line, can withstand only so much current and voltage. If the current becomes too high, the conductors will get so hot that they melt the dielectric material. If the voltage gets too high, arcing may occur between the conductors, again damaging the dielectric. This kind of problem rarely happens with power input levels of under 200 watts; but if you plan on using a kilowatt, you had better consider this possibility. For

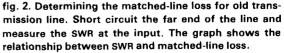


a given power level, the higher the SWR, the higher the observed current and voltage will be. If you're using RG-58/U or TV-type twin lead, *don't* use a power input over 200 watts. If you're using RG-8U with a kilowatt, beware if the SWR is more than 5.

Depending on the exact length of the transmission line, you might have a current or voltage maximum at your transmitter. Excessive current can cause the transmitter tank coil to overheat and become damaged. Excessive voltage can cause arcing in the output loading capacitor. Usually this problem occurs only with transmitters that have wide-range pi networks with components that aren't sufficiently rugged. It's also quite rare for SWR values of less than about 4; but it can still happen.

Components. Solid-state transmitters are often designed to work into a 50-ohm nonreactive load and have no matching circuits to compensate for other loads. With some of these rigs, there is an "SWR pro-





^{*}The power from the transmitter into the line is the sum of a) the power lost in and from the line and in any additional matching or other devices inserted into the line and b) the power delivered to the antenna (from reference 1).

table 1. Loss per foot, under perfectly matched conditions (SWR = 1), of various types of transmission lines commonly used by Amateurs. Accuracy is to two significant digits. (I foot = 3.05 meters.)

cable	characteristic impedance		free	quency (N	IHz)		
type	(ohms)	3.5	7	14	21	28	
RG-58/U*	52	0.0070	0.0100	0.0150	0.0190	0.0230	
RG-59/U*	75	0.0065	0.0090	0.0130	0.0160	0.0180	
RG-62/U*	93	0.0050	0.0072	0.0100	0.0130	0.0150	
RG-11/U*	75	0.0038	0.0055	0.0080	0.0100	0.0120	
RG-8/U*	52	0.0031	0.0045	0.0065	0.0081	0.0095	
TV ribbon							
lead	300	0.0020	0.0030	0.0042	0.0054	0.0061	

tection" circuit built in, but with many solid-state final amplifiers, damage can occur with an antenna system that presents a high SWR. If you have a solidstate rig, check your instruction manual to see if the final is "SWR protected." If it is not, check to see how high the SWR can be without risking damage to the output transistor(s), and be sure the SWR *is not above this limit*!

Output. Most commercially manufactured rigs will function all right if the SWR is less than 2 or 3. Some, with no output loading network, may tolerate an SWR of only 1.5 before power output is degraded. Certain of the older-vintage radios will compensate for SWR levels up to 5 or even 10. The radio's instruction manual should tell you how high the SWR can get before power output is degraded. If the SWR is beyond this limit, the final amplifier will not operate at full efficiency because the impedance mismatch will be too severe. If your loading control must be set to one extreme or the other to get a semblance of proper final-amplifier tuning, the SWR is too high for the circuit.

Harmonics. If the loading capacitor must be set all the way to maximum and you still can't get the transmitter to tune to the required power input level, you're being forced to under-load the transmitter. This can cause increased harmonic output, because under-loading causes nonlinearity in the final amplifier.

Of these problems, three can be eliminated by using a transmatch at the transmitter output. A "perfect" impedance match will then be presented to the transmitter. The transmatch will normalize the currents and voltages in the transmitter tank circuit, thus maximizing energy transfer and preventing under loading. The possibility of feed-line damage still remains, however, (as explained previously) unless the transmatch is placed at the point where the line feeds the antenna. This is generally not very convenient.

problems a high SWR doesn't cause

A high SWR, by itself, will *not* make a feed line radiate, and will *not* make an antenna less efficient. A perfectly matched feed line may radiate, and may be terminated by a grossly inefficient antenna (for example, a dummy load!). These problems are *not* related to SWR. If the transmitter can be properly loaded, a high SWR will *not* cause radio-frequency interference (RFI) that would not otherwise occur.

Surprisingly enough, at high frequency, a high SWR will generally not cause very much loss of signal. With several hundred feet of RG-58/U at 28 MHz and an SWR of, say 20, there will admittedly be a lot of loss because of the SWR. But this is an extreme case. In ordinary applications, the SWR can be surprisingly high before serious signal loss occurs.*

We will now investigate this phenomenon quantitatively. With the following information, you will be able to:

1. Determine how much signal loss your SWR is causing.

2. Find out whether or not it is significant.

3. Tell whether changing the feed line type, such as from RG-58/U to RG-8/U, is worth it, and perhaps

4. Save a lot of time, energy, money, and grief.

matched-line loss

Table 1 shows the loss in dB per foot for the most common types of transmission lines used by Amateurs. The numbers in **table 1** are based on perfectly matched systems; thus the antenna must have a pure resistance of 52 ohms for RG-8/U and RG-58/U, 75 ohms for RG-59/U, and 300 ohms for TV ribbon, for these data to be correct. However, the

^{*}Consider the end-fed Zepp antenna of 50 years ago. These antennas were efficient even with an SWR of 10 or more. **Editor**

table 2. Actual SWR as a function of matched-line loss (vertical) and the apparent SWR (horizontal). SWR values greater than 10 are indicated by x. SWR values above 6 are accurate to about \pm 0.5.

	apparent SWR (at transmitter)													
		1.0	1.2	1.4	1.6	1.8	2.0	2.5	3.0	4.0	5.0	6.0	7.0	
	0	1.0	1.2	1.4	1.6	1.8	2.0	2.5	3.0	4.0	5.0	6.0	7.0	
	0.2	1.0	1.2	1.4	1.6	1.9	2.1	2.6	3.1	4.2	5.5	6.8	8.0	
	0.4	1.0	1.2	1.4	1.7	1.9	2.1	2.7	3.4	4.7	6.3	8.0	9.5	
	0.6	1.0	1.2	1.5	1.7	2.0	2.2	2.9	3.7	5.2	7.6	10.0	x	
ଛ	0.8	1.0	1.2	1.5	1.7	2.0	2.3	3.1	4.0	6.0	9.0	x	x	
(dB)	1.0	1.0	1.2	1.5	1.8	2.1	2.4	3.3	4.3	7.0	x	×	x	
	1.2	1.0	1.3	1.5	1.9	2.2	2.6	3.6	4.8	8.3	x	x	x	
line loss matched	1.4	1.0	1.3	1.6	1.9	2.3	2.8	4.0	5.5	9.9	x	x	x	
Jat	1.7	1.0	1.3	1.6	2.0	2.5	3.0	4.3	6.5	х	x	×	х	
s	2.0	1.0	1.3	1.7	2.1	2.8	3.3	5.3	8.4	x	x	x	x	
los	2.5	1.0	1.4	1.8	2.4	3.2	4.0	8.0	x	x	x	x	x	
e	3.0	1.0	1.4	1.9	2.7	3.7	4.9	9.8	×	x	×	х	x	
3	3.5	1.0	1.4	2.1	3.1	4.6	6.9	×	×	x	×	x	x	
	4.0	1.0	1.5	2.3	3.7	6.0	10.0	x	x	x	x	x	х	
	4.5	1.0	1.6	2.6	4.7	7.9	x	x	x	×	×	x	х	
	5.0	1.0	1.8	3.0	6.0	x	×	x	×	×	x	x	x	
	5.5	1.0	2.0	3.4	8.5	х	x	х	x	х	x	x	х	

information in **table 1** is essential for determining the SWR loss in your system — so get it down.

For example, suppose you are using 60 feet of RG-58/U at 7 MHz. The loss is 0.01 dB per foot, so the total loss under perfectly matched conditions is 0.01 x 60 or 0.60 dB.

If your SWR happens to be 1, then you already know what the feed line loss is. You thus have no use for **tables 2** and **3**. But if the SWR is not 1, there will

be additional loss in the line. This additional loss is sometimes called "SWR loss."

Incidentally, the data in **table 1** are for brand-new cable. Coaxial cable gets lossier with age. Some types are better in this regard than others. Generally the foam dielectric type deteriorates faster than the solid dielectric type. There is a way to check old transmission line to find out what the loss really is. Short-circuit the far end of the line and place an SWR

table 3. Additional loss caused by standing waves. Find the line loss when perfectly matched in the vertical column; read across for the actual SWR. Find the figures that are closest to yours if yours are not exactly represented. Conditions above and to the left of the heavy line indicate an SWR loss of less than 1 dB.

	actual SWR (at antenna)												
		1.0	1.5	2.0	2.5	3.0	4.0	5.0	6.0	7.0	8.0	9.0	10.0
	0	0	0	0	0	0	0	0	0	0	0	0	.0
	0.2	0	0	0.1	0.1	0.1	0.2	0.3	0.4	0.5	0.6	0.6	0.7
	0.4	0	0	0.1	0.2	0.2	0.4	0.5	0.7	0.8	1.0	1.1	1.3
	0.6	0	0	0.1	0.2	0.3	0.5	0.7	0.9	1.1	1.3	1.5	1.7
ଳି	0.8	0	0	0.2	0.3	0.4	0.7	0.9	1.2	1.5	1.7	1.9	2.1
(qB)	1.0	0	0	0.2	0.3	0.5	0.8	1.2	1.4	1.7	1.9	2.2	2.5
	1.2	0	0	0.2	0.4	0.6	1.0	1.3	1.7	1.9	2.2	2.5	2.8
Ğ	1.4	0	0	0.3	0.4	0.6	1.1	1.5	1.8	2.1	2.4	2.7	3.0
line loss matched	1.7	0	0	0.3	0.5	0.7	1.3	1.7	2.0	2.3	2.6	3.0	3.3
S	2.0	0	0.1	0.3	0.5	0.8	1.3	1.8	2.1	2.5	2.8	3.2	3.6
os	2.5	0	0.1	0.3	0.6	0.9	1.5	1.9	2.3	2.8	3.1	3.5	3.7
e	3.0	0	0.1	0.4	0.6	1.0	1.5	2.0	2.5	2.9	3.2	3.7	4.0
÷.	3.5	0	0.1	0.4	0.7	1.1	1.6	2.1	2.6	3.1	3.4	3.8	4.1
	4.0	0	0.1	0.4	0.7	1.1	1.7	2.2	2.7	3.2	3.5	3.9	4.2
	4.5	0	0.1	0.4	0.7	1.1	1.7	2.3	2.8	3.2	3.6	4.0	4.3
	5.0	0	0.1	0.4	0.8	1.2	1.8	2.3	2.9	3.3	3.7	4.1	4.4
	5.5	0	0.1	0.5	0.8	1.2	1.8	2.4	2.9	3.3	3.8	4.2	4.5

meter at the transmitter output. Measure the SWR with a small amount of power. **Fig. 2** shows the relationship between SWR and perfectly matched line loss for this method.

The concept of reflected power is useful for obtaining the data for **fig. 2**. With a short circuit at the antenna end of the line, we may assume that 100 percent of the power is reflected, since a short circuit cannot dissipate power. Then, based on the SWR versus reflected power function, we can figure out what proportion of the power "came back" to the SWR meter to be registered as reflected power. The remainder was dissipated as heat in the line.

actual and apparent SWR

Most hams measure their SWR in the station at the transmitter output. However, measuring it at this point does not give an accurate reading. The *actual* SWR, the quality of the match between the antenna and the feed line, must be measured at the antenna if an SWR indicator is to give accurate results. The reading you get in the station is the *apparent* SWR, and is always less than the actual SWR. Sometimes the difference is considerable.

To accurately determine the SWR loss in your feed line, you need to know the actual SWR. This can be found out, of course, by climbing your tower or renting a "cherry picker" and physically placing an SWR indicator at the proper point. But there is an easier way.

If you know the line loss under perfectly matched conditions (from **table 1** or **fig. 2**), **table 2** will give you the actual SWR as a function of the apparent SWR.

Once you know the actual SWR as well as the line loss under perfectly matched conditions, you are ready to find the SWR *loss* — the extra feed line loss caused by standing waves on the line.

is it significant?

Use **table 3** to determine the SWR loss in your feed line. Does the result surprise you? How do you know whether or not it is significant?

In general, if the SWR loss is less than 1 dB, you are wasting your time by striving for a perfect match. A change in signal strength of 1 dB is recognized as the smallest detectable change; therefore, in practice, anything less is nothing. A gain of less than 1 dB will not be detected by the other station under any conditions. In **table 3**, the heavy black line differentiates between significant SWR losses (lower right) and those that are not significant (upper left).

Just because you have an SWR loss of less than 1 dB does not mean that you have a good antenna. It simply means there is no point in trying to get a bet-

ter match with the existing feed line. A 2000-foot (1 foot = 0.305 meter) run of RG-58/U at 21 MHz will be very lossy even if the SWR is 1. A 3-foot loaded vertical without radials at 3.5 MHz will have a lot of inductor and ground losses, even if it displays an SWR of 1. A small SWR loss does not make up for a poor antenna or feed line system.

The data here can be used to determine whether or not it is a good idea to change your feed line. Suppose, for example, that you are using 300 feet of RG-59/U with an antenna having a pure resistive impedance of 75 ohms, so the SWR is 1. At 21 MHz there will be 4.8-dB loss in the line (**table 1**), although none of it is attributable to the SWR. Would it be worthwhile to change to RG-8/U, which has a 52ohm characteristic impedance?

Using **table 1**, note that 300 feet of RG-8/U has 2.4 dB loss at 21 MHz. The SWR using this cable will be 1.5 (75/52), so there will be some SWR loss. How much? **Table 3** tells us it will be only 0.1 dB, giving an overall feed-line loss of 2.5 dB. This is an improvement of 2.3 dB over the perfectly matched run of RG-59/U! A change of 2.3 dB is fairly significant. If money is no object, this change would be worth-while.

The above is a rather unusual example, but in addition to showing you how to use these data, it illustrates another point as well: feed line loss is not a simple direct function of SWR.

conclusion

For all practical purposes, an actual SWR of 2.5 or better is just as good as a perfect match, as far as loss is concerned. This is apparent from examining **table 3**. So next time you try out a new antenna, don't throw your hands up in dismay and declare that it is no good just because your SWR meter reads 1.7!

The data presented here are primarily intended for the DXer or contester, since optimum antenna performance is especially important for them. Actually, a good antenna is important in any ham station. Of course we should try to optimize antenna systems. But there is a point of diminishing returns. The fear of an SWR that's not perfect or near perfect can take the fun out of hamming — not because there is anything actually wrong, but because you *think* there is. Now you know the real story. If your SWR loss is less than 1 dB and you're using the best grade of feed line you can, you have no more excuse not to get on the air and start operating!

reference

^{1.} For a detailed discussion of reflected power, see: Hubert Woods, "Power in Reflected Waves," *ham radio*, October, 1971, page 49. Also see "Comments," *ham radio*, December, 1972, page 76.

Amateur Radio equipment survey: synthesized 2-meter hand-helds

High-frequency rigs have received all of the attention in the *ham radio* equipment surveys thus far, and so the editors have decided that it's about time to let VHFers have a voice. This month, *ham radio* is surveying its readers' opinions of five of the most popular 2-meter hand-helds. The five are the Santec HT-1200, Tempo S-1 and S-5, ICOM IC-2A and 2AT, Yaesu FT-407, and Kenwood TR-2400.

All of these radios are frequency-synthesized, and all of them have now been around long enough for Amateurs to have had a chance to put them through their paces and come up with a fair appraisal. Twometer portables have become enormously popular in recent years, thanks mostly to the technological innovations that have made possible PLL synthesizers, multi-channel memories, and — just as important — reasonable prices. Attend any ham convention and you'll see just how well-liked these handhelds really are: they're everywhere.

The information that we're requesting is designed to help us get the clearest picture possible of what the owners of these radios think about them — what kind of problems they've had, what they think the best features are and what the worst. If you own or have owned one of these radios, please take the time to fill out the questionnaire. Only that way will other Amateurs be able to take advantage of your experiences.

The results of this survey will be published as soon as all of the results have been received and tabulated. It is not mandatory that you give your name and call, although we like to know who has replied. By checking the box marked *Yes* underneath your signature, you will be giving *ham radio* permission to quote from your comments and use your callsign. If you choose to check *No*, your wishes will be respected and your callsign will not be used. Those who do not wish to remove pages from their magazine are urged to photocopy the questionnaire, or send a stamped, self-addressed envelope to *ham radio*, Greenville, New Hampshire 03048; we will send you a copy.

Please mail in your completed form, one for each radio you want to report on, no later than September 30. Questionnaires received after that date may not be used. The readers of *ham radio* have always, in past surveys, been eager to let other hams in on their experiences with Amateur equipment — to let a fellow ham in on a good deal or warn him away from a bad one. Here is another chance to lend a hand to a fellow Amateur.

Owner's Report on 2-Meter Synthesized Hand-helds
(Please report only from your own experience. Type or print clearly. If you own more than one of these hand-helds, please use a separate report form for each.)
1. Make and Model (please circle the exact unit you are reporting on).
SANTEC HT-1200 TEMPO S-1 ICOM IC-2AT YAESU FT-207 KENWOOD TR-2400 S-5 IC-2A
2. What year did you buy it? New? Used?
3. Where did you buy it? Dealer Mail Order Individual Flea Market
800 Number Other
4. Would you buy from the same source again?
5. Amount of use: Daily Often Occasional Seldom
6. Is this your primaryor backup2-meter fm rig?
7. What is the rig's best feature?
8. Worst feature?
9. Have you had any problems? Explain
· · · · · · · · · · · · · · · · · · ·
10. Have you had the rig serviced? Where? Manufacturer Dealer Other
11. Was the service satisfactory? YesNo
12. What accessories have you purchased for this rig?
13. Have you been able to obtain all the accessories and parts you need?
14. Have you been satisfied with these accessories? YesNo
15. If not, why?
15. If flot, willy?

Additional features you would like to see built into a rig of this to	уре	
7. Give the equipment a score from 1 to 10 (with 1 being poorest,	4 to 6 average, and 10	perfect).
ase of operation		
Reliability		
Durability (in continuous use)		
nstruction Book		
Factory/Dealer Service		
Quality of Workmanship		
Performance		
Maintenance		
Parts Availability		
Accessories (ease of connection)		
Price		
Flexibility		
8. Would you buy this same rig again?		
What rig would you like to see reported on in the future?		
20. How long have you been licensed? Your Age	License Class	
21. (Optional: fill in the following only if you wish.)		
Submitted by: Name		Call
Address		
City	State	Zip
(Signature)		
(Signature)		
Your signature authorizes <i>ham radio</i> to quote portions of your comments in our report.) May we use your name and/or call?		
Yes No		
Completed survey forms must be returned no later th	an September 30, 198	31, to be included in our report.
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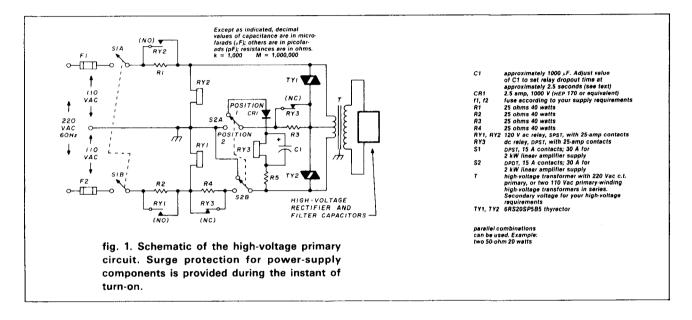


dual voltage surge-protection circuit for high-voltage power supplies

The circuit described here features dual output voltages with current surge protection. I developed the circuit so I could obtain, from one dc source, 2000 volts dc for my 3-500Z and 4000 volts dc for a 4-

charges, the voltage drop across resistors R1 and R2 decreases. When the voltage across the relay coils reaches the pickup voltage of the relays, the contacts of these relays close, shorting out resistors R1 and R2, which applies full voltage to the primary winding of the power-supply transformer.

When the output voltage switch, S2, is set for low voltage (position 1), the same process applies, but only relay RY2 operates. When switching the supply from low to high voltage, the voltage applied to the filter capacitor must again be increased slowly. This action is accomplished by components CR1, C1, R5, and relay RY3. When the power supply is switched from low to high voltage, capacitor C1 retains a voltage across the coil of relay RY3 for a time deter-



1000A linear amplifier conveniently and at minimum cost.

circuit description

Because of the very large secondary filter capacitor (or capacitors) in a linear amplifier power supply, it's necessary to apply voltage *slowly* to this capacitor to prevent tremendous current surges, which could destroy the power supply components. See **fig. 1**.

When the high-voltage power supply is first turned on, the filter capacitor presents a short circuit. This low impedance is reflected back through the highvoltage power transformer, T, and short circuits the coils of relays RY1 and RY2. The contacts of these relays are normally open, and resistors R1 and R2 decrease the 110 volts. As the filter capacitor

By Edwin Hartz, K8VIR, 108 Hartz Drive, Holly, Michigan 48442

mined by C1 and the internal dc resistance of the relay coil. (Time constant = RC.)

Since the contacts of RY3 are normally closed, resistors R3 and R4 are short circuited when the voltage across C1 decreases below the dropout voltage of the relay. This action brings the voltage of the filter capacitor to the high output-voltage value. An additional feature of the circuit is that, if a voltage breakdown short circuit should occur in the linear amplifier during operation, relay RY2 and/or RY1 will drop out and provide limited current surge protection.

This circuit has given excellent service. Components for the circuit are not critical as long as each has an adequate current and voltage rating.

bibliography

Marcellino, F.T., W3BYM, "Inrush Current Protection for the SB-220 Linear," (Weekender), ham radio, December, 1980, page 66.

Questions and Answers

Entries must be by letter or post card only. No telephone requests will be accepted. All entries will be acknowledged when received. Those judged to be most informative to the most Amateurs will be published. Questions must relate to Amateur Radio.

Readers are invited to send a card naming the question they feel is most useful that appears in each issue. Each month's winner will receive a prize. In addition we will have a drawing of all the voters each month, and that winner will receive a prize. We will also give a prize for the most popular question of the year. In the case of two or more questions on the same subject, the one arriving the earliest will be used.

radio propagation

In the "DX Forecaster" column, mention is made of the terms solarflux values and geomagnetic-field disturbances. I recognize these as the same as those I hear on WWV at 18 minutes after the hour. Could you explain these and other terms (A index, K index) and how they affect radiofrequency propagation? — Kathleen N. Freeman, KL7IFF

These terms were discussed by Jim Gray, W2EUQ (now W1XU), in the September, 1977, issue of *Ham Radio HORIZONS*. We have had other inquiries on this subject, so it seems worthwhile to explain these terms again.

solar flux and geomagnetic A indexes

For best propagation of radio signals on the high-frequency bands, a relatively high solar flux and low geomagnetic A index are desirable. The solar flux is measured at 2800 MHz and consists of radio noise whose magnitude has proven to be analogous to the magnitude of the energies that excite the E and F layers of the ionosphere. Solar-flux magnitudes range from about 65 to 150 and can be roughly equated to smoothed sunspot numbers and Maximum Usable Frequencies (MUF).

geomagnetic activity A index

This number tells how well the ionosphere has performed with respect to the reflection and absorption of radio signals, or both, particularly in the polar regions. An **A** index of 0-7 means excellent reflection of signals from the ionosphere, whereas an **A** index of 50 or more would indicate that a major magnetic storm is occurring, with severely disturbed, if not disrupted, radio-propagation conditions. Intermediate values indicate the relative condition of the earth's electromagnetic field.

solar K index

Also broadcast from WWV at 18 minutes past the hour, this data is updated every six hours. The K index ranges from 0 to 9 and corresponds to the A index of from 0 to 50 respec-

tively. **Fig. 1** shows how the various indexes relate to expected radio-propagation conditions.

tube or solid-state finals?

Would you please outline the primary advantages and disadvantages of tubes versus solid-state final amplifiers? – Lewis I. Hegyi, N2BPO

First, let's talk about power. Tubes are being used in most of today's high-powered linears. Solid-state technology, however, is beginning to produce devices that are capable of furnishing large amounts (several hundred watts) of rf energy. At least three manufacturers of Amateur equipment are now producing solidstate linear amplifiers that are capable of running at or near the Amateur legal limit. But it's probable that tubes will still be around for a while.

One of the advantages of tubes is that they are much more forgiving of mismatched loads. It's pretty hard to damage a tube amplifier when trying to make it work into a load presenting

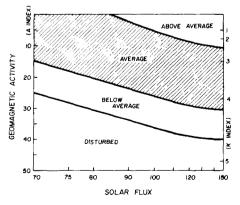


fig. 1. Relationship between geomagnetic and solar flux indexes and their combined effect on ionospheric propagation of radio signals.

a high standing-wave ratio. A transistor amplifier, on the other hand, will almost certainly be damaged under these conditions unless it has built-in overload protection. Also, tubes are easier to replace than transistors, especially in modern high-density solid-state circuits.

An advantage of solid-state finals is that transistors are smaller than tubes, which allows a lot of radio to be packaged into a small volume.

There's a difference in powersupply requirements for the two types of amplifiers. Solid-state amplifiers require supplies that produce large amounts of current at low voltage, while the opposite is true of tube rigs. An advantage of the solidstate amplifier in this case, of course, is that the danger of high voltages is eliminated.

selecting equipment

Having had no previous experience in Amateur Radio, I find myself drawing blanks when reading ads for ham gear. What are all those switches, dials, knobs, bells and whistles for anyway? How do I go about selecting ham gear when I have no idea what everything is or does? What are my options? — Dr. Crosby Pulliam.

Selecting ham gear pretty much depends on your sphere of interest, such as high-frequency communications, VHF/UHF, microwaves, or whatever. If you decide to work the high-frequency bands and want to select a commercially manufactured receiver, transmitter, or transceiver, you can write to the manufacturer of the radio of your choice and order an instruction manual. The manual will provide details of the radio's features, both from a theoretical and operating standpoint.

Or you can enlist the aid of a reputable ham-equipment dealer. Many have radios set up and operating, complete with antennas. Visit a dealer and ask for help in selecting a radio, then try it out (assuming you're licensed, of course). You'll find salespeople helpful and ready to answer your questions.

An alternative is to ask a ham friend to demonstrate his equipment. Some of the features you'll like, others perhaps not, depending on how you plan to operate. The controls and switches on today's equipment look mighty impressive to a beginner, but each one considered separately is not all that mysterious.

A good idea is to ask still another ham friend to demonstrate *his* equipment and explain its features. Keep in mind that your friends are partial to their equipment. Features that may appeal to one person may not appeal to others. Take a notebook and jot down your likes and dislikes.

For example, suppose you're interested in CW DX operating and want a receiver that will fit these requirements. You'll want a receiver with excellent mechanical and electrical stability as well as good sensitivity and selectivity.

Choose a receiver that has passband tuning, a notch control, and a sharp CW filter. With these three features you can virtually eliminate strong interfering signals and hear the desired DX signal, which would otherwise be inaudible. Passband tuning allows you to move the interfering signal around within the receiver's i-f passband. Then, by varying the notch control, you can null out the interfering signal or at least reduce its amplitude until you can hear the desired signal. Now, with the CW filter switched into the circuit, you'll have just about the best selectivity anyone could want.

A recommended book to help the beginner become acquainted with Amateur equipment is the ARRL's *Understanding Amateur Radio*, available from Ham Radio's Bookstore for \$5.00 plus \$1.00 postage and handling.

off-the-air tune up

I dislike the possible annoyance I might cause to other stations by put-

ting a signal on the air when matching my antenna to my radio. Is there a way to match my set to my antenna without causing interference? — Donald G. Ramras, M.D., KD6GR.

It would be great if others had your unselfish attitude toward this problem. Some hams are conscientious and take steps to tune their antennas without radiating an interfering signal. They use accepted methods for tune up and keep charts to indicate control settings for their equipment for different frequencies and bands. Others use the brute-force method and tune their rigs while on the air never mind the interference!

dummy load

First of all, you'll want to tune your transmitter to resonance without an antenna. A dummy load, representing the impedance into which your transmitter works, is the best device. A dummy load can be one of the many commercially available versions, consisting of a 50-ohm nonreactive resistor immersed in a cooling liquid, such as transformer oil. Most Amateur transmitters want to see a 50-ohm nonreactive load. After you've tuned up your transmitter into a dummy load of this impedance, you're ready to match the transmitter to your antenna.

Amateur antennas have an input impedance varying from 50 ohms (dipole or Yagi beam) to severalhundred ohms (extended double Zepp, multiple-band dipole, longwire). For antennas such as these, you'll need an impedance-matching device (sometimes referred to as a transmatch). This apparatus is designed to help you match your antenna impedance to that of your transmitter.

For complete off-the-air tune up you'll need, in addition to the dummy load and transmatch, a receiver with good frequency calibration, and an antenna noise bridge. The noise bridge will help you determine the resonant frequency of whatever antenna you are using.

the noise bridge

This instrument consists of two basic elements: a bridge circuit and a noise source, usually a diode. Connect the bridge to your receiver, and connect the receiver to your antenna. When the bridge is balanced, a null will occur in your receiver, indicating the resonant frequency of your antenna.

One arm of a typical bridge consists of a potentiometer and dial that indicates resistance. When you obtain a null, the potentiometer dial will indicate the resistance of the tuned circuit (antenna). The resonant frequency of the antenna will be indicated by your calibrated receiver.

Ideally, the null will occur only when the antenna input impedance is a pure resistance. Some noise bridges include a "reactance" control; others do not. If your noise bridge does contain a reactance control, this control should be adjusted, along with the other controls, to obtain a null. Noise bridges that do not contain a "reactance" control may or may not include a reactive component in their indication of antenna impedance.

When using the antenna noise bridge with your receiver, you'll probably have to adjust your antennamatching network (transmatch) to obtain a sharp null in the receiver and a resistance of 50 ohms, as indicated by the antenna noise bridge. Once these adjustments have been made, you can be sure that your antenna system is resonant and ready to take power from your transmitter.

Before connecting your transmitter to the antenna, however, the transmitter should have been tuned to resonance with its output connected to a dummy load. When this condition has been met, you know the transmitter is working into a nonreactive 50-ohm load. You can then disconnect the dummy load from the transmitter and connect the transmitter to your antenna, which has been tuned to resonance, as described previously. ham radio



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Garth Stonehocker, KØRYW

last-minute forecast

The higher frequency bands are expected to provide good DX openings during the first week of the month. Then DX conditions will probably wilt like a DXer in the August heat, while the nighttime DX will have the better openings. Of course, the QRN may take some of the joy out of that unless you're game for the early-morning, pre-dawn hours. The higher bands should again come into their own with good DX toward the end of the month. Disturbed propagation conditions may appear around the 8th and 25th for short, solar-flare-related periods of a one or two days' duration, then around the 17th for a coronal hole period of three to five days' duration.

The lunar perigee is on the 22nd and full moon on the 15th this month. The Perseids meteor shower occurs from the 10th to 14th, with its maximum the 11th and 12th at better than fifty meteors per hour.

August is a good month for shortskip openings by Sporadic-E (Es), but one of the last this Es season. The long hours of daylight will begin to fade noticeably by the end of the month.

forecasting

In the June DX Forecaster I gave some basic ideas on how propagation predictions are usually made. Here is some more shop talk about forecasting. Rather than trying to increase the accuracy of predictions by more closely defining variations of the ionosphere, better we shorten the length of the time in advance of when the prediction is needed - that is, until the day or hour before. We can of course try to keep track of the geophysical factors that could alter the prediction. This is like using low or high pressure areas, fronts, and winds in weather forecasting to say what the weather is going to be like. The geophysical factors found to be best for forecasting the ionosphere are the solar flux (in place of sunspot numbers) for predictions days in advance and the magnetic field variability within three hours (K figure) or daily (A figure) for predictions hours in advance.

As the solar radio flux goes up, expect the mid-latitude maximum usable frequency (MUF) to go up, after a two- to three-day delay. Or, as the geomagnetic field increases its variability (A or K becomes larger), the MUF will drop overnight. The amount of drop is a function of geomagnetic latitude, varying inversely from 30 degrees to the auroral zone (about 60 degrees). The MUF increases directly with geomagnetic latitude everywhere else. So you can forecast how the ionosphere propagation is going to behave from day to day or hour to hour by checking the daily solar flux value and geomagnetic A and K values from radio station WWV at 18 minutes after each hour.

Alternative to keeping track of the ionospheric propagation, you can plot signal strength for your favorite DX QSOs to a particular location. Plot each QSO signal strength at a point on a linear graph with abscissas of solar flux (70-200 units) and ordinates of the geomagnetic A value

(0-50 units). As the number of QSOs plotted grows, a family of similar signal strengths builds so as to be usable for forecasting signal strengths from solar flux and A values. For more detail and examples of this method see articles in CQ magazine for September, 1974, by Cohen; March, 1975, by Jacobs and Cohen; August, 1979, by Cohen and Jacobs; and in QST magazine for September, 1977, by Sartoni. It works! The North Atlantic forecast on radio station WWV over the years was done essentially by this method on a six-hour forecast period using the past two measured geomagnetic K values.

band-by-band forecast

Ten and fifteen meters should provide excellent daytime openings. The hours of daylight should begin to be noticeably shorter particularly toward the end of the month. Watch for the best openings during periods of high solar flux. They should be northsouth paths to Africa, South America, and the South Pacific. A sun-following sequence of Africa in the AM local time, South America in the early PM, and South Pacific in the late afternoon is usually seen from most QTHs in the U.S.A. Fifteen will be open a little before ten and will last a little longer in the evening.

Twenty meters will be open around the clock on most days and will be the long-skip/short-skip workhorse of the high-frequency DX bands. Signals will peak in the morning and afternoon hours, but will be readable to one area or another all the day and night.

Forty meters is going to start coming back strong, except for the high QRN levels during local thunderstorms.

Eighty and one-sixty meters will become active once again for DX during the nighttime hours, with strong openings into the south. High QRN levels will still be a problem generally, but they will be almost usable between local thunderstorms.

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*Look at next higher band for possible openings.

Hear there and everywhere.



Easy tuning, digital display, professional quality



The R-1000 is an amazingly easy-tooperate, high-performance, communications receiver, covering 200 kHz to 30 MHz in 30 bands. This PLL synthesized receiver features a digital frequency display and analog dial, plus a quartz digital clock and timer. Its easy-singleknob tuning and high sensitivity, selectivity, and stability make the R-1000 a favorite amongst Radio Amateurs, shortwave listeners, engineers, maritime communicators, and others who demand high quality in a general-coverage communications receiver.

R-1000 FEATURES:

- Continuous frequency coverage from 200 kHz to 30 MHz Receives shortwave, medium-wave, and
- long-wave bands. **30 bands, each 1 MHz wide** Easy-to-use band switch with large knob.
- Five-digit frequency display and analog dial

Accurate digital display with 1-kHz resolution and illuminated analog dial with precise gear dial mechanism.

- Built-in quartz digital clock with timer Precise 12-hour clock with AM and PM indicators. Timer turns on radio for scheduled listening, and even controls a recorder through remote terminal.
- Up-conversion PLL, wideband RF circuits

Provide exceptional performance and easy operation without the need for bandspread, preselector, or antenna tuning. Excellent sensitivity, selectivity, and stability.

Step attenuator

0-60 dB in 20-dB steps. Prevents overload.

- Three IF filters for optimum AM, SSB, CW
- 12-kHz and 6-kHz (adaptable to 6-kHz and 2.7-kHz) filters for AM wide and narrow, and 2.7-kHz filter for high-quality SSB (USB and LSB) and CW reception.
- Communications-type noise blanker
 Eliminates ignition and other pulse-type noise. Superior to noise limiter.
 Recording terminal
- For external tape recorder.
- Tone control For desired audio response.
- Built-in 4-inch speaker For quality sound reproduction.
- Dimmer switch
- Controls S-meter and other panel lights and digital-display intensity. • Three antenna terminals

Wire terminals for 200 kHz to 2 MHz and 2 MHz to 30 MHz. Coax (SO-239) terminal for 2 MHz to 30 MHz.

Selectable operating voltage AC voltage selector for 100, 120, 220 and 240 VAC. Also adaptable to operate on 13.8 VDC (with optional DCK-1 kit).

More information on the R-1000 is available from all authorized dealers of Trio-Kenwood Communications, Inc., 1111 West Walnut Street, Compton, California 90220.

Matching accessories:

- SP-100 external speaker
 HS-5 deluxe headphones
- HS-5 deluxe neadphones Other accessories
- not shown:
- HS-4 headphones
- DCK-1 easy-to-install modification kit for 12-VDC operation



HC-10 Digital World Clock Two 24-hour displays with quartz time base

Right display: local (or UTC) hour, minute, second, day. Left display: month, date, world time in various cities, memory time (QSO starting time), and time difference (in hours from UTC).

- Time in 10 cities around the world Plus two additional programmable time zones.
- "TOMORROW" and "YESTERDAY" indicators
- Memorizes present time
- And recalls later, for logging purposes. • High accuracy

±10 seconds/month



Ξ

New 2-meter direction.



A compact transceiver with FM/SSB/CW plus...

The exciting TR-9000 2-meter all-mode transceiver combines the convenience of FM with long-distance SSB and CW in a very compact, affordable package, ideal for mobile installation. With its fixed-station accessories it becomes the obvious choice for your ham shack.

TR-9000 FEATURES:

- . FM, USB, LSB, and CW All the popular 2-meter modes.
- Extended frequency range Covers all 2-meter Amateur frequencies as well as MARS and CAP frequencies (simplex and any repeater split) between 143.9000 and 148.9999 MHz.
- Digital dual VFOs With selectable tuning steps of 100 Hz,

5 kHz, and 10 kHz, convenient for each mode of operation.

- Digital frequency display Five, four, or three digits, depending on selected tuning step.
- Scan of entire band

Automatic busy stop and free scan.

Five memories

M1-M4...for simplex or ±600 kHz repeater offset. M5... for nonstandard offset (memorizes transmit and receive frequency independently).

SSB/CW search

Sweeps between 0 and 9.9 kHz around the selected frequency in 100-Hz steps, while the main knob selects in 10-kHz steps. Easy way to find SSB or CW activity. **UP/DOWN** microphone

- "Beep" sounds with each frequency step. (Supplied with TR-9000.)
- Effective noise blanker

Suppresses pulse-type noise on SSB and CW.

 Improved receiver front-end characteristics

Low-noise, dual-gate MOSFET and twostage monolithic crystal filter.

RIT control

Receiver incremental tuning, to tune only the receiver slightly off frequency in the SSB/CW mode. Functions on memory, also

RF gain control

Threshold-type control, permitting accurate S-meter readings on SSB/CW and FM modes.

CW sidetone

Enables monitoring of keying during CW operation.

Automatic AGC selection

AGC time constant selected automatically with MODE switch (slow for SSB and fast for CW)

 HI/LOW power switch
 10 watts/1 watt RF output on FM/CW. Always 10 watts on SSB. Improved power module for reliable and stable linear RF output.

- **LED** indicators VFO A/B, RIT, ON AIR, and BUSY.
- Rear-panel accessory terminals Key, memory back-up voltage, tone input, standby, external speaker, DC supply voltage, and antenna.
- **Compact size** Only 6-11/16 inches wide by 2-21/32 inches high by 9-7/32 inches deep.
- Adjustable-angle mobile mount With quick-release levers for easy removal.

More information on the TR-9000 is available from all authorized dealers of Trio-Kenwood Communications, Inc., 1111 West Walnut Street, Compton, California 90220.



Matching accessories for fixed-station operation:

- · PS-20 power supply
- SP-120 external speaker · BO-9 System Base ... with power switch, SEND/RECEIVE switch for CW operation, backup power supply for memory retention (BC-1 backup power adaptor may also be used for this application), and
- headphone jack MC-46 16-button autopatch (DTMF) UP/DOWN microphone

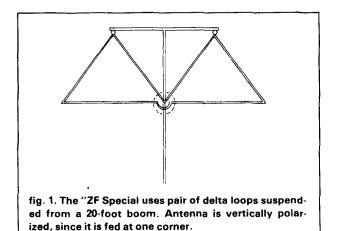


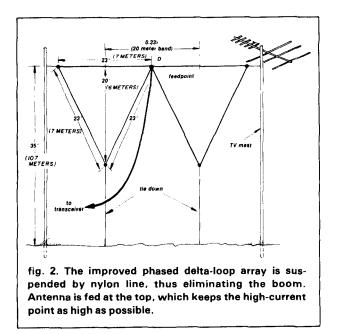
Specifications and prices are subject to change without notice or obligation.

two delta loops fed in phase

Another version of this popular antenna eliminates the suspension boom

Many months ago Ken Bale, W7VCB, sent me a copy of an antenna article describing a "ZF Special" antenna by ZF1MA in Grand Cayman, B.W.I. It used a pair of connected delta loops suspended from a 20-foot boom, as in **fig. 1**. Once in a while I would look at this idea and discard it, because 20-foot booms are pretty difficult to come by, not to mention to erect.





Then one day I said to myself, "Why a boom?" I turned the page upside down and mentally suspended the antenna from a nylon rope running from my 50-foot TV tower to a mast 48 feet away. It looked interesting. I'd often used both 20-meter and 15-meter loops suspended this way, but never two of them at a time.

I decided that what ZF1MA had was two full waves in phase, and that the array was vertically polarized, since they were fed at a corner.

improved phased delta-loop array

Fig. 2 shows the result of my efforts. I changed the feed point to the flat top to keep the high-current point at the top. Feed-point detail is shown in **fig. 3**.

So what do we have? A single delta loop has a gain of close to 2 dB referenced to a dipole. A pair of them at the close spacing shown will add another 2 dB. The pattern will be as in **fig. 4**.

If the antenna is oriented so that one lobe is toward Europe, the other lobe will be toward the South Pacific. Or you can orient the antenna so that Japan and the South Atlantic are covered.

My antenna is aimed at Antarctica in one direction and European USSR and the Persian Gulf in the other. It has provided excellent results with phonepatch traffic between my location and McMurdo Station, Antarctica. It has also performed well with many other DX stations on the 20-meter band.

If I had more space, I'd use the feed system in **fig. 5**, which provides horizontal polarization. (Incidentally, Bill Orr's Antenna Handbook shows that a delta loop fed in this manner yields a substantial, lowangle, vertically polarized signal.)

element-spacing considerations

Although the elements of my antenna are spaced as closely as possible without overlap (0.33 wavelength, or 23 feet, center-to-center, for the 20meter band), wider spacing would increase gain, as shown below.

For two elements (20-meter band):

spacing wavelength	feet	(meters)	gain (dB)
0.33	23	(7)	2.0
0.40	27	(8)	2.8
0.50	34	(10)	4.0
0.65	44	(13)	4.8

The gain numbers shown are in *addition* to the 2-dB gain for a delta loop over a dipole.

By Jerrold A. Swank, W8HXR, 657 Willabar Drive, Washington Court House, Ohio 43160

For more than two elements, with half- and threequarter wavelength spacing, here are the gain figures (from the ARRL *Antenna Book*):

number of parallel elements	half wave- length spacing gain (dB)	3/4 wave- length spacing gain (dB)
3	5	7.0
4	6	8.5
5	7	10.0
6	. 8	11.0

other arrangements

I can picture some lucky ham who lives on a farm and who could stretch out a series of six elements with half-wavelength spacing for 10-dB total gain; or with 3/4-wavelength spacing for 13-dB gain, suspended from trees or a row of telephone poles. Or how about two arrays at right angles to each other, with the intersection at a cross point of two fence rows, using no farmland space?

With an antenna gain of 13 dB, you'd have a power gain of 20, which would make a 2 kW PEP signal sound like 40 kW!

For two band operation, spacing for the 15-meter band at 3/4 wavelength would be the same as halfwavelength spacing for the 20-meter band. And you could suspend the 15-meter-band loops inside each 20-meter-band loop.

Each delta loop has a feed-point impedance of about 120 ohms, so a pair is about 60 ohms, which is what mine measure. For more loops, the impedance should be as follows, with each feed line being the same length:

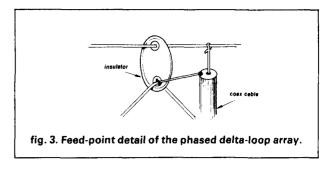
Four loops, 30 ohms with an SWR of 1.7:1 and six loops with a 2.5:1 SWR. None of these will be difficult to match.

It would probably be well to use a 1:1 balun at each feedpoint.

concluding remarks

Stations from the South Atlantic, the Caribbean, and South America really pound in here during the day, with many pinning my meter at 60 dB over S9. At night, the Antarctic stations are often over S9 and are easy phone-patch quality both ways.

Oh, yes — if you want to make a unidirectional beam, you can stretch a director or reflector element, or both, about 8 feet from the center of the array. Use a single wire 31 feet, 2 inches for the director and 34 feet, 6 inches for the reflector. This arrangement will, of course, reduce the feed point impedance, making some sort of tuner or matching section necessary to reduce the SWR. This is especially important if you have one of the new solid-state rigs, which are intolerant of high SWR.



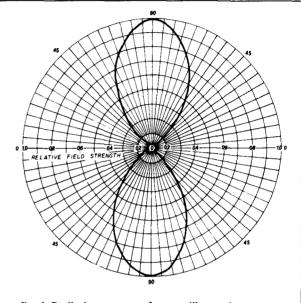
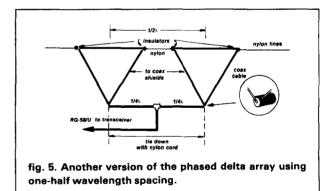
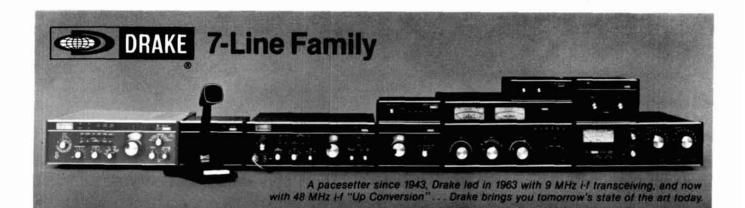


fig. 4. Radiation pattern of two collinear elements.



Antennas are great fun. They are much more inexpensive than amplifiers. Let me know how you make out - l'm always interested in what my antenna ideas do for others.

One caution: if you use a really high-gain array, a polar map is essential for pointing directions. I read about a ham who put up a ten-element Bruce array, and used a regular orthogonal map and completely missed his objective.





Full general coverage reception, 0-30 MHz, with no gaps or range crystals required.

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- · Complete front-end bandpass filters are included that operate from hf thru vlf. External vlf preselectors are not required.
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- Tunable i-f notch filter effectively reduces heterodyne interference from nearby stations.

Model 1240

• The famous Drake full electronic passband tuning system is employed, permitting the passband position to be adjusted for any selectivity filter. This is a great aid in interference rejection.

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Synthesized

General Coverage

- . Three agc time constants plus "Off" are switch-selected from the front panel.
- · Complete transceive/separate functions when used with the · Drake TR7 transceiver are included, along with separate R7 R.I.T. control.
- * Special multi-function antenna selector/50 ohm splitter is switch-selected from the front panel, and provides simultaneous dual receive with the TR7. This makes possible the reception of two different frequencies at the same time. Main and alternate antennas and vhf/uhf converters may also be selected with this switching network.
- The digital readout of the R7 may be used as a 150 MHz counter, and is switched from the front panel. Access thru rear panel connector.
- The built-in power supply operates from 100, 120, 200, 240 V-ac, 50/60 Hz, or nominal 13.8 V-dc.
- . Various optional selectivity filters for cw, RTTY and a-m are . The R7 includes a built-in speaker, or an external Drake MS7 speaker may be used.
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The outstanding Yaesu FT-707, FT-902DM, FT-107M or the FT-101ZD

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The Federal Communications Commission is quite explicit on the following point:

United States Amateur Regulations: 97.112 a) An amateur station shall not be used to transmit or receive messages for hire, nor for communication for material compensation, direct or indirect, paid or promised.

In other words, you can't make money from Amateur Radio. Or can you? Well, you can't charge fees for passing traffic, and running commercials is strictly *verboten*. However, there are legal ways of earning money indirectly from your hobby. As a matter of fact, the FCC says, in another part of its rules (Section 97.1), that the aims of the Amateur service are to provide an expansion of the existing reservoir...of trained operators, technicians and electronics experts. In the past, this section of the Amateur regulations has been interpreted by many as meaning a pool of experts available to the military in time of war, but it's obvious that such people can serve private industry as well. And Amateurs serving private industry means jobs for hams.

cashing in on Amateur Radio

Perhaps like no other hobby or pastime, Amateur Radio provides its participants with a real opportunity

By John Edwards, KI2U, 78-56 86th Street, Glendale, New York 11385

to acquire skills immensely useful in the world of time clocks and paychecks. The fact that this training can also be lots of fun is an extra benefit. Just think of some recent ham-radio contacts you've had. Odds are that half (probably more) of the people you spoke to were employed in some type of radio or electronics-related field. Now, whether ham radio got these people interested in their jobs or vice versa is subject to debate. But the fact of the matter is that ham radio and professional electronics go hand in hand.

What this means to you is that all those hours spent poring over electronics theory books and attending licensing classes didn't go to waste the moment you received your ticket. Depending on how high up the Amateur licensing ladder you've progressed, you may be qualified for scores of jobs with only a little extra training. Other, more advanced positions, may require additional formal study; but with an Amateur Radio head start, you'll probably jump to the top of your class in no time. Whether you're a high-school student looking for a career, a person in mid life frustrated with your present job, or retired and in need of some extra income, Amateur Radio may be the pathway to your financial goal.

types of jobs

There are many types of occupations open to jobseeking Amateurs. The FCC, in a published list,* has counted more than 65 different career categories covering radio alone. So, contrary to public opinion, not all hams professionally employed in electronics are electrical engineers (although many are), but run the full gamut of occupations within the industry. For the purpose of this article, I've narrowed the list down to five specific fields: broadcasting, television repair, electrical engineering, computing, and twoway radio servicing. While this compilation doesn't cover the entire spectrum of jobs having an Amateur-Radio connection, it should give you a rough idea of the choices available and show how you can match your skills and interests to a specific radio-electronics career.

broadcasting

Although nearly everyone seems to have forgotten by now, at one time Amateur and broadcast radio were one and the same. Stations such as Pittsburgh's KDKA and New York's WQXR (originally, experimental station W2XR), among others, were pioneered by Amateurs fascinated with the concept of radiotelephone communications. Only when advertisers realized the potential of mass electronic communications, and began pouring money into the

*A publication called "Careers in Broadcasting," routinely sent to those writing the FCC for job information.



Television repair, either as a full or part-time job, has long been a traditional field for Amateurs. (*Photo courtesy Hewlett-Packard*.)

broadcast radio service, did Amateur and professional radio enthusiasts part ways.

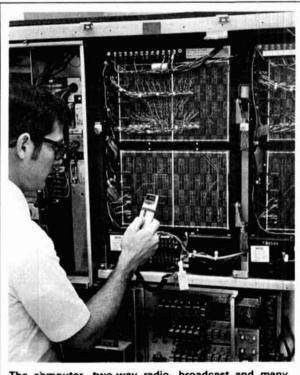
Practical experience. Even though the gap between Amateur and professional radio is now large, the distance separating the two isn't all that far. For instance, if you know how to build and operate an Amateur am transmitter, chances are you can do the same with a commercial rig. And, of course, when it comes to troubleshooting a circuit, it makes no difference whether the rig you're working



From technicians to engineers — as sex barriers continue to fall, more jobs in all areas of the electronics industry will open to women. (*Photo courtesy Hewlett-Packard*.)

on is Amateur or commercial. Actually, the greatest barrier separating technical people in the two services is the FCC. Just as the Commission has an incentive licensing ladder for hams, a similar structure also exists for broadcast engineers. To operate or service any type of commercial equipment within the United States, you must hold one of three levels of commercial radiotelephone licenses, which are designated as First, Second, and Third Class tickets. A First, in most cases, is required for broadcast station technical employment.

Technical requirements. The examinations for these licenses are rigorous (although there is some



The computer, two-way radio, broadcast and many other industries have a real need for trained technicians. And many newcomers will come from the ranks of Amateur Radio. (*Photo courtesy Hewlett-Packard*.)

disagreement on whether the Amateur Extra or First-Class Radiotelephone is tougher), but not beyond the comprehension of most Amateurs. While the material in these tests is much too detailed to cover here, your local FCC field office can provide you with the latest syllabus together with an examination time schedule for your area.

While a radiotelephone ticket is necessary to get your foot into broadcasting's door, you'll probably need some outside education, too. This can include technical school, junior college and even, in some lucky cases, on-the-job training. These days, in a switch from the past, many prospective broadcast engineers are also going the four-year college route. An extra benefit here is that most schools offering a broadcasting or communications major will have a radio station and television studio available for students for hands-on experience. Such practice, besides training the student, also gives him or her a chance to decide which specific area of broadcast engineering to enter.

Compensation. Pay for broadcast engineers is good, from about \$9,000 a year for a beginner at a small-town station to over \$35,000 for a chief engineer at a "Top 10" market outlet. In a middle-level position, pay ranges between \$15,000 and \$20,000 — somewhat higher at larger, unionized stations, and, of course, the networks.* In addition, scanning the want ads in *Broadcasting* magazine (available at most larger libraries) will give you a good idea of job opportunities around the country.

television repair

The receiving end of the broadcasting industry also has ample job opportunities for Amateurs. Although being a television service technician may not sound as glamorous as working at a radio or TV station, it's still an excellent field in which to make an adequate, steady income.

Independent shop or salaried employee? There are two traditional ways to earn money from television repairs. You can either open your own shop and be your own boss, or you can become someone else's salaried employee. If you operate your own shop, there's really no simple way of telling how much income you'll make in a year. You might hit the big time and decide to start a chain of television service centers in shopping malls across the nation, or you could fail and lose everything.

Factors affecting your chances for success will hinge on the size of your shop's community, potential competition and, certainly, your own business sense and technical acumen. On the other hand, working as a technician in someone else's business offers the comforting prospects of a weekly paycheck and job security. As a tradeoff, many TV technicians with an eye toward opening their own business will opt to put in at least five years work as a service shop employee before setting off on their own. This trial period gives the budding entrepreneur time to acquire the skills of the trade while simultaneously learning from his boss's mistakes. The average \$11,000-\$15,000 a year such an employee can expect to earn should also help to provide at least some of the nest egg for his eventual business.

^{*}For more information about careers in broadcast engineering, write to the National Association of Broadcast Employees and Technicians (NABET), 135 West 50th Street, New York, New York 10019.

Free lancing. Besides the two ways mentioned above, there's also a third path to making a TV servicing income that's especially appealing to part-timers — free lancing. With this method you can use your own ham-radio workbench to service TVs on a small scale in your own home. Of course, you'll have to scrounge your own customers; but by advertising in local newspapers and undercutting the high-overhead competition, you should be able to make more than a few bucks. Just be sure to fine out *in advance* if your community has any laws regulating TV repair shops.

Shoddy practices by a minority of TV technicians have forced severe crackdowns by some localities, and you'll want to be sure that you're operating within the law. But whichever route you take, if you're good at troubleshooting and perhaps have a year or two of technical-school training, TV servicing may be your answer.

electrical engineering

"If you've got the degree, we've got the job." That seems to be the rallying cry for EEs springing from virtually every major newspaper's employment pages these days. Whether you want to be a design engineer, technical writer, or consultant, it's certainly a job hunter's market with companies literally all over each other to get qualified applicants. So tight is the current EE job squeeze that many firms in California's "Silicon Valley" have even taken to building on-site tennis courts and swimming pools to attract workers.

What's available. Needless to say, with some employers even going so far as to fly applicants across the country at no charge just to look over the facilities, many EEs can almost write their own ticket. It's not unusual for someone in such especially desirable fields as microprocessor and energy work having say, five years' experience, to be earning upwards of \$30,000. Not bad. But there is a price to pay for such a rosy picture; and the cost here is education — the more the better.

While a bachelor's degree is a must, the everincreasing flood of technical information swamping the industry is making engineers holding masters degrees and even doctorates especially sought-after. Of course, you'll have to pay to get that education; but with the salaries currently being offered, you should be able to make the money back quickly. Many firms will even refund your education costs as a bonus for signing up.

Using your ham experience. Obviously, a ham ticket isn't as important here as it might be in broadcast engineering or TV servicing. Just because you passed a few FCC exams doesn't mean you can design an aircraft radar system. Yet, Amateur experience can be a big boost to your education. The mere fact that you hold an Amateur license is evidence of your interest in the field. While it may not be your only key to success, it will give you a leg up on the competition. And when it comes to the tough schooling an EE needs, every bit of an advantage helps. Anyway, who knows? With your Amateur Radio background, someday you may even get a job designing ham rigs!*

computing

As in electrical engineering, the job situation here is pretty wild. Although employers won't go to quite the same lengths to get qualified computer people as they do for EEs, the money is tops, the job market very open and, perhaps best of all, you won't have to spend six or more years at college to get started.

Depending on the exact area of computing you're contemplating, you may be able to get away with as little as a year or two at a technical school or junior college. Data processing and computer maintenance are just a couple of areas open to applicants with as little as a few months of classroom training. Of course, for higher-level positions in such areas as programming or analysis, you'll need a four-year college degree; but, as in electrical engineering, the financial rewards are substantial, so an educational investment can be most worthwhile.

Although the computer field isn't directly related to Amateur Radio, many hams do manage to combine the two. Just ask your friend who spends his weekends swapping programs, through ASCII RTTY, on 2 meters.

So, if you're the type of person who can't enjoy getting on the air without a keyboard at hand, you may be able to enter the computer field without any additional schooling at all — or you can finance your advanced education by working at a part-time computer job.

Depending on your exact skills, many openings exist for computer operators and program troubleshooters with virtually no knowledge of computer electronics but requiring a deep understanding of computer languages, something many advanced computer hobbyists seem to possess almost by second nature. Also, most companies specializing in games and other personal-computer software are constantly searching for people who can supply them with new and interesting software. So take a look around! If you regularly operate a home computer, chances are there's a job waiting for you somewhere.

^{*}For more information on electrical-engineering careers, drop a line to the Institute of Electrical and Electronics Engineers (IEEE), 345 East 47th Street, New York, New York 10017.

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ICOM 251A WWW Constants Konstant Repeater Offsets, Noise Band Scan, Repeater Offsets, Noise Band S	Other fine lines we carry: Ameco MFJ Amidon J.W. Miller Antenna Specialists Mirage ARRL March Astatic Radio Amateur Callbook Barker & Williamson Regency Bash Rohn Belden Sams Bencher Saxton Cushcraft Signals DenTron Trac Drake Turner Global Specialties Unadilla/Reyco Gold Line Valor Ham-Key Van Gorden Engineering Hydein W2AU Larsen W2AU

two-way radio servicing

A long-time career choice for many Amateurs, the two-way radio servicing business is experiencing something of a renaissance. An explosion in twoway radio use, combined with a serious drought of expert technicians, has created an ever-widening demand for skilled people to repair everything from police radios to CB sets.

Since two-way radio technicians (like their broadcasting counterparts) must make transmitter repairs, workers here are required to hold an FCC radiotelephone license. On the brighter side, however, only a Second-Class ticket is usually needed for most two-way equipment servicing.

Pay for these radio technicians is superb, especially when you consider that a reasonably intelligent Amateur holding a Second-Class commercial license doesn't need anything more than a basic high-school diploma to begin his career. Such a person can expect to make around \$10,000 a year to start, with salary prospects going as high as \$25,000 for a supervisory position, and even more for repair-department heads for such major employers as the airlines and government.

Uncle Sam

Speaking of the government, many good opportunities exist within Civil Service for radio technicians. Federal, state and local governments are in almost constant need of qualified people to repair and maintain radios for police, fire, safety, and other public services. For the latest line on government jobs in your area, write to your state or local Civil Service department.*

By the way, if you feel you can't drag your interests too far away from Amateur Radio, the U.S. Office of Personnel Management will also send you information on how to apply for a job with the FCC. Now, there's a job with a real Amateur-Radio connection!

final comments

So put your Amateur-Radio skills to work. There's a big, wide world out there crying for people with your technical knowledge, and they're willing to pay good money for what you learned from those study manuals. One last word: when applying for a job, be sure to mention *Amateur Radio* on your resume or application. More than one ham has been pleasantly surprised when that imposing person interviewing him or her for a job suddenly looked up from the form, smiled, and said, "Say, I'm a ham, too."

*If you're interested in working for the federal government, contact the U.S. Office of Personnel Management, 1900 E. Street Northwest, Washington, D.C. 20415, for details on Civil Service applications and examinations.



More Details? CHECK-OFF Page 94

DXer's diary

By Bob Locher W9KNI

I arrive home from work at the usual time. It was a full day for sure, and I'm a bit tired. But a good dinner – and a chance to relax a bit while reading the paper – refresh me and I'm ready to go. By 0030 Zulu, 7:30 Central Daylight Saving Time, I'm easing into the operating chair.

I flip the switches to turn on the gear and move the bandswitches to 15 meters. With the elevated sunspot count these days, 15 is queen of the DX bands during late summer. Usable during the day for DX, it really comes alive in the late afternoon and early evening, and probably will stay open all night if conditions are any good at all.

I'm hoping that LU3ZY will show up again tonight. This station is run by the Argentine Navy from the exceedingly rare South Sandwich Islands. Until this station became operational recently there had been no activity from these islands for over ten years, the last operation having been put on by a British VP8 station.

The islands harbor an environment hostile to man. The small group is actively volcanic, and a volcanic eruption in fact caused the emergency evacuation of a scientific team a few years ago. The island group is claimed by both Argentina and the United Kingdom, but neither party seemed too concerned about the matter until international law developed the concept of 200-mile zones of economic control around islands such as the South Sandwiches. Suddenly, these deserted, remote, Antarctic islands became the sacred soil of the motherland, with both Argentina and Great Britain heatedly arguing their claims.

But the DXCC program recognizes operation from either country, and I will be perfectly content to receive a QSL from either LU3ZY or a VP8, as long as it says, "South Sandwich Islands" on it.

I bring the antenna around to 151 degrees, the bearing my *Second Op* says is correct for the American Midwest bearing to the South Sandwiches, and I begin tuning the band.

LU3ZY is proving to be an elusive catch. I have heard him twice this summer, once on that memorable night when I snagged the A51 in Bhutan (the thought still brings a grin to my face — sure hope the QSL shows up soon), and I heard him again three weeks later, again on 15 CW. A YV station told me that he was often on 10 meters, but that won't do us much good in the Northern Hemisphere in summer. Openings are possible, but unlikely that far into the Antarctic.

The DX Tip Sheet shows that he has been operating a little on 20 meters: at 5 AM local one night, 2 AM another. I'd get up and watch for him if he would ever show a pattern; and I did stay up till 2 AM one night with insomnia. Sure enough, not only did he get on at 5 AM that morning, but when he did, he QSO'd VK9NV, the Norfolk islander that I still need. Aggh!

I just have not been able to get a pattern on him. And maybe there isn't one. Most DX stations have a pattern, and if you can get enough data, it usually is pretty easy to find the fellow, often requiring nothing more than a little bookkeeping and a little lost sleep. Of course, working him after you find him can be another thing, especially if lots of other hungry DXers also know the DX station's pattern.

But some fellows just don't follow a pattern, and they are the kind that give you grey hair. I've got a bit of help lined up: K6NA and I have a pact to call each other if we hear the LU3. Glenn is two hours behind us, so he'll be up later, which helps; and besides, I don't think he ever sleeps. With the score that he's got, he couldn't.

The band tonight seems in good shape. As I tune up from the bottom, strong signals from South America seem everywhere. There aren't all that many CW stations in South America but the ones you do find are superb operators, almost to a man.

But it's not South Americans that I'm after, it's the Antarctic islands. I keep tuning, listening especially for signals that are a little weaker. Of course, most of the signals that are a little weaker are from stations using low power and simple antennas, or they are stations coming in off the back or the sides of my antenna. But, in any case, there are lots and lots of South and Central American signals to filter through, and always the necessity of watching them to be sure one isn't a CE9 or an LU/Z station.

I tune the band slowly and carefully, but without result. I hear VP8s; all in the Falklands. I hear one of the Russian 4K1 stations in the Antarctic. There's a pretty good pileup on a PYØ, but it's Fernando de Noronha, an island off the Brazilian coast and one I don't need.

Almost before I know it, it's 0300Z, 10 PM local, time to close up shop. I muse on the evening's DXing as I head up the stairs to catch the news. Nothing at all to show, really which, of course, is the way it is most of the time. But it was still enjoyable, listening to the world talking. And I'll be back tomorrow.

August 8th

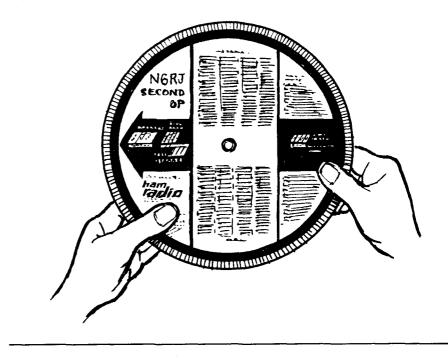
I start tuning about the same time as last night, with the same results as last night too. Lots of South Americans; little else. Band conditions don't seem to be as good. Maybe tomorrow.

August 10th

I couldn't get on the band last night. I had to stay late at work and got home late. Got to watching the game on TV during dinner and decided to stay with it, since it was already 9:30. The Cubs won again; their sixth straight on the road. LU3ZY was probably working all comers and having to call CQ to raise a little action, the way my luck runs. But the 2meter DX net was quiet, and no phone calls. Maybe it's tonight instead. I can always hope.

The band sounds a little better tonight. I barely lift the receiver above 21,000 when I hear FB8YC calling CQ. FB8Y is the French Antarctic, a curiosity station. A lot of DXers would given up next year's vacation for FB8W, which is Crozet, and FB8X and FB8Z are almost as rare; FB8Y counts as Antarctica, which is fairly common. But again, the station is on the Indian Ocean side of the Antarctic continent, so it is a long shot over a transpolar path to a station whose existence is widely know, but one that is rarely heard.

I bring the VFO up on him. He signs, and, pausing a second, I hear several stations start calling him. I call, but when I open the receiver again, I hear him calling W1FB. Regretfully, I decide to move on. I



don't need him for a country counter, and I do need the LU3ZY QSO. If I could have nailed him quick and easy, I would have, but I can't. Maybe later, after I've had a look across the band, I'll find him without a pileup. If I do, I'll try again.

With that behind me, I continue tuning up the band. It quickly becomes apparent that conditions are pretty good, with lots and lots of signals. Hey, there's a bit of a pileup. OK, it's 8R1G. And the pileup on him is mostly JA stations; some of them are pretty loud, and off the back of my antenna at that.

At 21,033 kHz, I run across ZD8TC negotiating his way through a crowd of admirers in contest style. And three kHz above him is VP5FM. And everywhere are South Americans; PYs, LUs, plus the occasional CE, ZP, and CX.

There's a weak signal calling a CO. OK, another JA. And there's KC4AAF, one of the American scientific stations in the Antarctic. Well, the band's open, for sure. I decide to speed up my tuning for a few minutes, and use that old Novice technique, "tune for the pileup."

Normally, I never use this technique on purpose. I much prefer to crawl up a band, listening to every signal at least for a moment. It's a heck of a lot more productive, and frankly, a lot less boring. In fact, it's fun, even if I don't catch a new one. A good open band is a dynamic, living thing, with a possible surprise at every kHz: to hear a band, to explore it, to hear signals from everywhere on different paths, to eavesdrop on old friends chatting or people making new friends.

Careful tuning brings other thrills, too. Digging a really rare one out, getting him in a one-on-one situation, having a nice ragchew with liberal use of his callsign — and then the enormous pileup on him when you sign clear with your 100 watts — it's heady stuff.

Tuning for pileups is a great way to

run up an SWL DXCC, but getting the QSO can be tough, really tough. Especially when you're after a rare one. Still, suppose that the LU3ZY station is already on the band? A half hour's careful tuning from the bottom up that leads to finding him at 21,080 with a horrendous pileup isn't necessarily the right way to do it either.

I start tuning higher, but much more rapidly. Instead of individual signals, I look for concentrations of signals; in a word, for pileups.

At 21,057, I find one. I should say centered on 057. This one's huge; hundreds of stations frantically calling; but calling whom? The calling stations seem to be spread out a bit, over at least five kHz. That's a pretty good sign that whomever they are chasing is working split frequency. And yes, they all start calling at once, and they all stop calling at almost the same time.

Whoever's running this pileup is a real pro; he's got everyone marching to his drumbeat. And the way they are calling so neatly in unison suggests that he has a pretty loud signal. He's probably lower in frequency; I drop back to 21,050 and have a look.

Almost immediately I find him; a good 579 signal, really pounding it out:

"W1DA 5NN K" . . . "R TU"

"4SMX ? 5NN K" . . . "R K4SMX EE" "N4AR HI BILL 5NN K" . . . "CU E E"

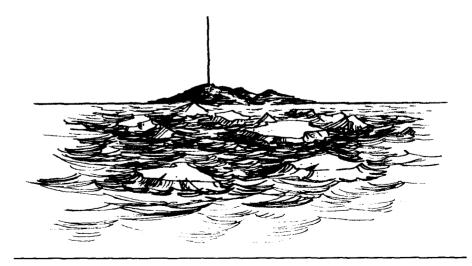
"PY7PO 5NN K" . . . "R QSL VIA KØHGB QTH SO GEORGIA 73 DE VP8MU QRZ UP8"

South Georgia. A real goody indeed, and one I don't need, but that's another story. I call it on 2 meters; "VP8MU, that's Victor Pappa Eight Mexico Uniform, South Georgia, twenty one oh five oh, South Georgia, twenty one oh five oh, listening up eight, he's got a good signal, here's W9KNI."

I get a terse response: "All right!" Hmm. Whoever that was really should sign a call. Oh well.

Anyhow, it's time to keep moving. I'm soon up to 21,150 and, as I suspected after hearing the South Georgian pileup, no LU3ZY. In a non-contest situation, it's rare to have two or more huge pileups on a band. If nothing else, two rare stations split the band population between them, limiting the size of either pileup.

Anyhow, it's back to slow, careful tuning for me, now that I'm pretty confident that the LU3ZY station isn't a featured guest in a pileup. I move the receiver back down to the bottom of the band and start tuning up higher again. My interest is heightened now, helping me to keep the concentration needed for the job at hand. Hearing that VP8 on South Georgia has gotten my fires lit, even though I don't





need him. After all, South Georgia and the South Sandwich isles are neighbors in that frozen part of the globe. Maybe there is hope. The band is open, at least, and that pileup is so huge that there can't be too many people after LU3ZY if he does show.

Another factor is in my favor. It's summer here and winter there: the long night of the polar regions, and so cold that the personnel at the station have to pretty well stay inside. Maybe there's a shortage of entertainment; operating 15-meter CW could be an ideal pastime.

I tune much more carefully this time, more the way I prefer to tune. If I didn't have a specific target (that is, the LU3) I would tune through 100 kHz, then rotate the antenna 90 degrees and do it again. If the band is open at all, it can be a fascinating exercise, listening to the world. I listen to this signal, "QTH HR BUENOS AIRES," and that signal, "QSL SURE VIA BUREAU," obviously a statesider with backscatter, and if he is going to QSL via the bureau he obviously isn't too excited.

Then, another backscatter statesider, "OK ISIDRO QSL VIA LU2CN TNX NEW ONE ES 73 . . ." That almost has to be him! The right name, the right manager, and a new country for whoever's in QSO. I punch the switch to turn on the linear. Bzzzt - zap - the surge relay drops out, and it's operational. It's already tuned on 15; I only need to trim the drive level from the exciter. I back off the drive control on the exciter; I will advance it to the proper level as I start my call.

There — the other station is signing — sure enough, "LU3ZY DE W3KT SK." I listen intently. There he is, almost dead on 3KT's frequency.

"R W3KT DE LU3ZY OK OM TNX ES 73 MUST ORT FOR DUTY W3KT DE LU3ZY SK CL EE."

Arrrhhhghl Ohhhhgh!

I call, desperately, a one by two. I listen. Nothing. Absolutely nothing; no competition and no QSO. I call again, but I know I'm dead. I listen. I'm right.

After a pause, I hear, "SRI HOPE U GET HIM NEXT TIME W9KNI 73 DE W3KT" That KT's a gentleman, a class act.

I respond, "R TNX JESSE ES CONGRATS UR FINE CATCH 73 ES GL W3KT DE W9KNI EE."

Gad. Maybe I should go back to stamp collecting.

But even as I consider such options, I make notes on my little chalk board, "LU3ZY 21032 0154Z," I make a mental note as well that the slight chirp he appeared to have last time I heard him seems to be gone; his signal was nice and clean this time. Hmmm. The time is interesting. Obviously his watch must change at 0200 Zulu, so that he has to go on duty, and QRT. I'm pretty sure that he could not have been on the band long, or there would have been a terrific pileup on him. I wonder how W3KT caught him? On a CQ maybe? Or hunting? 3KT's a sharp old fox, and his QSO with the LU3 was probably no accident. He knew what he was doing.

Well, I don't really have enough data to mark any kind of a trend, although I have a good deal more than I did yesterday at this time. But you better believe one thing — I'm going to be watching tomorrow around 21,032 at 0130Z like a hawk.

I turn off the gear and head up the stairs. Might as well go for a walk.

August 11

I had to work late again tonight, but still, I managed to get home in good time. My wife, sympathetic to the cause, has dinner ready when I get home, and I'm in front of the rig with a few minutes to spare against my self-appointed schedule.

OK. Antenna south-southeast. Check. Linear tuned for 15. Check. Attenuator out. Check. I settle the headphones into place, bring up the receiver gain — and there he is, square in the middle of the passband.

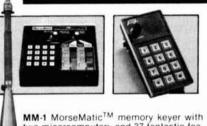
"CQ CQ DE LU3ZY LU3ZY AR K.



QRZ W1's, W2's and W3's...

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MM-1 MorseMatic^{1M} memory keyer with two micorcomputers and 37 fantastic features including up to 2000 characters of memory plus virtually every capability of all the other keyers & trainers listed below... call for super-low price! CK-1 Contest Keyer with 500 character memory, soft message partitioning, automatic serial number, and much, much more. call for super-low price!

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call for super-low price! KT-1 Keyer Trainer with all the features of the MK-1 above and the MT-1 below call for super-low price!

MT-1 Morse Trainer for pulling up that code speed the easy way with automatic speed increase, five letter or random word length and more, more. *call for super-low price!*



Bzzzt. Snap, the linear's on. I pause a moment as I back down the exciter drive level control. Ahh, no one else seems to be calling, as I yank the VFO dead on him. I start calling him and quickly bring the drive up as I go. I can't believe it; this is all happening so fast I haven't even had time to get excited or get butterflies yet. "LU3ZY DE W9KNI W9KNI AR K." I open the receiver without yet having to touch the receiver dial from where I left it last night.

Huh! There are two signals calling me! Hah! Yes, one is the LU3ZY station. OK, the other's an LU2. Guess he thought I was signing from a CQ. I ignore him.

"KNI DE LU3ZY R GE OC ES TNX QSO UR RST 589 589 NAME HR ISIDRO ISIDRO QSL VIA LU2CN LU2CN HW CPY W9KNI DE LU3ZY KN."

How sweet it is! "R LU3ZY DE W9KNI FB DR ISIDRO ES TNX NEW COUNTRY RST 579 579 HR NR CHICAGO NR CHICAGO ES NAME BOB BOB QSL SURE LU2CN OK? LU3ZY DE W9KNI AR."

"R FB BOB QSL SURE ES GLD BE NEW FOR U 73 ES GOOD DX W9KNI DE LU3ZY SK QRZ? K."

As he starts the QRZ, I shove the headset back, grab the telephone, and start punching digits. Let's see. It's almost 7 PM out there on the Coast; Glenn should be home by now. The phone rings, but no answer. I pull the headphone jack out to activate the speaker, and desperately start dialing again.

But wait! "HOW CPY? K6NA DE LU3ZY AR KN." Hah! No wonder the phone went unanswered.

"OK GLENN TNX QSO MUST QRT FOR DUTY 73 ES GL K6NA DE LU3ZY SK CL EE."

I keep listening. Sure enough, a couple of stations are calling, but I know how well that is going to work out. I swing the antenna around to the west. "K6NA UP 2 DE W9KNI." Glenn responds with a brief "R" and we go up for a very happy ragchew.

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Softbound \$4.75 RP-CO THE RADIO AMATEUR ANTENNA HANDBOOK by William I. Orr, W6SAI and Stuart Cowan, W2LX

If you are pondering what new antennas to put up, we recommend you read this very popular book. It contains lots of well illustrated construc-tion projects for vertical, long wire, and HF/VHF beam antennas. But, you'll also get information not usually found in antenna books. There is an information are intervented information and here. an honest judgment of antenna gain figures, information on the best and worst antenna locations and heights, a long look at the quad vs the yagi antenna, information on baluns and how to use them and some new information on the increasingly popular Sloper and Delta Loop antennas. The text is based on proven data plus practical, on-the-air experience. We don't expect you'll agree with everything Orr and Covan have to say, but we are convinced that **The Radio Amateur** Anterna Handbook will make a valuable and often consulted addition to any Ham's library. 190 pages. © 1978. Softbound \$6.95 RP-AH

BEAM ANTENNA HANDBOOK

Here's recommended reading for anyone thinking about putting up a yagi beam this year. It answers a lot of commonly asked questions like: What is the best element spacing? Can different yagi antennas be stacked without losing performance? Do monoband beams outperform tribanders? Lots of construction projects, diagrams, and photos make reading a pleasurable and informative experience. 198 pages. © 197 Softbound \$5.95 RP-BA

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1900 MHz to 2500 MHz DOWN CONVERTER This receiver is tunable over a range of 1900 to 2500 mc and is intended for amateur radio use. The local oscillator is voltage controlled (i.e.) making the i-f range approximately 54 to 88 mc (Channels 2 to 7). PC BOARD WITH DATA\$19.99 PC BOARD WITH CHIP CAPACITORS 13..... PC BOARD WITH ALL PARTS FOR ASSEMBLY \$69.95 PC BOARD WITH ALL PARTS FOR ASSEMBLY PLUS 2N6603 \$89.99 \$79.99 PC BOARD ASSEMBLED AND TESTED PC BOARD WITH ALL PARTS FOR ASSEMBLY, POWER SUPPLY AND ANTENNA..... \$159.99 POWER SUPPLY ASSEMBLED AND TESTED \$49.99 YAGI ANTENNA 4' LONG APPROX. 20 TO 23 dB GAIN . . \$39.99 YAGI ANTENNA 4' WITH TYPE (N. BNC, SMA Connector) \$64.99 2300 MHz DOWN CONVERTER HMRII, with dish antenna — 6 months warrantee \$200.00 2300 MHz DOWN CONVERTER \$200.00 Includes converter mounted in antenna, power supply, plus 90 DAY WARRANTY OPTION #1 MRF902 in front end. (7 dB noise figure)..... \$299.99 OPTION #2 2N6603 in front end. (5 dB noise figure) \$359.99 2300 MHz DOWN CONVERTER ONLY DATA IS INCLUDED WITH KITS OR MAY BE PURCHASED SEPARATELY **Shipping and Handling Cost:**

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(Satellite Receiver Boards)

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47 pF CHIP CAPACITORS For use with dual conversion board. Consists of 6 — 47 pF.	\$6 .00
70 MHz IF BOARD. This circuit provides about 43 dB gain with 50 ohm input and output impedance. It is designed to drive the HOWARD/COLEMAN TVRO De- modulator. The on-board band pass filter can be tuned for bandwidths between 20 and 35 MHz with a passband ripple of less than ½ dB. Hy- brid ICs are used for the gain stages.	\$25.00
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95H90DC	350 MHz Prescaler Divide by 10/11	\$9.50	2N1562	15.00	2N5591	11.85	MM1552	50.00
95H91DC	350 MHz Prescaler Divide by 5/6	9.50	2N1692	15.00	2N5637	22.15	MM1553	56.50
11C90DC	650 MHz Prescaler Divide by 10/11	16.50	2N1693	15.00		6.00	MM1601	5.50
11C91DC	650 MHz Prescaler Divide by 5/6	16.50	2N2632		2N5641			
11C83DC	1 GHz Divide by 248/256 Prescaler	29.90		45.00	2N5642	10.05	MM1602/2N5842	7.50
11C70DC	600 MHz Flip/Flop with reset	12.30	2N2857JAN	2.52	2N5643	15.82	MM1607	8.65
11C58DC	ECL VCM	4.53	2N2876	12.35	2N6545	12.38	MM1661	15.00
11C44DC/M	IC4044 Phase Frequency Detector	3.82	2N2880	25.00	2N5764	27.00	MM1669	17.50
11C24DC/M	IC4024 Dual TTL VCM	3.82	2N2927	7.00	2N5842	8.78	MM1943	3.00
11C06DC	UHF Prescaler 750 MHz D Type Flip/Flop	12.30	2N2947	18.35	2N5849	21.29	MM2605	3.00
11C05DC	1 GHz Counter Divide by 4	50.00	2N2948	15.50	2N5862	51.91	MM2608	5.00
11C01FC	High Speed Dual 5-4 input NO/NOR Gate	15.40	2N2949	3.90	2N5913	3.25	MM8006	2.23
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TOW BOO	ADBAND AMPLIFIER MODEL CA615B		2N3287	4.30	2N5942	46.00	MMT72	1.17
			2N3294	1.15	2N5944	8.92	MMT74	1,17
	response 40 MHz to 300 MHz		2N3301	1.04	2N5945	12.38	MMT2857	2.63
	300 MHz 16 dB Min., 17.5 dB Max.		2N3302	1.05	2N5946	14.69	MRF245	33.30
	50 MHz 0 to – 1 dB from 300 MHz		2N3304	1.48	2N6080	7.74	MRF247	33.30
Voltage: 2	24 volts dc at 220 ma max.	\$19.99	2N3307	12.60	2N6081	10.05	MRF304	43.45
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			2N3375	9.32	2N6083	13.23	MRF450	11.85
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Size: 3.20 m	m	3.58		2.80	2N6096	20.77		1.08
CRYSTAL	FILTERS: TYCO 001-19880 same as 2194F		2N3866JAN		2N6097	29.54	MRF504	6.95
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	arrow Band Crystal Filter		2N3924	3.34	2N6166	38.60	MRF511	8.15
	ridth 15 kHz min. 20 dB bandwidth 60 kHz min. 40 dB band	width 150	2N3927	12.10	2N6439	45,77	MRF901	5.00
kHz min.			2N3950	26.86	2N6459/PT9795	18.00	MRF5177	21.62
	dB: Insertion loss 1.0 dB max. Ripple 1.0 dB max. Ct. 0 + / -	- 5 pf 3600	2N4072	1.80	2N6603	12.00	MRF8004	1.60
ohms.		\$5.95	2N4135	2.00	2N6604	12.00	PT4186B	3.00
ABUDATA			2N4261	14.60	A50-12	25.00	PT4571A	1.50
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	FD-455D 455 kHz	\$3.00	2N4957	3.62	BLY568C	25.00	PT4628	5.00
	FB-455D 455 kHz	2.00	2N4958	2.92	BLY568CF	25.00	PT4640	5.00
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S	FE-10.7 10.7 MHz	5.95	2N4976	19.00	HEP76/S3014	4.95	PT9784	24.30
			2N5090	12.31	HEPS3002	11.30	PT9790	41.70
			2N5108	4.03	HEPS3003	29.88	SD1043	5.00
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TEST EQU	IIPMENT — HEWLETT PACKARD — TEKTRONIX	– ETC.	2N5179	1.05	HEPS3007	24.95	SD1119	3.00
Hewlett Pac	tkard:		2N5184	2.00	HEPS3010	11.34	TRWMRA2023-1.5	42.50
491C	TWT Amplifier 2 to 4 Gc 1 watt 30 dB gain	\$1150.00	2N5216	47.50	HEPS5026	2.56	40281	10.90
608C	10 to 480 mc .1 uv to .5 V into 50 ohms Signal Generator	500.00	2N5583	4.55	HP35831E/	2.00	40282	11.90
608D	10 to 420 mc .1 uV to .5 V into 50 ohms Signal Generator	500.00	2N5589	6.82	HXTR5104	50.00	40290	2.48
	450 to 1230 mc .1 uV to .5 V into 50 ohms Signal Generato		2113305	4.02	MM1500	32.20	40200	6.40
					1414113000	56.20		
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614A 616A 616B 618A 618B	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator	400.00 500.00 400.00 500.00					200-1	20-4
614A 616A 618B 618A 618B 620A	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7 to 11 Gc Signal Generator	400.00 500.00 400.00 500.00 400.00			1pf	27pf)0pf
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614A 616A 618B 618B 620A 623B 626A 895A	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set	400.00 500.00 400.00 500.00 400.00 900.00	value chip	p capac-	1pf 1.5pf 2.2pf 2.7pf	27pf 33pf 39pf 47pf	240pf 150 270pf 180 300pf 220	Nopf Nopf Nopf
614A 616A 618B 618B 620A 623B 626A 895A Alitech:	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7.0 to 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00	value chip		1pf 1.5pf 2.2pf 2.7pf 3.3pf	27pf 33pf 39pf 47pf 56pf	240pf 150 270pf 180 300pf 220 330pf 270	NOpf NOpf NOpf NOpf
614A 616A 618B 618B 620A 623B 626A 895A	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7 to 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator	400.00 500.00 400.00 500.00 400.00 900.00 2500.00	value chij itors you	p capac- may need.	1pf 1.5pf 2.2pf 2.7pf 3.3pf 3.9pf	27pf 33pf 39pf 47pf 56pf 68pf	240pf 150 270pf 180 300pf 220 330pf 270 360pf 330	Nopf Nopf Nopf Nopf Nopf
614A 616A 618B 618B 620A 623B 626A 895A Alitech: 473	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7.0 to 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00	value chip	p capac- may need.	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf	240pf 150 270pf 180 300pf 220 330pf 270 360pf 330 390pf 390	NOpf NOpf NOpf NOpf NOpf NOpf
614A 616A 616B 618A 618B 620A 623B 626A 895A Alitech: 473 Singer:	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator	400.00 500.00 400.00 500.00 900.00 2500.00 900.00 750.00	value chip itors you PRICI	p capac- may need.	1pf 1.5pf 2.2pf 2.7pf 3.3pf 3.9pf 4.7pf 5.6pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf 100pf	240pf 150 270pf 180 300pf 220 330pf 270 360pf 330 390pf 390 430pf 470	NOpf DOpf DOpf DOpf DOpf DOpf DOpf
614A 616A 618B 620A 623B 622A 623B 626A 895A Alitech: 473 Singer: MF5/VR-4	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7.0 to 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator	400.00 500.00 400.00 500.00 900.00 2500.00 900.00 750.00	value chij itors you	p capac- may need. ES	1pf 1.5pf 2.2pf 2.7pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf 100pf 110pf	240pf 150 270pf 180 300pf 221 330pf 271 360pf 330 390pf 390 430pf 470 470pf 560	NOPF NOPT NOPT NOPT NOPT NOPT NOPT
614A 616A 618B 620A 623B 620A 623B 626A 895A Alltech: 473 Singer: MF5/VR-4 Keltek:	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00	value chip itors you PRICI 1 to 10 11 - 50	p capac- may need. ES \$1.49 1.29	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf 100pf 110pf 120pf	240pf 150 270pf 180 300pf 220 330pf 277 360pf 330 390pf 390 430pf 470 470pf 566 510pf 680	Nopf Nopf Nopf Nopf Nopf Nopf Nopf Nopf
614A 616A 618B 620A 623B 620A 623B 626A 895A Alltech: 473 Singer: MF5/VR-4 Keltek:	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator	400.00 500.00 400.00 500.00 900.00 2500.00 900.00 750.00	value chip itors you PRICI 1 to 10 11 - 50 51 - 100	capac- may need. ES \$1.49 1.29 .89	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf 100pf 110pf 120pf 130pf	240pf 150 270pf 180 300pf 221 330pf 271 360pf 390 430pf 390 430pf 470 470pf 560 510pf 688 560pf 820	X0pf X0pf X0pf X0pf X0pf X0pf X0pf X0pf
614A 616A 618B 620A 623B 626A 895A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad:	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 GC 100 watts 40 dB gain	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00	value chi itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000	capac- may need. ES \$1.49 1.29 .89 .69	1pf 1.5pf 2.2pf 2.7pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf 100pf 110pf 120pf 130pf 150pf	240pf 156 270pf 186 300pf 222 330pf 270 360pf 330 430pf 390 430pf 477 470pf 566 510pf 680 560pf 822 620pf .01	XDpf XDpf XDpf XDpf XDpf XDpf XDpf XDpf
614A 616A 618B 620A 623B 620A 623B 626A 895A Alltech: 473 Singer: MF5/VR-4 Kettek: XR630-100	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 GC 100 watts 40 dB gain	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00	value chip itors you PRICI 1 to 10 11 - 50 51 - 100	capac- may need. ES \$1.49 1.29 .89	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf 100pf 110pf 120pf 130pf 130pf 150pf	240pf 150 270pf 180 300pf 221 330pf 271 360pf 330 430pf 390 430pf 470pf 560 510pf 680 560pf 820 620pf 0.01 680pf 0.01	X0pf X0pf X0pf X0pf X0pf X0pf X0pf X0pf
614A 616A 618B 620A 623B 626A 895A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad:	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7.8 to 7.2 Gc Signal Generator 7.10 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 Gc 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chi itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000	capac- may need. ES \$1.49 1.29 .89 .69	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 15pf 18pf	27pf 33pf 39pf 47pf 68pf 82pf 100pf 110pf 120pf 150pf 150pf 160pf 180pf	240pf 150 270pf 180 300pf 221 330pf 271 360pf 390 430pf 390 430pf 470 470pf 560 510pf 680 560pf 820 620pf 0.01 680pf 0.01 820pf 0.01	00pf 00pf 00pf 00pf 00pf 00pf 00pf 00pf
614A 616A 618B 620A 623B 626A 895A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad:	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 Gc 100 watts 40 dB gain 102A	400.00 500.00 400.00 500.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chi itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000	capac- may need. ES \$1.49 1.29 .89 .69	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf	27pf 33pf 39pf 47pf 56pf 68pf 82pf 100pf 110pf 120pf 130pf 130pf 150pf	240pf 150 270pf 180 300pf 221 330pf 271 360pf 330 390pf 390 430pf 470 470pf 560 510pf 680 550pf 820 620pf 0.01 680pf 0.01 820pf 0.01	X0pf X0pf X0pf X0pf X0pf X0pf X0pf X0pf
614A 616A 618B 620A 623B 626A 895A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad:	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7.8 to 7.2 Gc Signal Generator 7.10 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 Gc 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chi itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000	capac- may need. ES \$1.49 1.29 .89 .69	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 15pf 18pf	27pf 33pf 39pf 47pf 68pf 82pf 100pf 110pf 120pf 150pf 150pf 160pf 180pf	240pf 150 270pf 180 300pf 221 330pf 271 360pf 390 430pf 390 430pf 470 470pf 560 510pf 680 560pf 820 620pf 0.01 680pf 0.01 820pf 0.01	00pf 00pf 00pf 00pf 00pf 00pf 00pf 00pf
614A 616A 618B 620A 623B 626A 895A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad:	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7.8 to 7.2 Gc Signal Generator 7.10 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 Gc 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chi itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000	capac- may need. ES \$1.49 1.29 .89 .69	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 15pf 18pf	27pf 33pf 39pf 47pf 68pf 82pf 100pf 110pf 120pf 150pf 150pf 160pf 180pf	240pf 150 270pf 180 300pf 221 330pf 271 360pf 390 430pf 390 430pf 470 470pf 560 510pf 680 560pf 820 620pf 0.01 680pf 0.01 820pf 0.01	00pf 00pf 00pf 00pf 00pf 00pf 00pf 00pf
614A 616A 618B 620A 623B 620A 623B 626A 695A Alltech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad: 2038/2436/1	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7.8 to 7.2 Gc Signal Generator 7.10 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 Gc 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chip itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000 1,001 up	p capac- may need. ES \$1.49 1.29 .89 .69 .49	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 15pf 18pf	27pf 33pf 39pf 47pf 56pf 68pf 100pf 110pf 120pf 130pf 150pf 160pf 180pf 200pf	240pf 15(270pf 18(300pf 222 330pf 27(360pf 330(390pf 390) 430pf 47(470pf 56(510pf 68(560pf 820) 620pf 0.01 680pf 0.01 820pf 0.01 1000pf 0.01	00pf 00pf 00pf 00pf 00pf 00pf 00pf 00pf
614A 616A 618B 618A 620A 623B 620A 623B 626A 895A Alltech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad: 2038/2436/1	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7 to 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 Gc 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10 40 mc Single Tone Synthesizer N SOLID STATE RELAYS	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chip itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000 1,001 up	p capac- may need. ES \$1.49 1.29 .89 .69 .49	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 18pf 22pf	27pf 33pf 39pf 47pf 56pf 68pf 100pf 110pf 120pf 130pf 150pf 160pf 180pf 200pf	240pf 15(270pf 18(300pf 222 330pf 27(360pf 330(390pf 390) 430pf 470 470pf 56(510pf 68(560pf 820(620pf 0.01 680pf 0.01 820pf 0.01 1000pf 0.01	00pf 00pf 00pf 00pf 00pf 00pf 00pf 00pf
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614A 616A 618B 620A 623B 620A 623B 626A 895A Alitech: 473 Singer: MF5/VR-4 Kettek: XR630-100 Polarad: 2038/2436/1 HAMLIII 120 Vac a	900 to 2100 mc Signal Generator 1.8 to 4.2 Gc Signal Generator 1.8 to 4.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 3.8 to 7.2 Gc Signal Generator 7 to 11 Gc Signal Generator Microwave Test Set 10 to 15 Gc Signal Generator 12.4 to 18 Gc Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 Gc 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10 40 mc Single Tone Synthesizer N SOLID STATE RELAYS	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chip itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000 1,001 up ATLAS CRYS 5.52-2.7/8 5.595-2.7/8/U	p capac- may need. ES 51.49 1.29 .69 .69 .49	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 18pf 22pf	27pf 33pf 39pf 47pf 56pf 68pf 100pf 110pf 120pf 130pf 150pf 160pf 180pf 200pf	240pf 15(270pf 18(300pf 222 330pf 27(360pf 330(390pf 390) 430pf 470 470pf 56(510pf 68(560pf 820(620pf 0.01 680pf 0.01 820pf 0.01 1000pf 0.01	00pf 00pf 00pf 00pf 00pf 00pf 00pf 00pf
614A 616A 618B 618B 620A 623B 620A 623B 626A 695A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad: 2038/2436/1 120 Vac a Input Vol	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7.to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 GC 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10 40 mc Single Tone Synthesizer N SOLID STATE RELAYS at 40 Amps. Itage 3 to 32 Vdc.	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chig itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000 1,001 up ATLAS CRYS 5.52-2.7/8	p capac- may need. ES 51.49 1.29 .69 .69 .49	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 18pf 22pf	27pt 33pf 33pf 47pf 56pf 68pf 100pf 100pf 120pf 130pt 130pt 150pf 180pf 200pf	240pf 15(270pf 18(300pf 222 330pf 27(360pf 330(390pf 390) 430pf 470 470pf 56(510pf 68(560pf 820(620pf 0.01 680pf 0.01 820pf 0.01 1000pf 0.01	Nopf Sopf Sopf Sopf Sopf Sopf Sopf Sopf S
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614A 616A 618B 623B 623B 626A 695A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630-100 Polarad: 2038/2436/1 120 Vac // Input Vo 240 Vac // Input Vo	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug II TWT Amplifier 8 to 12.4 GC 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10 40 mc Single Tone Synthesizer N SOLID STATE RELAYS at 40 Amps. Itage 3 to 32 Vdc.	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chig itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000 1,001 up ATLAS CRYS 5.52-2.7/8 5.595-2.7/8U 5.595-2.7/SU	p capac- may need. ES 51.49 1.29 .69 .69 .49	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 18pf 22pf	27pt 33pf 33pf 47pf 56pf 68pf 100pf 100pf 120pf 130pt 130pt 150pf 180pf 200pf	240pf 15(270pf 18(300pf 221 330pf 27(360pf 33(390pf 39(430pf 47(470pf 56(510pf 68(510pf 68(560pf 011 680pf 011 820pf 011 1000pf 011	Nopf Sopf Sopf Sopf Sopf Sopf Sopf Sopf S
614A 616A 618A 618B 623B 623B 626A 895A Alitech: 473 Singer: MF5/VR-4 Keitek: XR630.100 Polarad: 2038/2436/1 120 Vac / Input Vo 240 Vac / Input Vo	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7 to 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 GC 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10 40 mc Single Tone Synthesizer N SOLID STATE RELAYS at 40 Amps. Itage 3 to 32 Vdc. at 40 Amps.	400.00 500.00 400.00 500.00 400.00 900.00 2500.00 900.00 750.00 1200.00 9200.00	value chip itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000 1,001 up ATLAS CRYS 5.52-2.7/8 5.595-2.7/8U 5.595-2.7/8U 5.595-2.7/8B	p capac- may need. ES 51.49 1.29 .69 .69 .49	1pf 1.5pf 2.2pf 3.3pf 3.9pf 4.7pf 5.6pf 6.8pf 8.2pf 10pf 12pf 18pf 22pf	27pt 33pf 33pf 47pf 56pf 68pf 100pf 100pf 120pf 130pt 130pt 150pf 180pf 200pf	240pf 15(270pf 18(300pf 221 330pf 27(360pf 33(390pf 39(430pf 47(470pf 56(510pf 68(510pf 68(560pf 011 680pf 011 820pf 011 1000pf 011	Nopf Sopf Sopf Sopf Sopf Sopf Sopf Sopf S
614A 616A 618B 620A 623B 620A 623B 626A 695A Alitech: 473 Singer: MF5/VR-4 Kettek: XR630-100 Polarad: 2038/2436/1 120 Vac a Input Vo 240 Vac a Input Vo 240 Vac a	900 to 2100 mc Signal Generator 1.8 to 4.2 GC Signal Generator 1.8 to 4.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 3.8 to 7.2 GC Signal Generator 7.10 11 GC Signal Generator Microwave Test Set 10 to 15 GC Signal Generator 12.4 to 18 GC Sweep Generator 225 to 400 mc AM/FM Signal Generator Universal Spectrum Analyzer with 1 kHz to 27.5 mc Plug In TWT Amplifier 8 to 12.4 GC 100 watts 40 dB gain 102A Calibrated Display with an SSB Analysis Module and a 10 40 mc Single Tone Synthesizer N SOLID STATE RELAYS at 40 Amps. Itage 3 to 32 Vdc. at 40 Amps. Itage 3 to 32 Vdc. 2000 State St	400.00 500.00 400.00 500.00 900.00 2500.00 900.00 750.00 9200.00 9200.00	value chig itors you PRICI 1 to 10 11 - 50 51 - 100 101 - 1,000 1,001 up ATLAS CRYS 5.595-2.7/8/U 5.595-2.7/8/U 5.595-2.7/8 5.595-2.7/8 5.595-2.7/8 5.595-2.7/8 9.0USB/CW	p capac may need. ES \$1.49 1.29 .89 .69 .49	1pf 1.5pf 2.2pf 3.3pf 3.9pf 6.8pf 6.8pf 8.2pf 10pf 12pf 15pf 18pf 22pf	27pt 33pf 39pf 47pf 56pf 68pf 100pf 110pf 120pf 130pf 130pf 180pf 200pf	240pf 15(270pf 18(300pf 22(330pf 27(360pf 39(430pf 47(470pf 56(510pf 68(560pf 82(620pf 01 680pf 01 1000pf 01 1000pf 01	Nopf Nopf Nopf Nopf Nopf Nopf Nopf Nopf
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MHZ electronics

MOTOROLA Semiconductor

\$21.83

NPN SILICON RF POWER TRANSISTORS

. . . designed for power amplifier applications in industrial, commercial and amateur radio equipment to 30 MHz.

 Specified 12.5 Volt, 30 MHz Characteristics – Output Power = 80 Watts Minimum Gain = 12 dB Efficiency = 50%



NPN SILICON RF POWER TRANSISTOR



MRF454

... designed primarily for use in large-signal output amplifier stages. Intended for use in Citizen-Band communications equipment operating at 27 MHz. High breakdown voltages allow a high percentage of up-modulation in AM circuits.

\$2.50

 Specified 12.5 V. 27 MHz Characteristics – Power Output = 4.0 Watts
 Power Gain = 10 dB Minimum
 Efficiency = 65% Typical

MRF475

NPN SILICON RF POWER TRANSISTOR

... designed primarily for use in single sideband linear amplifier output applications in citizens band and other communications equipment operating to 30 MHz.

- Characterized for Single Sideband and Large-Signal Amplifier Applications Utilizing Low-Level Modulation.
- Specified 13.6 V, 30 MHz Characteristics Output Power = 12 W (PEP) Minimum Efficiency = 40% (SSB) Output Power = 4.0 W (CW) Minimum Efficiency = 50% (CW) Minimum Power Gain = 10 dB (PEP & CW)

Common Collector Characterization

Tektronix Test Equipment

В	Wideband High Gain Plug In
ĘΑ	Dual Trace Plug In
ĸ	Fast Rise DC Plug In
N	Sampling Plug In
R	Transistor Risetime Plug In
¥	Righ Gain Differential Comparator Plug In
TU-2	Test Load Plug In for 530/540/550 Main Frames
IA2	Widehand Dual Trace Plug In
151	Sampling Unit With 350PS Risetime DC to 1GH7
2A61	AC Differential Plug In
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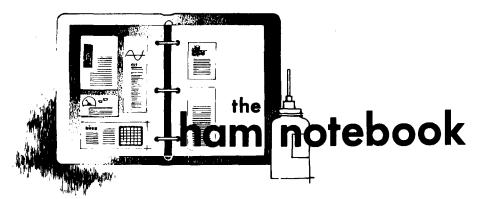
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squirrel-cage motors make field-day power supplies

One of the perrenial deterrents to portable and field Amateur Radio operation is the cost of a suitable energy source. Small gasoline engines can be had quite reasonably priced from lawnmower shops and Goodwill as-is stores, to name but two of many sources, but generators that produce 115 volts at 60 Hz are a scarce and expensive commodity. Or are they?

Used washing-machine and dryer motors are easy to come by. Goodwill, Salvation Army, and other usedmerchandise stores, or scrap-yards, sell such motors for from \$2 to \$5; most often you have to take the washing machine and do the salvaging yourself. Motors? But it's an alternator we need for the field power unit, not an electric motor!

the motor as

a generator

Fractional and multi-horsepower motors operate quite well as ac generators. Coupled to a suitable gasoline engine and driven at approximately the same speed at which they run as a motor, they will deliver alternating current power almost as well as an alternator designed for the job, and without any rewinding whatsoever. All that is needed is a little know-how and a willingness to tinker, plus a careful curbing of the standard ham tendency to overload equipment.

Washing-machine motors are generally rated at 1/4 to 1/3 hp and an occasional one may be found at 1/2 hp. These are supplied with resistance-starting, phase-splitting windings. **Fig. 1** shows a typical example.

Dryer motors are rated from 1/6 to 1/3 hp, and are as likely as not to be capacitor split-phased, sometimes just for starting, but occasionally for

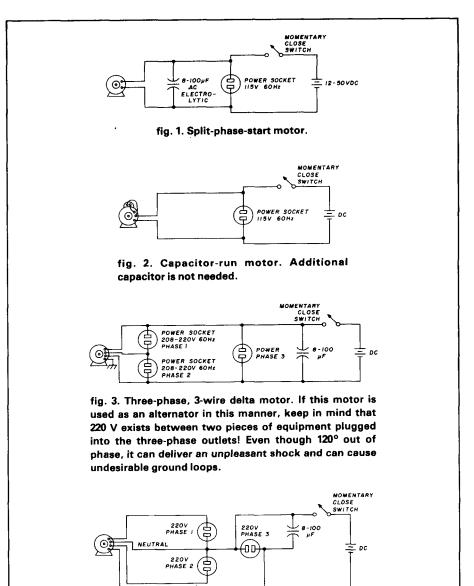


fig. 4. Three-phase, 4-wire star-connected motor. The note in *fig. 3* applies to this alternator as well.

unning as well. The key to coaxing a squirrel-cage motor (virtually all largeappliance, fractional-hp motors are his type) to perform as an alternator s in the use of capacitance to supply a field voltage.

split-phase-start motor

Without going too deeply into notor theory, a capacitor placed across the power leads of a motor creates a split-phase condition in the notor windings, which produces a resonance between the inductance of the winding and the induced current n the rotating squirrel-cage armature. Within certain power limits, the notor will perform as an alternator. Because such motors are "induction" notors with a 5 per cent slippage facor, the squirrel-cage must rotate about 5 per cent faster than synchronous speed to produce 60 Hz: a notor rated at 1750 RPM would have to be turned at 1890 RPM to obtain 60 Hz (synchronous speed: 1800 plus 5 per cent).

Sometimes such a motor used as a power source will start to deliver curent by itself as it comes up to speed. Frequently, however, it is necessary to supply a momentary dc pulse to the motor windings to induce magnetic-field poles into the soft-iron arnature. This pulse may be applied either before starting the engine, or after the motor has come up to speed. If the latter, it is obvious that the pulse should be just that -aoulse, because the dc source becomes a short circuit to the alternator output the moment power generation begins.

Such an alternator will not produce as much power as the labeled motor ating, for two or more reasons: effisiency losses, power-factor losses, and perversity. The latter, which seems somewhat facetious, will be found to hold quite true in practice.

capacitor-run motor

A split-phase, capacitor-run motor will not need an additional capacitor (**fig. 2**). It is ready to operate as an alternator with only the occasional

glossary of terms

induction motor — An ac motor featuring a primary power winding (usually the stator) and a unique "squirrel-cage" secondary (usually the armature). Currents, and therefore magnetic poles, are *induced* into the secondary by the rotating primary magnetic field; thus the armature rotates at slightly slower than synchronous speed, called "slip." The amount of slip increases with an increase in mechanical loading.

split-phase motor — Single-phase induction motors are not self-starting. They are momentarily converted to two-phase (polyphase, self-starting) motors by inclusion of additional primary power windings displaced 90° from the main windings. Occasionally the auxiliary windings are designed to operate full time rather than momentarily.

resistance-start motor — A split-phase induction motor employing a resistive auxiliary starting winding (copper alloy or iron wire) to shift the phase angle approximately 90°. Used almost exclusively for low-starting-power requirements, this winding is switched out of the circuit when the armature approaches near-synchronous speed.

capacitor-start motor — When greater starting torque is required for a split-phase induction motor, a capacitor is used in series with the auxiliary starting winding to provide a leading current vector of 90° , which momentarily converts the motor to a polyphase induction motor.

capacitor-run motor – A single-phase induction motor using a capacitor in series with a second primary winding displaced 90° from the first. This winding remains in the circuit full time, converting the machine to polyphase operation.

three-phase motor -A true polyphase induction motor with three or more power windings designed specifically to operate from a three-phase power source. It is self-starting.

synchronous motor — A motor that runs at a speed in step with the power-station frequency (fundamental or harmonic of the alternator speed). Often started as an induction motor and converted to synchronous operation as it approaches synchronous speed.

universal motor — A motor with wound field and armature, with commutator to make dc operation possible. When all windings are connected in series-aiding, it will operate on ac as well as dc, although the higher the ac frequency the more carefully the iron must be laminated to prevent excess eddy-current heating loss.

need for a momentary dc starting pulse. A resistance-start, split-phase motor will require from 8 to 100 μ F across the power terminals; final size to be determined by cut-and-try. Ac-rated electrolytic capacitors should be used, obtainable from motor-rewinding shops, although standard dc electrolytics may be used by connecting them in series, backto-back. Choose at least 300-volt ratings, as electrolytics must be derated for use in power ac circuits. The capacitance of a back-to-back combination is the capacitance of one they are not additive or subtractive in this circuit. The momentary starting pulse may be applied through a simple touch-switch circuit from a 12volt automobile battery.

three-phase machines

Larger squirrel-cage motors are sometimes available from discarded power-shop equipment. Table saws, planes, and sanders use motors rated from 3/4 to 2 hp, usually capacitorrun variety. Even larger motors are used for water pumping and many industrial demands, but usually are three-phase types (figs. 3 and 4). These are equally useful as alternators, but the output is three phase, although single phase across any two legs (except when four wires — see fig. 4). The voltage, however, is usually either 208 or 220 volts across each leg. When used as an alternator, it is necessary to parallel only one leg of the three with a capacitor.

It is almost impossible to burn out a squirrel-cage motor operating as an alternator. An overload will null the rotating field, and the alternator will simply cease to generate. Generation can be started again by a pulse of dc across the output leads. It may therefore be advisable to breadboard a power supply before building it into a fancy metal frame, to make sure that the motor (alternator) is large enough to supply the rig plus a small-wattage bulb for nighttime log-keeping or whatever. For test, parallel a number of bulb sockets and screw in one bulb at a time. Add to the load until the alternator suddenly ceases to function. The sum of the bulb wattages is your absolute top output power.

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Chapters include: Frequency and Power Measurements; Power Supplies; Proportional Temperature Control; I-f Amplifiers; Antennas; Television and Computer Data Links; and much more.

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This book is a basic primer for those interested in exploring the fascinating world of 10 GHz. Author Richardson writes in an easy-going style that gives you the impression that he's looking over your shoulder as you build the Gunnplexer system, from the first building blocks to the finished station. The book is profusely illustrated with clear, sharp photos and construction drawings — everything is here to get you on the Amateur 10-GHz band. It's a must for those who want to explore something more than the "dc bands." The Gunnplexer Cookbook, 16 chapters, 335 pages, softbound. Published by The Ham Radio Publishing Group, Greenville, New Hampshire 03048. Price: \$9.95 plus \$1.00 shipping and handling.

double zepp antenna

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(# 4GHZ MAG 15DB MGF1412 NF 0.8DB		\$28.5
@ 4GHZ MAG 18 DB		\$75.0
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50 OHM T NETWORK 3DB P		\$6 D.
SMA Chassis Hount Square		56 1
SMA Chassis Mount Plug sq SMA Chassis Mount Strip-lin	Flange	\$8.5 \$6.7
SMA Plug for RG-58 SMA Plug for RG-174		\$6.7 \$6.7
SMA Plug for 141 Semi-rigid	BONE	\$3.9
GUNN SOURCE 10 525 GHZ	105MW	
WR-90 WAVEGUIDE MOUT	NTING	\$37.0
50+20MW WR-90 MOUNT FILTER/MIXER 8 2 to 12 4GF	ING	\$39.0
WR90 MOUNTING HORN ANTENNA 18108 (\$30.0
10.525GHZ WR-90 MOUN WAVE GUIDE FLANGE WR-	TING	\$13.7 \$4.0
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Will plate Copper, Brass, Bro		
Nickel, Tin. Pewter, Gold and white metal alloys	i most	\$30.0
RF CA	BLE	
141 Semi-rigid Cable Appro Loss per 100 ft @ 4GHZ Pr	x 24 DB	
It i/- inch max length is 51 Other lengths by special or	1	\$4.0
PISTON TR	IMMER	S
	-3 pt 1-8 pt	\$2.5
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Coming Events ACTIVITIES "Places to go..."

ALABAMA: The Central Alabama Amateur Radio Association's 4th Annual Hamfest, Sunday, September 13, Civic Center, downtown Montgomery. Free admission; free parking. Air-conditioned Flea Market. Set-up 0600. Doors open from 0800-1500. Prize drawing 1400 CDST. Restaurants and motels within walking distance of Civic Center. Refreshments available. Talk-in on 146.04/.64 or 146.52/S. For information or reservations: Hamfest Committee, PO Box 3141, Montgomery, AL 36109.

ALABAMA: The Huntsville Hamfest (formerly the North Alabama Hamfest), Saturday and Sunday, August 15 and 16, Von Braun Clvic Center, Huntsville. Free admission. Prizes, exhibits, forums, air-conditioned indoor flea market and ladies' activities. Flea market tables \$3.00/day. Family tours of the Alabama Space & Rocket Center available. Limited camp sites with hookups. Talk in on 3.965 and 34/94. For information: Huntsville Hamfest, PO Box 4563, Huntsville, AL 35802.

ARIZONA: The Scottsdale Amateur Radio Club, Inc. will host the 1981 Southwestern Division ARRL Regional Convention, October 9-11, Ramada Safari Resort in Scottsdale. Talk-in on 146.52 and 147.84/24 MHz. Further information: S.A.R.C. Convention Committee, PO Box 3073, Scottsdale, AZ 85257.

CALIFORNIA: The Tri-County Amateur Radio Association's annual Hamfest, Saturday, August 8, 9 AM to 3 PM, Los Angeles County Fairgrounds (Thummer's Patio). No charge. Bring a picinci lunch. Refreshments available. A noon raffle is featured. Grand Prizes: Quasar 10" TV; Tempo S-1 Hand-Held. Winner must be licensed Ham but need not be present. Donation: \$1.00. Talk in on 146.34/94. For information: TCARA, PO Box 142, Pomona, CA 91767.

DELAWARE: The Sixth annual New Delmarva Hamfest, Sunday, August 16, Gloryland Park, Bear. 8 AM to 4 PM. Admission: \$2.25 advance; \$2.75 gate. YL and jr ops free. Tailgating or covered table space \$3.50. Refreshments available. First prize: ICOM IC-2A and many other prizes. Talk-in on 52 and 13/73. For tickets, map, info: SASE to Stephen J. Momot, K3HBP, 14 Balsam Rd., Wilmington, DE 19804.

GEORGIA: Augusta Amateur Radio Club's annual Hanfest will be held September 20, 1981, at the Julian Smith Casino. Prizes will be a DenTron Clipperton L, a Cushcraft A4 Tribander, and an Icom IC2A. Bingo for the family. Talk-in 34-94. Tailgating \$3.00 includes one ticket. Tickets \$1.00 each. Further information call Diane, WB4YHT, (404) 860-3700.

ILLINOIS: Illiana Repeater Systems 12th annual Danville area Hamfest, September 5 and 6, Georgetown, Illinois, Fairgrounds. Flea market, free parking, forums, family entertainment, many prizes and more. Gates open at 6:30 AM. Talk in 22/82 and 146.52. Tickets: \$1.50 advance, \$2.00 gate. For more info: Lowell Wells, WD9AFG, Hamfest Chairman, RR 3, Box 215, Danville, IL 61832. (217) 759-7560.

ILLINOIS: Radio Expo '81 sponsored by the Chicago FM Club will be held, rain or shine, on September 19th and 20th at the Lake County Fair Grounds, routes 45 and 120 in Grayslake. Grayslake is 30 minutes north of Chicago and 45 minutes south of Milwaukee. This year we will have a super large flea market with plenty of indoor and outdoor space, free with a gate ticket. Just bring your own table and chair or tailgate it. Parking is free. We will also have new camping sites complete with power hookups. There will be Ham seminars both Saturday and Sunday. YL's have a ladies program and door prizes both

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LLINOIS: State ARRL Convention and annual Hamfest sponsored by the Fox River Radio League, Sunday, August 23, Kane County Fairgrounds, St. Charles. Forums on antennas, DX and ARRL operations, contests and demonstrations. Advance tickets \$1.50, \$2.00 at gate. For tickets and information: SASE to Jerry Frieders, W9ZGP, 1501 Molitor Rd., Aurora, IL 60505. Commercial exhibitors contact Mike Pittard, KA9EVT, [312) 896-7383. Talk-in on 146.940 MHz.

LLINOIS: Peoria Area Amateur Radio Club's Peoria Superfest '81, September 19 and 20, Exposition Gardens, W. Northmoor Rd., Peoria. Gate opens 6 AM, commercial building 9 AM. Admission: \$2.00 advance, \$3.00 gate. Forums, Amateur/ computer product displays, Ilea market, Ladies and children's programs, camping facilities. Saturday night smorgasbord, Heritage House, 8209 N. Mt. Hawley Rd. Movies at hamfest site. Talk-in on 146.16/76, call W9UVI. For information and reservations: Superfest '81, 5808 N. Andover Ct., Peoria, IL 61615.

SOUTHERN ILLINOIS: Shawnee Amateur Radio Association's 25th anniversity Silver Jubilee Hamfest will be August 30 at JOHN A. LOGAN College in Cartersville, Illinois. Offerings include Air Conditioned Flea Market — Prizes — Forums — Computers — Food — Refreshments — Contests. For details QSL Bill May, KB9QY, B00 Hilldale, Herrin, IL 82948 or (618) 942-2511 days.

NDIANA: The Bloomington Area Amateur Radio Hams will hold their 4th annual "Hoosier Backyard Hamfest," Sunday, September 6, 7 AM to 5 PM, 2335 Vernal Pike, Bloomington. Admission: \$2.00. Door prizes, swap'n'shop, vendors, free set-ups, balloon rides, 50/50 drawing, refreshments, ATV demonstrations. Grand prize: Aptron Lab ATV Converter. Rain or shine. Talk-in 7.78/18, 04/64, 233.26/224.86. For information: Bob Myers, K9KTH at above address or call (812) 332-2433.

NDIANA: The Tippecanoe Amateur Radio Association's 12th annual Hamfest, Sunday, August 16, Tippecanoe County Fairgrounds, Teal Rd. and 18th Street, Lafayette. Grounds open 7 AM. Taik-in on 146.13/73 or 146.52. Flea market, dealers, fun and prizes. For tickets and information: J.B. Van Sickle, K9KRE, RR 1, Box 63, Westpoint, IN 47992.

MINNESOTA: The St. Cloud Amateur Radio Club's Hamfest, August 9, 8 AM to 4 PM, Whitney Senior Center, St. Cloud. Swapfest, prizes, refreshments. Talk-In on 146.34/94. For information: Mike Lynch, KA0HQS, 2115 First Street South, St. Cloud, MN 56301. (612) 251-2297.

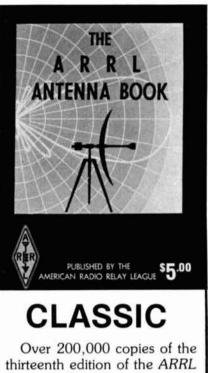
MISSOURI: The St. Charles Amateur Radio Club's Hamfest '81, August 23, Wentzville Community Club. Admission \$1.00 per car. Advance tickets: \$1.00 each/ 4 for \$3.00. At door: \$1.50 each/ 4 for \$5.00. Prizes, contests, flea market, fun for all. 'Air conditioned exhibition building. For tickets and information: SCARC Hamfest '81, c/o Bill Graham, 512 Bermuda Dr., O'Fallon, MO 63366.

NEW JERSEY: The Sussex County Amateur Radio Club's third annual hamfest, SCARC 81, Saturday, September 12, Sussex County Farm and Horse Show grounds, Plains Rd. off US Highway 206, Augusta. Outdoor felae market sellers: \$4.00 pre-registered, \$5.00 gate. Indoor sellers: \$5.00 pre-registered, \$6.00 gate. Registration: \$2.00. Door prizes. For information: Sussex County Amateur Radio Club, P.O. Box 11, Newton, NJ 07860 or Lloyd Buchholtz, WA2LHX, 10 Black Oak Drive, Vernon, NJ 07462. Talk-in on 147.90/30 and 146.52.

NEW YORK: The Suffolk County Radio Club's ARRL supported 4th annual electronic flea market, Sunday, September 13, (rain date — September 20) Odd Fellows Hall, Jayne Bivd., Port Jefferson, L.I. Gates open 7 AM. Sellers — \$3.00 — one car, one driver. Walk-Ins \$1.50. No charge for XYL's and Harmonics of attending Hams. Bargains, prizes, food and Hamship. Talk-In on 2 meters .52 and 94, 223.5 MHz. For details: Floyd Davis, 516-234-9376.

NEW YORK: HAM-O-RAMA 81, Friday, September 18, 6 PM to 9 PM, Saturday, September 19, 7 AM to 5 PM, Erie County Fairgrounds, south of Buffalo. Outside/inside Flea Markets, Equipment displays, technical programs, women's programs and more! General admission for both days: \$3.00 Advance, \$4.00 gate. Refreshments available. Talk in on W2EUP/R 146.31/91. For information: Nelson Oldfield, 126 Greenway Blvd., Cheektowaga, NY 14225, 634-6394 or Mike Merrick, 419 Sommerville Avenue, Tonawanda, NY 14150, 835-0866.





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NEW YORK: The Seaway Valley Hamfest, Saturday, September 12, Municipal Arena, Louisville, near Massena and the St. Lawrence Seaway. Activities include flea market, commercial exhibitors, movies, magic show, snack bar and more. For information: Louis G. Ierlan, WA2RXO, Secretary, Seaway Valley Hamfest Committee, 725 Proctor Avenue, Ogdensburg, NY 13669.

NEW YORK: The Hall of Science ARC's Sixth annual Hamfest, September 13, from 9 to 4, Municipal Parking Garage, one block north of Queens Boulevard, 80-25 — 126th Street, Kew Gardens. Sellers, \$3.00; Buyers \$1.00. Refreshments, free parking. Talk-in on. 52. For information: Tom Doyle, KA2DTB (212) 738-8887 or (212) 641-1700.

OHIO: The Original Forty-fourth Annual Hamfest, Sunday, September 20, 1981, at Stricker's Grove on State Route 128, one mile west of Venice (Ross) Ohio. Exhibits, prizes, food and refreshments available. Flea Market (radio related products only), music, talks, hidden transmitter hunt and sensational air show. Admission and registration \$4.00. For information: Lillian Abbott, K8CKI,317 Greenwell Road, Cincinnati, Ohio 45238.

OHIO: The 24th Annual Warren Hamfest, August 16, Kent State Branch, Trumbull County, Warren. Major prizes include: Ten-Tec Omni C with power supply; Ten-Tec Delta: Icom 2ATs. Tickets \$2.50 advance; \$3.00 gate. Banquet Saturday evening, August 15, tickets \$10.00. For information and tickets: Warren Hamfest, PO Box 809, Warren, Ohio 44482.

OREGON: The Willametta Valley DX Club's 1981 Northwest DX Convention, August 8 and 9, Greenwood Inn, Beaverton, just west of Portland. Speakers, WA4ZNH/ WB4FVU. Grand prize: ICOM-730. For information: Willamette Valley DX Club. PO Box 555, Portland, OR 97207.

RHODE ISLAND: The Bristol County Amateur Radio Association's annual indoor/outdoor flea market, Sunday, September 13, V.F.W. Hall, Tiverton. 12 noon til 4 PM. Flea market spaces, \$6.50. Admission: \$1.00. Talk-in on 147.63/03 and .52 direct. Door prizes. For info and maps: SASE to Ann M. Carro, KA1DNB, 652 Old Colony Terrace, Tiverton, RI 02878.

SOUTH DAKOTA: The 15th annul CSVHF Society Conference, July 30, 31 and August 1, Holiday Inn-Airport, Sioux Falls, Fun and activities for the entire family. For details: W@SD.

VIRGINIA: ARRL Roanoke Division Convention September 26 and 27 in the Virginia Beach, Virginia Pavillion. Free transportation to the oceanfront where the Neptune Festival is also taking place. FCC Amateur Exams given to those sending form 610 request in advance. Admission \$3.50. Advance ticket drawing for FM transceiver. Flea market tables, \$5 day, \$7 both days. TRC PO Box 7101, Portsmouth, Virginia 23707. 804-587-1695.

WEST VIRGINIA: The East River Amateur Radio Club's Hamfest '81 on Sunday, August 23, Bluefield, Armory-Civic Center, one mile north of Bluefield on US 52. Admission: \$2.00 advance, \$3.00 gate, includes prize ticket. Tables \$5.00 (\$4.00 each for 3 or more). Tailgaters \$2.00. Food, dealers, flea market, forums, entertainment something for the whole family. Talk-in 89/49, 52/52. For info: Bluefield Hamfest '81, 2113 Hemlock Hill, Bluefield, WV 24701.

QUEBEC: RAQI, Radio Amateur du Quebec's annual convention, August 1 and 2, CEGEP Levis-Lauzon, 205 mgr Bourget, Lauzon, near Quebec City. Inquiries to Jean-Marc Labarre, VE2BZL.

OPERATING EVENTS "Things to do..."

JULY 27 - AUGUST 4: Amateur Radio Station K2BSA will operate from the 1981 Boy Scout Jamboree at Fort A.P. Hill. Operation will be on 80-10 meters. See August QST for frequencies. No formal schedule. Best bet evenings. QSL card and SASE to K2BSA/4, c/o ARRL HQ, 225 Main St., Newington, CT 06111.

AUGUST 1 & 2: Micro-expedition to Forest County, Pennsylvania. Call WB3IQE/3. Mode: CW only. Frequencies: 80, 40 and 15 meters. Will operate in bottom 50 kHz of each of two bands, exact frequencies depending on conditions. Time will be spent outside the extra-only subbands. QSL to WB3IQE, Rt. 1, 3 ox 297, Brockway, PA 15824. US/Canadian stations, send SASE. DX stations, 1 IRC for QSL via ship, 2 IRCs for QSL via air.

AUGUST 15: The Englewood Amateur Radio Association's 22nd annual New Jersey QSO party. Time: 2000 UTC Saturday, August 15 to 0700 UTC Sunday, August 16; 1300 UTC Sunday, August 16 to 0200 UTC Monday, August 17. Phone and CW are considered same contest. General call "CQ NEW JERSEY" or "CQ NJ." Suggested frequencies: 1810, 3535, 3900, 7035, 7135, 7235, 14035, 14280, 21100, 21355, 28100, 28610, 50-50.5 and 144-146. Phone activity on even hours; 15 meters on odd hours

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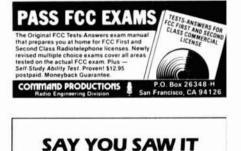
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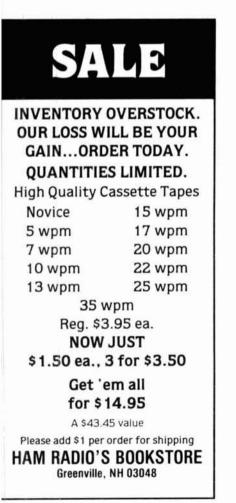
AUGUST 22: TMI FUN DXpedition: The Central Pennsylvania DX Club will hold the world's first DXpedition to Three Mile Island, the site of the Nuclear Power Facility made famous in news recently, from 1200Z Saturday August 22 until 2100Z, Sunday, August 23. Tentative frequencies: Phone - 3900, 7240, 14260, 14290, 21325, 21375, 28625 and 146,58, CW - 21125 and 7125. For an attractive photo QSL card for contacts, send SASE or IRCs to: CPDXC, c/o WB3DNA, T. Fanus, 6140 Chambers Hill Road, Harrisburg, PA 17111.

AUGUST 27: The Pend Oreille Amateur Radio Club's "Special Event Station" will be operating August 27, 28, 29 and 30 from 1600Z to 0500Z each day during the Pend Oreille County Fair at Cusick, Washington, using the Newport High School's ARC call WB7TBN. Frequencies: 14.340, 21.400, 28.700, 39.45, 37.15 CW. RTTY 28.090, 21.090, 14.080, 36.50. For a special commemorative QSL card SASE to: Mike Bice, WB7SGU, Star Rt., Box 251, Spirit Lake, Idaho 83869

SEPTEMBER 9: "CQ YL" Howdy Days. Wednesday, September 9 at 1800 UTC through Friday, September 11 at 1800 UTC. All bands and modes of emission may be used YI RI member or non-YI RI member. All licensed women operators world-wide are invited to participate. For information: Kay Eyman, WAØWOF, RR2, Garnett, Kansas 66032.

SEPTEMBER 12: "FIRE MUSTER," a special events sta-tion N@ARU@, will be on the air September 12 from 1400-2200 UTC, commemorating Burnsville, Minnesota's third annual Fire Muster. Amateur Radio in Public Service is this year's theme. Operating frequencies: 7.260, 14.285, 21.285, 28.550 MHz ± 5 kHz. Local 2 meter con tacts will be on 16/76. For a commemorative QSL certificate SASE to: NØARU, c/o David L. Justis, 14129 Frontier Lane, Burnsville, MN 55337 or via the ARRL DX Bureau.

SEPTEMBER 12 & 13: The Albuquerque, New Mexico, DX Association will sponsor a QSO party. Suggested fre-guencies: CW - 63 kHz from the low end of each band. SSB - 3900, 7265, 14285, 21365, and 28650 kHz. Novice 3705, 7105, 21105, 28105 kHz. Entries must be post-marked no later than October 15, 1981. SASE to Albuquerque DX Association, PO Box 997, Corrales, NM 37048 for complete results.





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1	1 uH \cdots 1.00 ea. or 10/7.50	12 mH
	$1.2 \text{ uH} \cdots 1.00 \text{ ea. or } 10/7.50$	15 mH
ļ	$1.5 \text{ uH} \cdots 1.00 \text{ ea. or } 10/7.50$	
	2.2 uH \dots 1.00 ea. or $10/7.50$	
	2.7 uH \dots 1.00 ea. or $10/7.50$	19.6 mH
	3.3 uH1.00 ea. or 10/7.50	20 mH2.99
ĺ	6.5 uH \dots 1.00 ea. or $10/7.50$	20.5 mH
	7.5 uH1.00 ea. or 10/7.50	22.6 mH2.99
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-	1630 uH 1.50	100 mH
· 1	.1 mH	120 mH
ļ	.2 mH	150 mH
:	. 22 mH 2. 99	175 mH
-	. 27 mH	200 mH
1	. 33 mH 2. 99	205 mH
	. 39 mH 2. 99	237 mH
"	. 240 mH 2. 99	240 mH
"	1.2 mH	300 mH
	1.5 mH2.99	360 mH
"	1.65 mH2.99	390 mH
"	1.75 mH2.99	430 mH
	1.9 mH2.99	500 mH
- {	1 mH1.69	600 mH
"	1.88 mH	1000 mH
	2 mH 2.99	1.5 Hy2.99
	2.4 mH	2.0 Hy
•	2.5 mH \dots 1.00 ea. or 10/7.50	2.5 Hy
00	2.7 mH	3.0 Hy
~	3.0 mH	5.0 Hy
_	3.6 mH	10 Hy
	4.3 mH	10 11y
RS	7. 0 IIII ·······························	۱ - <u>محمد محمد محمد محمد محمد محمد محمد محم</u>
	HIGH VOLTAGE CAPS	New Fairchild Prescaler Chip
77	420 MFD @ 400 VDC 3.99 each	95H90DCQM 6.50 each
99	600 MFD @ 400 VDC 3.99 each	350 MHz prescaler divide by 10/11
99		

ohnson AIR Variables

Transistors

 $1/4 \ge 2 \frac{1}{2}$ shaft \$2.50 each 193-10-6 2.2 to 34 pF 1.5 to 27.5pF 193-193-.6 to 6.4pF

	\$1.00 each	
160-107-16	.5 to	12 pF
193-10-9		34 pF
193-10-104	2 . 2 to	34 pF
193-4 -5	3 to	30 pF

Power Device RF

MRF454 Same as MRF458 12.5 VDC, 3-30 MHz 80Watts output, 12dB gain \$17.95 ea.



#124-0311-100 6.99 each For 8072 etc.

#124-0107-001..... 13.99 each For 4CX250B/R, 4X150A etc.

#124-0111-001..... 4.99 each Chimney for 4CX250B/R and 4X150

#124-0113-001 and 124-0113-021 \$12.99 each Capacitor for #124-0107-001

#123-209-33 Sockets....6.99 each For 811A, 572B, 866, etc.



43pF	350V	\$1.00 each
36pF	1000pF	
33pF	470pF	
24pF	380pF	
20pF	240pF	
14pF	200pF	
13pF	180pF	
1 2 pF	160pF	
10pF	100pF	
8. 2pF	62pF	
6.8pF	47pF	

86 Pin Motorola Bus Edge Connectors

Gold plated contacts	
Dual 43/86 pin . 156 spacing	
Soldertail for PCB	\$3.00 each

110VAC MUFFIN FANS

New															•					\$11.95
Used	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	•	. \$5.95

View Name State S						513					
2N3959 3.00 2N5643 14.00 40282/2N3927 10.48 CRYSTALS \$4.95 each 5.120 7.4825 9.565 10.150 11.155 11.905 17.315 7.4335 7.4865 9.575 10.160 11.275 11.905 17.355 7.4585 7.4925 9.585 10.170 12.000 17.365 7.4665 7.5025 10.020 10.240 11.730 12.100 37.600 7.4685 7.5065 10.030 10.595 11.755 17.015 37.750 7.4725 7.8025 10.040 10.605 11.800 17.065 37.800 7.4725 9.545 10.030 10.625 11.850 17.15 37.900 7.4815 9.545 10.140 10.635 11.900 17.265 37.900 7.4815 9.555 10.140 10.635 11.900 17.265 37.900 7.4815 9.500 VDC 1.69 10.925 11.900 17.265 37.900 10.60 1	2N2949 2N2947 2N2950 2N3375 2N3553 2N3818 2N3866 2N3866JA 2N3866JA 2N3866JA	3. 15. 4. 8. 1. 5. 1. N 2. NTX 4. 10.	60 00 60 57 00 00 50 00 00 00	2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2	4072 4427 4429 4877 4959 4976 5070 5071 5108 5109 5179 5583 5589 5590 5591 5635 5636	NTX	$\begin{array}{c} 1.\ 60\\ 1.\ 10\\ 7.\ 00\\ 1.\ 00\\ 2.\ 00\\ 15.\ 00\\ 8.\ 00\\ 15.\ 00\\ 4.\ 00\\ 1.\ 50\\ 1.\ 00\\ 4.\ 00\\ 6.\ 00\\ 8.\ 00\\ 11.\ 00\\ 5.\ 44\\ 11.\ 60\\ \end{array}$	2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2N 2	15842 15849 15946 15946 15862 16080 16081 16082 16083 16084 16095 16096 16096 16097 16166 16368 210/MRF5 2138		$\begin{array}{c} 8.00\\ 20.00\\ 40.00\\ 14.00\\ 50.00\\ 7.00\\ 10.00\\ 11.00\\ 13.00\\ 14.00\\ 11.00\\ 20.00\\ 28.00\\ 38.00\\ 22.99\\ 2.00\\ 5.00\\ \end{array}$
Start Start Start as a St05 but only 1/2 Amp							5.00				
5. 1207. 48259. 56510. 15011. 15511. 90517. 3157. 34357. 48659. 57510. 16011. 27511. 90517. 3157. 45857. 498510. 00010. 18011. 70012. 00017. 3657. 46157. 498510. 00010. 18011. 70512. 05037. 6007. 46257. 501510. 01010. 24011. 75016. 96537. 7007. 46257. 506510. 03010. 59511. 75517. 01537. 7007. 46857. 506510. 03210. 62511. 85017. 16537. 8007. 47859. 54510. 13010. 62511. 85017. 16537. 8007. 47859. 55510. 13010. 62511. 85017. 16537. 9007. 47859. 55510. 14010. 63511. 90017. 26537. 9007. 47859. 55510. 14010. 63511. 90017. 26537. 9007. 47859. 55510. 14010. 63511. 90017. 26537. 9007. 48159. 55510. 14010. 63511. 90017. 26537. 90030 MFD @ 450 VDC2. 2922to 30pF30 MFD @ 450 VDC2. 293. 9 to 18pF30 MFD @ 450 VDC2. 2930 01 @ 2 KV3. 90 each9.01 @ 1. 6KV4. 1009.02 @ 8 KV2.00.01 @ 1. KV6/1.00NEW 2'' ROUND SPEAKERS100 Ohm coil\$. 99 each	2N3959	3.	00					40	282/2N39	27	10.48
5. 120 7. 4825 9. 565 10. 150 11. 155 11. 905 17. 315 7. 3435 7. 4865 9. 575 10. 160 11. 275 11. 905 17. 315 7. 4485 7. 4985 10. 000 10. 180 11. 700 12. 000 17. 365 7. 4615 7. 4985 10. 010 10. 240 11. 700 12. 000 37. 650 7. 4625 7. 5015 10. 020 10. 245 11. 750 16. 965 37. 700 7. 4625 7. 5025 10. 030 10. 595 11. 750 16. 965 37. 700 7. 4665 7. 5025 10. 040 10. 605 11. 850 17. 105 37. 750 7. 4715 7. 7985 10. 0525 10. 615 11. 850 17. 125 37. 900 7. 4785 9. 555 10. 140 10. 635 11. 900 17. 265 37. 950 30 MFD @ 500 VDC 1. 69 17. 215 37. 950 38. 000 30 MEW & Y 2.00 2.9 2.00 30F 3.9 to 139F 225 MFD @ 450 VDC 2.99 2.1 c 2 to 139F		_		<u> </u>							
30 MFD @ 500 VDC 1.69 22 MFD @ 500 VDC 1.69 100 MFD @ 450 VDC 2.29 150 MFD @ 450 VDC 3.29 25 MFD @ 450 VDC 4.29 .001/1000pF @ 10 KV .89 .001 @ 2 KV 4/1.00 .01 @ 1.6KV .79 .01 @ 1.6KV .79 .01 @ 1.6KV .79 .01 @ 1.6KV 4/1.00 .02 @ 8 KV 2.00 .01 @ 1 KV 6/1.00 NEW 2" ROUND SPEAKERS 00 100 Ohm coil \$.99 each J.7 MHz arrow band 3 dB bandwidth 15 KHz min. .01 dB bandwidth 15 KHz min. .20 dB bandwidth 15 KHz min. .02 dB bandwidth 15 KHz min. .20 dB bandwidth 15 KHz min. .02 dB bandwidth 15 KHz min. .500 .03 dB bandwidth 15 KHz min. .60 .40 dB bandwidth 15 KHz min. .60 .04 db bandwidth 15 KHz min. .50 .05 H455D 455 KHz .50 .83.99 each \$3.99 each .78MO5 \$3.99 each 78MO5 \$30 but only 1/2 Amp	7.3435 7.4585 7.4615 7.4625 7.4665 7.4685 7.4685 7.4715 7.4725 7.4765 7.4785	7.4865 7.4925 7.4985 7.5015 7.5025 7.5065 7.7985 7.8025 9.545	9.5 9.5 10.0 10.0 10.0 10.0 10.0	75 85 000 010 020 030 040 525 130	10.19 10.10 10.11 10.24 10.24 10.59 10.60 10.6	50 50 70 80 40 45 55 55 55 25 35	11. 155 11. 275 11. 700 11. 705 11. 730 11. 750 11. 755 11. 800 11. 850 11. 855 11. 900		11.955 12.000 12.050 12.100 16.965 17.015 17.065 17.165 17.215 17.265	17 17 31 31 31 31 31 31 31 31 31 31 31	7.355 7.365 7.600 7.650 7.700 7.750 7.800 7.850 7.850 7.900 7.950
30 MFD @ 500 VDC1.6922 MFD @ 500 VDC1.69100 MFD @ 450 VDC2.29150 MFD @ 450 VDC3.29225 MFD @ 450 VDC4.29.001/1000pF @ 10 KV.89.001 @ 2 KV4/1.00.01 @ 1 KV.79.01 @ 1 KV.79.01 @ 1 KV6/1.00NEW 2" RO UND SPEAKERS100 Ohm coil\$.99 eachPLASTIC TO-3 SOCKETS $4/$1.00$.07 MLz narrow band3 dB bandwidth 15 KHz min.20 dB bandwidth 15 KHz min.21 dB max.Ripple 1 dB max.St 9 each78MO520 Sb but only 1/2 Amp	High	Volta	ige	Cá	aps		RIM	M	ER (PS
100 Ohm coil\$.99 eachPLASTIC TO-3 SOCKETS 4/\$1.00J310 N-CHANNEL J-FET 450 MHz Good for VHF/UHF Amplifier, Oscillator and MixersCRYSTAL FILTERS Tyco 001-19880 Same as 2194F 10.7 MHz narrow band 3 dB bandwidth 15 KHz min. 20 dB bandwidth 150 KHz min. 40 dB bandwidth 150 KHz min. Ultimate 50 dB insertion loss 1 dB max. Ripple 1 dB max. Ct. 0+/-5 pF 3600 Ohms \$3.99 eachMURATA CERAMIC FILTERS SFD 455D 455 KHz 2.00 SFB 455D 455 KHz 1.60 CFW455E 455 KHz 5.50 CFU 455H 455 KHz 3.00 SFE 10.7MA 10.7 MHz 2.9978MO5 Same as 7805 but only 1/2 AmpTEXAS INSTRUMENTTIL-305P 5 x 7 array alphanumeric display	22 MFD 100 MFD 150 MFD 225 MFD .001/1000 .001 @ .0015 @ .01 @ .01 @ 1 .02 @	 @ 500 VD@ @ 500 VD@ @ 450 VD@ @ 450 VD@ @ 450 VD@ 0pF @ 10 1 2 KV 3 KV 4 KV .6KV 8 KV 		4/ 3/ 4/	1.69 1.69 2.29 3.29 4.29 .89 1.00 1.00 .79 1.00 2.00		. 5 arbide Ci or PCB B	0 ea not 1. 2 3. 3. 3. 3. 1 rcu 5 m	ch or 10/4 sold mixe 2 to 13pF to 30pF 9 to 18pF 9 to 40pF 9 to 55pF it Board L is ix for \$5.4	4. 00 d)
J310 N-CHANNEL J-FET 450 MHzPLASTIC TO-3 SOCKETS 4/\$1.00CRYSTAL FILTERS Tyco 001-19880 Same as 2194F 10.7 MHz narrow band 3 dB bandwidth 150 KHz min. 20 dB bandwidth 150 KHz min. 40 dB bandwidth 150 KHz min. Ultimate 50 dB insertion loss 1 dB max. Ripple 1 dB max. Ct. 0+/-5 pF 3600 Ohms \$3.99 eachJ310 N-CHANNEL J-FET 450 MHz Good for VHF/UHF Amplifier, Oscillator and Mixers 3/\$1.00MURATA CERAMIC FILTERS SFD 455D 455 KHz 2.00 SFB 455D 455 KHz 1.60 CFW455E 455 KHz 5.50 CFU 455H 455 KHz 3.00 SFE 10.7MA 10.7 MHz 2.9978MO5 Same as 7805 but only 1/2 Amp			PEAK					J -	Fet		
10.7 MHz narrow band 3 dB bandwidth 15 KHz min. SFD 455 LHZ 2.00 20 dB bandwidth 150 KHz min. SFB 455D 455 KHz 1.60 40 dB bandwidth 150 KHz min. SFB 455D 455 KHz 1.60 Ultimate 50 dB insertion loss 1 dB max. CFM 455E 455 KHz 5.50 Signe as 7805 but only 1/2 Amp TEXAS INSTRUMENT TIL-305P 5 x 7 array alphanumeric display	PLASTIC TO-3 SOCKETS 4/\$1.00					Go Os	Good for VHF/UHF Amplifier,				
Same as 7805 but only 1/2 Amp 5 x 7 array alphanumeric display	10.7 MHz narrow band 3 dB bandwidth 15 KHz min. 20 dB bandwidth 60 KHz min. 40 dB bandwidth 150 KHz min. Ultimate 50 dB insertion loss 1 dB max. Ripple 1 dB max. Ct. 0+/-5 pF 3600 Ohms \$3.99 each					SI SI C C	FD 455D FB 455D FM455E FU455H FE 10.7M	A	455 KHz 455 KHz 455 KHz 455 KHz 10. 7 MHz		2.00 1.60 5.50 3.00 2.99
	Same as						5 x 7 arra			disp	olay

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ARCO CAPS

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304 100-550pF	1.50	469	170-780pF	1.40		
400 .9-7pF	1.00	4615	390-1400pF	2.02		
402 1.5-20pF	1.00	404	8-60pF	1.00		
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423 7-100pF	1.00	422	4-40pF	1.00		
426 37-250pF	1.01	424	16-150pF	1.00		
464 25-280pF	1.00	427	55-300pF	1.00		
465 50-380pF	1.39	462	5-80pF	1.50		
467 110-580pF	1.03		-			
-	TU	DEC				
		BES				
6KD6	5.00	6939		7.99		
6LQ6/6JE6	6.00	6146		5.00		
6MJ6/6LQ6/6JE6C	6.00	6146A		5.69		
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811A	20.00	3-500Z		141.00		
6360	4.69					
	RF Tra	nsistors				
		11515(015				
\sim	MRF449	12.65	BFR91	1.25		
	MRF449A		BFR96	1.50		
	MRF450	11.00	BFW92A	1.00		
	MRF450A	11.77	BFW92	. 79		
	MRF452	15.00	MMCM918	14.30		
B	MRF453	13.72	MMCM2222	15.65		
MRF203 P.O.R.	MRF454A		MMCM2369	15.00		
MRF216 19.47	MRF455	14.08	MMCM2484	15.25		
MRF221 8.73	MRF455A	14.08	MMCM 3960A	24.30		
MRF226 10.20	MRF474	3.00	MWA120	7.80		
MRF227 2.13	MRF475	2.90	MWA 130	8.08		
MRF238 10.00	MRF476	2.25	MWA210	7.46		
MRF240 14.62	MRF477	10.00	MWA 220	8.08		
MRF245 28.87	MRF485	3.00	MWA230	8.62		
MRF247 28.87	MRF492	20.40	MWA 310	8.08		
MRF262 6.25	MRF502	. 93				
MRF314 12.20	MRF604	2.00	NEW MRF472			
MRF406 11.33	MRF629	3.00	12.5 VDC, 27			
MRF412 20.65	MR F648	26.87	4 Watts output	t		
MRF421 27.45	MRF901	3.99	10 dB gain	_		
MRF422A 38.25	MRF902	9.41		1.69 ea.		
MRF422 38.25	MRF904	3.00		10/9.50		
MRF428 38.25	MRF911	4.29		0/69.00		
MRF428A 38.25	MRF5176	11.73	1000)/480.00		
MRF426 8.87	MRF8004	1.39				
MRF426A 8.87	BFR90	1.00	1			
TO-3 TRANSISTOR SOC Phenolic type			PL259 TERMINATION 52 Ohm 5 Watts \$1.50 each			
NEW SIMPSON 260-7	\$99.99	TORIN TA7	00 FANS NEW \$2	9.99 each		
		Model A 30	340			
RG174/U - \$15.00 per 1 Factory new	ου π.	230 VAC @ Will also v).78 Amps vork on 115 VAC			

CRYSTAL FILTERS

EFCL455K13E	3.99
EFCL455K40B2	2.99
FX-07800L, 7.8 MHz	12.99
FHA 103-4, 10.7 MHz	12.99
·	

CB type crystals

	-7 P C	
	\$4.95 ea	ach
	51-T	
ГІ	T 15	Т28
Т2	T 16	Т 29
Т3	T17	Т 30
Т4	T18	T31
Т5	T19	Т 32
Т6	Т20	Т 33
Т7	T21	Т 34
Т8	Т22	Т 35
т9	Т23	Т 36
T10	Т24	T 37
T11	Т25	Т38
T12	Т 26	Т 39
T13	T2 7	Т40
T14		
	51-R	
R1	R15	R28
R2	R16	R29
R3	R17	R30
R4	R18	R31
R5	R19	R 32
R6	R20	R33
R7	R 2 1	R 34
R8	R22	R35
R9	R23	R36
R 10	R24	R37
R11	R 25	R38
R12	R 26	R39
R13	R27	R40
R14		
	HERRY BCD	SWITCH
New end		1.29 each
	ohn	c o p
J	ohns	S U II
		ishlac
	var	iables
	\$1.00 e	ach
T-3-5	•	l to 5 pF
T-6-5		1.7 to 11 pF
T-9-5		2 to 15 pF
189-6-		.1 to 10 pF
189-502		1.3 to 6.7pF
189-503-		1.4 to 9.2pF
189-504	4-5	1.5 to 11.6pF
189-505		1.7 to 14. lpF
189-505-		1.7 to 14.1pF
189-506-		1.8 to 16.7pF
189-507-		2 to 19.3pF
189-50		2.1 to 22.9pF
189-50	9-5	2.4 to 24.5pF
545-0	43	1.8 to 11.4pF

1 0-2 5	C	RTERS
1.9-2.5		NILNS

1900 MHz to 2500 MHz DOWNCONVERTERS
Intended for amateur radio use.
Tunable from channel 2 thru 6.
34 dB gain 2.5 to 3 dB noise.
Warranty for 6 months Model HMR 11
Complete Receiver and Power Supply
(does not include coax) \$225.00
4 foot Yagi antenna only \$39.99
Downconverter Kit - PCB and parts \$69.95
Power Supply Kit -
Box, PCB and parts \$49.99
Downconverter assembled \$79.99
Power Supply assembled \$59.99
Complete Kit form\$109.99
(includes Yagi antenna and instructions)
REPLACEMENT PARTS
MRF901\$ 3.99
MBD101 1.29
.001 Chip Caps 1.00
Power Supply PCB 4.99
Downconverter PCB 19.99
Instructions for any separate item 10.00

NEW TRANSFORMERS

		Price each
F-18X	6.3 VCT @ 6Amps	6.99
F-46X	24V @ 1Amp	5.99
F41X	25.2VCT @ 2Amps	6.99
P-8380	10VCT @ 3Amps	7.99
P-8604	20VCT @ 1Amp	4.99
K-32B	28VCT @ 100 MA	4.99
E30554	Dual 17V @ lAmp	6.99

DIODES

	#43 Shield Bead			
HEP 170 3.5 A, 1000 PIV .20 ea., 100 for \$15.00	High-voltage diode EK500 5000 Volts, 50 mA .99 each	#61 Toroid #43 Balun #61 Balun #61 Balun		
D61005 1.5 A, 1000 PIV .15 ea., 100 for \$12.00	Motorola SCR TO-92 Case, 0.8 Amp, 30 V. lgt 0.2 Vgt 0.8. Same as #NS060.	#61 Balun #61 Beads Ferrite Rod 1/4 x 7 1/2		
HVK 1153 25 mA, 20,000 PIV \$1.00 ea., 10 for \$8.00	4/\$1.00 or 100/\$15.00 Dialco Type 555-2003	Ferrite Beads 1/8" long Ferrite Beads 3/8" long Ferrite Beads 1/16" long		
Fairchild LEDs FLV 5007 & 5009 red. Case type TO-92.	LED 5 VDC with built-in resistor. .69 each	DOOR KNOB CAPS 470 pF @ 15 KV		
6/\$1.00 SCMS 10K 15 mA, 10,000 PIV \$1.69 ea., 10 for \$12.50	Motorola MA 752 Rectifier 6 Amps, 200 PIV 4/\$1.29	Dual 500 pF @ 15 KV 680 pF @ 6 KV 800 pF @ 15 KV 2700 pF @ 40 KV		

ORDERING INSTRUCTIONS

Check, money order, or credit cards welcome. (Master Charge and VISA only.) No personal checks or certified personal checks for foreign countries accepted. Money order or cashiers check in U.S. funds only. Letters of credit are not acceptable. Minimum shipping by UPS is \$2.35 with insurance. Please allow extra shipping charges for heavy or long items.

All parts returned due to customer error or decision will be subject to a 15% restock charge. If we are out of an item ordered, we will try to replace it with an equal or better part unless you specify not to, or we will back order the item, or refund your money

PRICES ARE SUBJECT TO CHANGE WITHOUT NOTICE. Prices supersede all previously published. Some items offerd are limited to small quantities and are subject to prior sale. We now have a toll free number, but we ask that it be used for charge orders only. If you have any questions please use our

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544-4001-002, similar to type MHW 401-2 1.5 Watts output. 440-512 MHz.

TRANSFORMERS

26.8 VCT @ 660 MA

21.9 VCT @ 1.1 Amps

#2899652-01

#18000711P

#2099459-00

24 V @ 100 MA

28 V @ 1.5 Amps

9.6V@9 Amps

16.8 V @ 300 MA

JUMBO LED'S

MEDIUM LED'S

NE555V TIMERS

.39 each or 10/\$5.00

NEW DUAL COLON LED

TRANSISTORS/IC S

Motorola MHW 252 VHF power amplifier.

EC9814. 2-W audio amplifier. \$1.29 ea., 10 for \$9.50

ECG no. 707 Chroma demodulator. \$1. 29 ea., 10 for \$8.50

Selection Guide & Cross-Reference

.69 each or 10/\$5.00

PLATE CHOKES

Frequency range: 144-148 MHz. Output power: 25W. Minimum gain: 19.2 dB.

House no. same as HEP C6073 & EC9814.

Red

Clear

Yellow

Green

Amber

Red

Green

75 uH

.94 mH

Motorola MC 1316P

Fairchild 007-03 IC.

Catalog.

RCA Triacs

Sensitive gate.

43 pgs.

Motorola rf transistors.

Type T2310A. TO-5 Case with heat sinks

RCA nower transistors. IPN RCS 258

Vceo 60 NFE 5mA

RCA Triacs.

RCA Triacs.

Stud type.

IC 20 Amps Vce 4V. 250 Watts, Ft 2 MHz.

Type T4121B/40799, 200 VDC 10 Amps.

Type 40805/T6421D.

30 Amps, 400 VDC.

15 dB gain min.

Motorola rf amplifier.

1.6 Amp, 100 VDC, lgt 3mA.

\$9.99 each

\$1.99 each

\$12.99 each

8/\$1.00

6/\$1.00

6/\$1.00

6/\$1.00

6/\$1.00

6/\$1.00

6/\$1.00

3.00

3.99

\$29.67 each

\$1.99 each

\$1.00 each

\$3.00 each

\$3.69 each

\$5.00 each

\$19,99 each

NEW BCD SWITCH 8 switch with end plates

36 in. long x 1/2 in.

MAGNET WIRE

#24

#26

#25

#30

#31

CORES

T 20-12

T25-6

T 30 - 2

#/T - 18R

CV31D350 HM00-4075-03

CERAMIC STAND OFFS

CORES AND BEADS

#NL523W03-010 3/4 x 1 1/4"

300425

E5-25A

#CNP-5

#N54W0112

CABLE TIES

Model TSM 200-1011 (CDI) \$16.87

\$4.99 each

9 lb.

9 lb.

9 lb.

6 lb.

8 3/4 lb.

T37-6

T37-10

T44-6

100 per bag

2 to 8 pF

3.5 to 11 pF

3.5 to 13 pF

5 to 25 pF

5.1 to 40 pF

3.5 to 15 pF

5.2 to 40 pF

2.5 to 6 pF

. 29 each

.39 each

.49 each

. 79 each

4/1.00

3/1.00 10/1.00

8/1.00

6/1.00

4/1.00

2.99

10/1.00

12/1.00

6/1.00

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HAM CALENDAR

Sunday	Monday	Tuesday	Wednesday	Thursday	Friday	Saturday
						CENTRAL STATES VHF SOCIETY'S 15TH ANNUAL CON- FERENCE – Holiday Inn Aliport, Sloux Falls, SD ~ W05D 1 ILLINOIS GSO PARTY 12. MICRO-EXPEDITION IS Forest County, PA. Call W830[6/3, Mode CW only Frequencies 30, 40, and 15 meters – W830E 1-2. STRD ANNUAL UPPER PENISULA HAMFEST (MICHIGAN) – Datia County Repeater Assoc, Fiel Rock Township Hall, Escaraba, MI – WA8DHB 12. JACKSONVILLE HAMFEST & NORTHERN FLORIDA ARRL CONVENTION – Orange Park Kennel Club, I-295 G US 17 south of Jacksonville 1.2. RADIO AMATEVIA DE QUEBEC – Annue convention at CEGEP Levis Lauton, 200 Mgr Bourget, Lauton near Duebec City, Call V2282L 48577-2860 - L.
ZRD ANNUAL CROOKED LAKE FM PICNIC & HAMFEST - by Sieuben Courty Radio Amareus, Angoia, IN - WB911 2 NORTHWEST KANSAS IST AMATEUR RADIO SWAP MEET - community Bldg. Color, KS - WARGEN or KARFBG 2 ILLINGIS BELYTIORFE BIG THUNDER ARC ANNUAL HAMFEST - Boone County Fairgrounds, Hwy 76 KBOGG 2 SRD ANNUAL NORTH CENTRAL MISSOURI HAMFEST Municoal Auditorium, Moberly, MD - WBBENV 2.	WEST COAST BULLETIN B PM PST 10400 UTC: 2540 NHz A 1 22 WPM 3	AMSAT Eastcoast Net 3860 vHz 9:00 PM EDST (01002 Wednesday Morring) AMSAT Mid-Continent Net 3850 vHz 9:00 PM CDST (02002 Wednesday Morring) AMSAT Westcoast Net 3850 vHz 8:00 PM PDST (03002 Wednesday Morring)	5	West coast Qualifying RUN 6.	7	EUROPEAN C.W. CONTEST 8.9.* UHF CONTEST – Begins 1900 UTC Sat and ends at 1900 UTC on Sivn 8.9 NORTHWEST DX CONVENTION – by Wilamette Valley DX Club, The Greenwood Inn, Beaverran, OR Write POB 556. Portinal, DA 92/07 8.9 EURLINGTON ARC ANNUAL INTERNATIONAL HAMFEST – Old Lattere Canground, Of Tours 7, Charotner, VT – VIVSA 8.3. TRI-COUNTY AMATEUR RADIO ASSN. ANNUAL HAMFEST – Los Angeles County Fairgrounds, Thummer's Patio, Write TCARA, POB 142, Pamona, CA 91767 8.
ST. CLOUD AMATEUR RADIO CLUB HAMFEST — Whitney Sr. Ceniw, St. Coud, MN. Contact Make Lynch, KABHOS 9. CENTRAL KENTUCKY ARRL HAMFEST — by the Bluegrass Amateur Radio Society at Tates Cuesk Jr. H S. Centre Parkwar, Leangion, KY – KACIN 9. HAMFESTERS RADIO CLUB 5 47TH ANNUAL HAMFEST Santa Fe Park, 91: B Woll Rd. Willow Springs. IL NBBIA 9.		AMSAT Eastcoast Net 3850 kHz 9:00 PM EDST (01002 Wednesday Morning) AMSAT Mid-Continent Net 3850 kHz 9:00 PM CDST (02002 Wednesday Morning)		WIAW QUALIFYING RUN 13	TEXAS VHF FM SOCIETY STATE CONVENTION 5 2ND ANNUAL SUPER CEN- TRAL TEXAS SWAPFEST – Hitton Inn, Austin, TX. Con- tact VHF 781, POB 13473, * Capital Sta, Austin, TX. 78711 14-16	ALL ASIAN CW CONTEST 22.22°., RHODE ISLAND QSO PARTY 15-17. RADIO CLUB OF TACOMA ANNUAL HAMFAIR — Pacific Luheran University. Tacoma. WA — WB7ONS 15-16. NEW JERSEY QSO PARTY 15-17.
9	10	AMSAT Westcoast Net 3850 kHz 8:00 PM PDST (0300Z Wednesday Morning)	12	13	14	15
SIXTH ANNUAL NEW DELMARVA HAMFEST — Granyland Park, Bear, D.E. Contact K3HBP, 14 Balisam R.E., Wilmington, DE 19904 16. IDTH ANNUAL LAFAYETTE HAMFEST — Trippecance County Feargrounds in Lafayette, IN 16.	WEST COAST BULLETIN - 8 PM PST (000 UTC) 350 LHz, A-1 22WPM 17	AMSAT Eastcoast Net 3850 kHz 9:00 PM EDST (01002 Wednesday Morring) AMSAT Mid-Continent Net 3850 kHz 9:00 PM CDST (02002 Wednesday Morring)				OHID OSO PARTY - 22-23. TMI FUN DAEDITION - Central PA DX Club's first DXpedition to Three Mile Island. OSL to CPDXC, c/o WB3DNA 22-23.
16	17	MIGRING) AMSAT Westcoast Net 3850 kHz 8:00 PM PDST (0300Z Wednesday Morning)	19	20	21	22
WIAW QUALIFYING RUN 22. ST CHARLES ARC HAMFEST – Wentzville Missouri Community Cub. Context SCARC Hamless II, <i>cir</i> Bitl Graham, 512 Ber- muda Dr., 075alon, MO SG56 23. BULFFLD HAMFEST BI – 50¢ East River ARC at Bluefield Army Cvic Center, 1 m north of Bluefield, WV on US 52 Wirte Bluefield Hamlest, 2113 Hemlock Hill, Bluefield, WV 24701 23. 1981 ILLINOIS STATE ARR. CONVENTION – by Fox River		AMSAT Eastcoast Net 3850 kHz 9:00 PM EDST (01002 Wednesday Morning) AMSAT Mid-Continent Net 3850 kHz 9:00 PM CDST (02002 Wednesday Morning)		PEND ORFILLE ARC "SPECIAL EVENT STA. TION" - fram Pend Oreite County Fairgrounds at Cusce, WA, on Bool to 66020 uning call W87TBN 27-30.		
Radio League, Kane County Fairgrounds in St. Charles, IL 23. 1004 75 METER NET PICNIC & SWAPPEST - Ewing park in southeast Des Moines, I.A. Contact WBBUFF 23.	24	AMSAT Westcoast Net 3850 kHz 8:00 PM PDST (0300Z Wednesday	26	27	28	29
SHAWNEE AMATEUR RADIO ASSOC. 25TH ANNUAL HAMFEST — ALJohn A Logan College west of Canerville, IL on Salla Ri, 13. 30. IDULESTER COUNTRY ARC HAMFEST — ALGeucesian Country College, Tanyard, Rd. Sevell, NJ. Contact Hamfest Committee, POB 370. Priman. NJ 08071. 30. APORTE-COUNTY HAMFEST — County Fargrounds on Hwy		Morning)				W1AW Schedule April 26-October 25, 1981 UTC Slow Code Practice MWF: 0200, 1300, 2300; Fisst Code Practice MWF: 2000; TTh S.Sn. 2000; Siss Code Practice MWF: 2000; TTh; 0200, 1300; TTh; 0200;
2 west of LaPore. IN 30.	31					CW Bulletins Dy: 0000, 0300, 2100; MTWThF: 1400 Code practice and CW bulletin frequencies: 1.835, 3.58, 7.08, 14.08, 21.08, 28.08, 50.08, 147.555 MHz.

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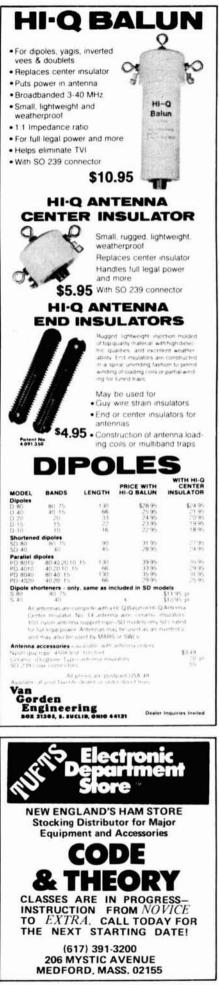
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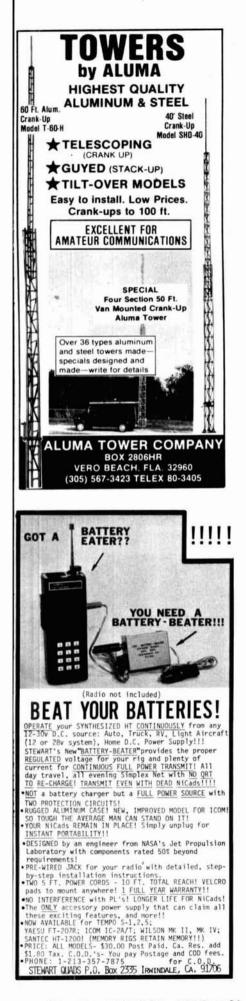
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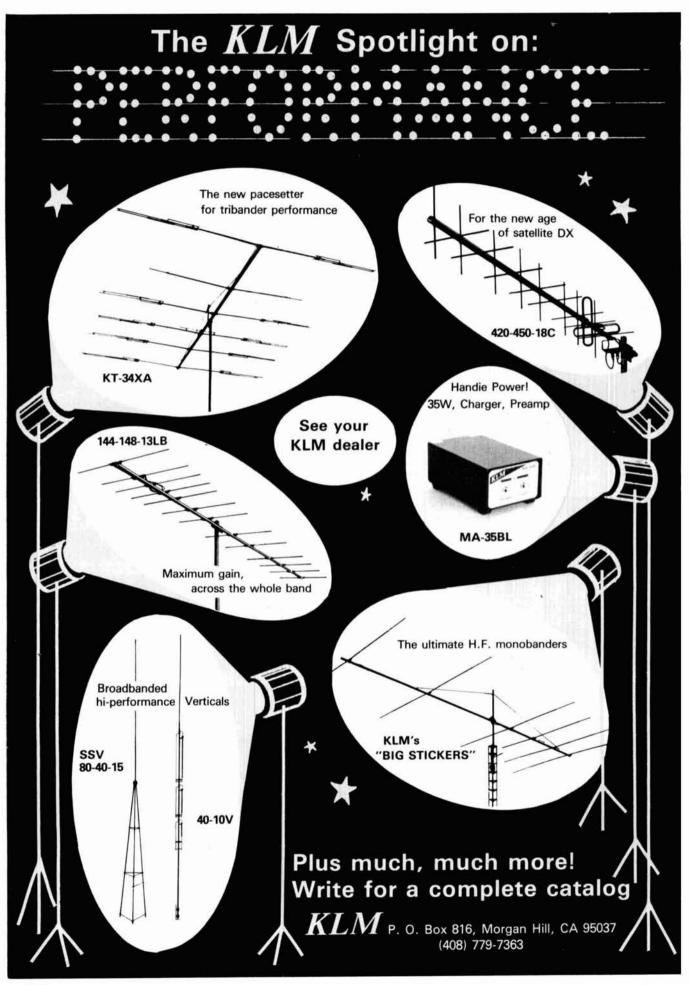


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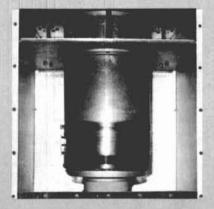
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